

# Antenne

Corso di Laurea in Ingegneria Informatica, Biomedica e delle  
Telecomunicazioni

a.a. 2023–2024 – Laurea “Triennale” – Secondo semestre – Terzo anno

**Università degli Studi di Napoli “Parthenope”**

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# Color legend

New formulas, important considerations,  
important formulas, important concepts

Very important for the discussion

Memo

Mathematical tools to be exploited

Mathematics

# Antenna Parameters

Parameters of the Tx Antenna

Parameters of the Rx Antenna

# Parameters of the Tx Antenna

- Effective length
  - Radiation pattern
  - Radiation pattern lobes
  - Beamwidth
- Directivity
- Gain
- Radiation Resistance
- Equivalent circuit of the tx antenna
- Input Impedance and Input Resistance



# Effective Length

$$r \gg D$$

$$r > \frac{2D^2}{\lambda}$$

$$r \gg \lambda$$

$$\mathbf{E}(\vec{\mathbf{r}}) = \mathbf{E}(r, \vartheta, \varphi) = \frac{j\zeta I e^{-j\beta r}}{2\lambda r} \mathbf{l}(\vartheta, \varphi)$$

$$\zeta \mathbf{H}(\vec{\mathbf{r}}) = \hat{i}_r \times \mathbf{E}(\vec{\mathbf{r}})$$

$$\mathbf{l}(\vartheta, \varphi) = l_g(\vartheta, \varphi) \hat{i}_g + l_\varphi(\vartheta, \varphi) \hat{i}_\varphi$$

$$\vec{\mathbf{S}} = \frac{1}{2\zeta} |\vec{\mathbf{E}}|^2 \hat{i}_r = \frac{\zeta}{2} |\vec{\mathbf{H}}|^2 \hat{i}_r$$

$$\mathbf{l}(\vartheta, \varphi) = l_g(\vartheta, \varphi) \hat{i}_g + l_\varphi(\vartheta, \varphi) \hat{i}_\varphi \quad \text{effective length of the antenna}$$

# Antenna Parameters

Parameters of the Tx Antenna

**Parameters of the Rx Antenna**

# Receiving mode

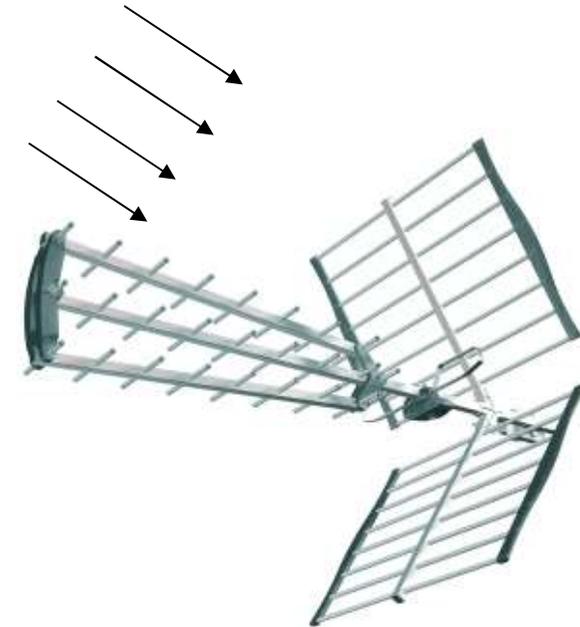
When an antenna is operating as a receiving antenna, it extracts a certain amount of power from an incident electromagnetic wave.

Since an incident wave comes from a far distance may be thought of as a uniform (local) plane wave being intercepted by the antenna.

The use of the antenna in the receiving mode is shown in Figure.

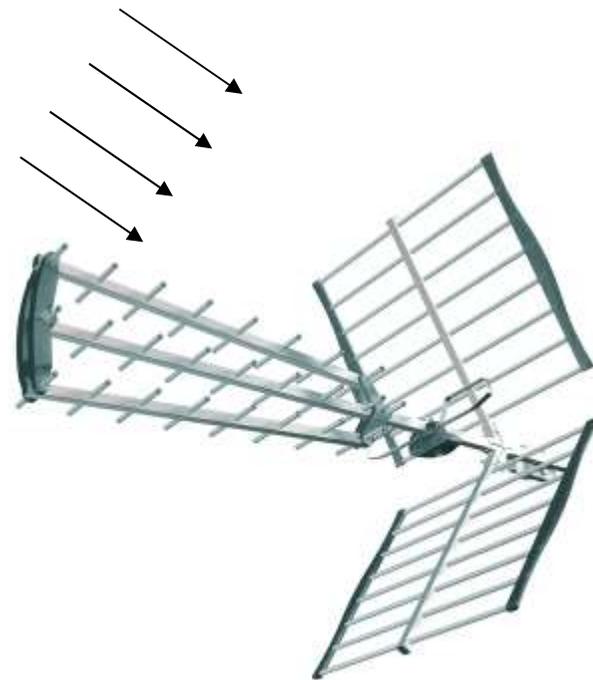
The incident wave impinges upon the antenna, and it induces a voltage  $V_0$  at the input terminals.

**Incident wave**



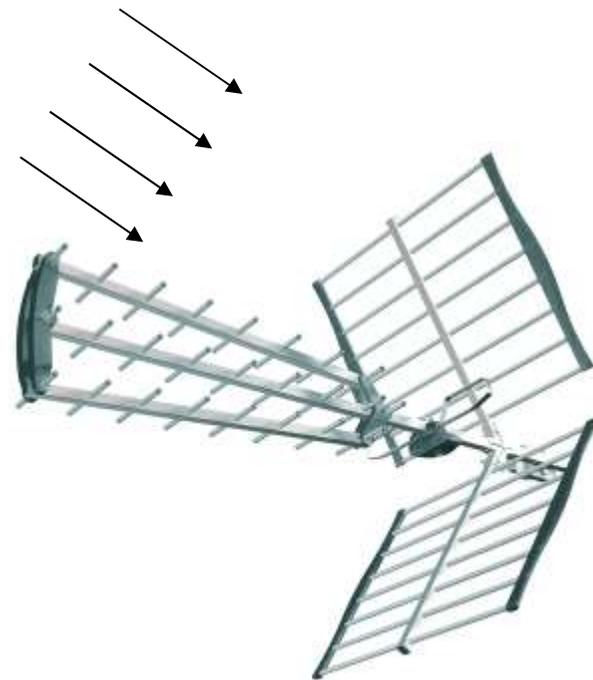
# Parameters of the Rx Antenna

- Rx effective length
- Equivalent circuit of the rx antenna
- Effective Area



# Parameters of the Rx Antenna

- Rx effective length
- Equivalent circuit of the rx antenna
- Effective Area



# Rx effective length

**Effective length**  $\mathbf{l}(\vartheta, \varphi) = l_g(\vartheta, \varphi) \hat{i}_g + l_\varphi(\vartheta, \varphi) \hat{i}_\varphi$

**Fraunhofer region**

$$\mathbf{E}(r, \vartheta, \varphi) = \frac{j\zeta I}{2\lambda} \frac{e^{-j\beta r}}{r} \mathbf{l}(\vartheta, \varphi)$$

**Elementary electrical dipole**  $\mathbf{l}(\vartheta, \varphi) = \Delta z \sin \vartheta \hat{i}_g$

**Small loop antenna**  $\mathbf{l}(\vartheta, \varphi) = -j\beta \Delta S \sin \vartheta \hat{i}_g$

It can be shown that **for an elementary electrical dipole or for a small loop antenna**, the following property is valid:

$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

$\mathbf{l}$  is the tx antenna effective length

$\mathbf{E}_i$  is the incident, locally plane, field

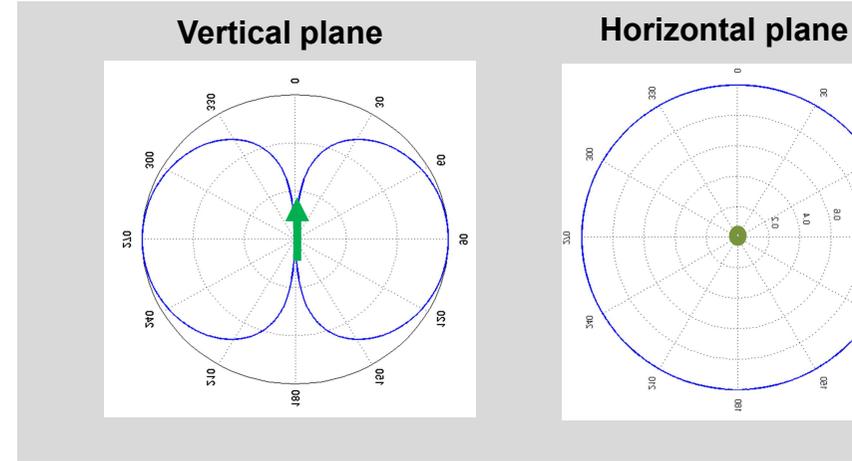
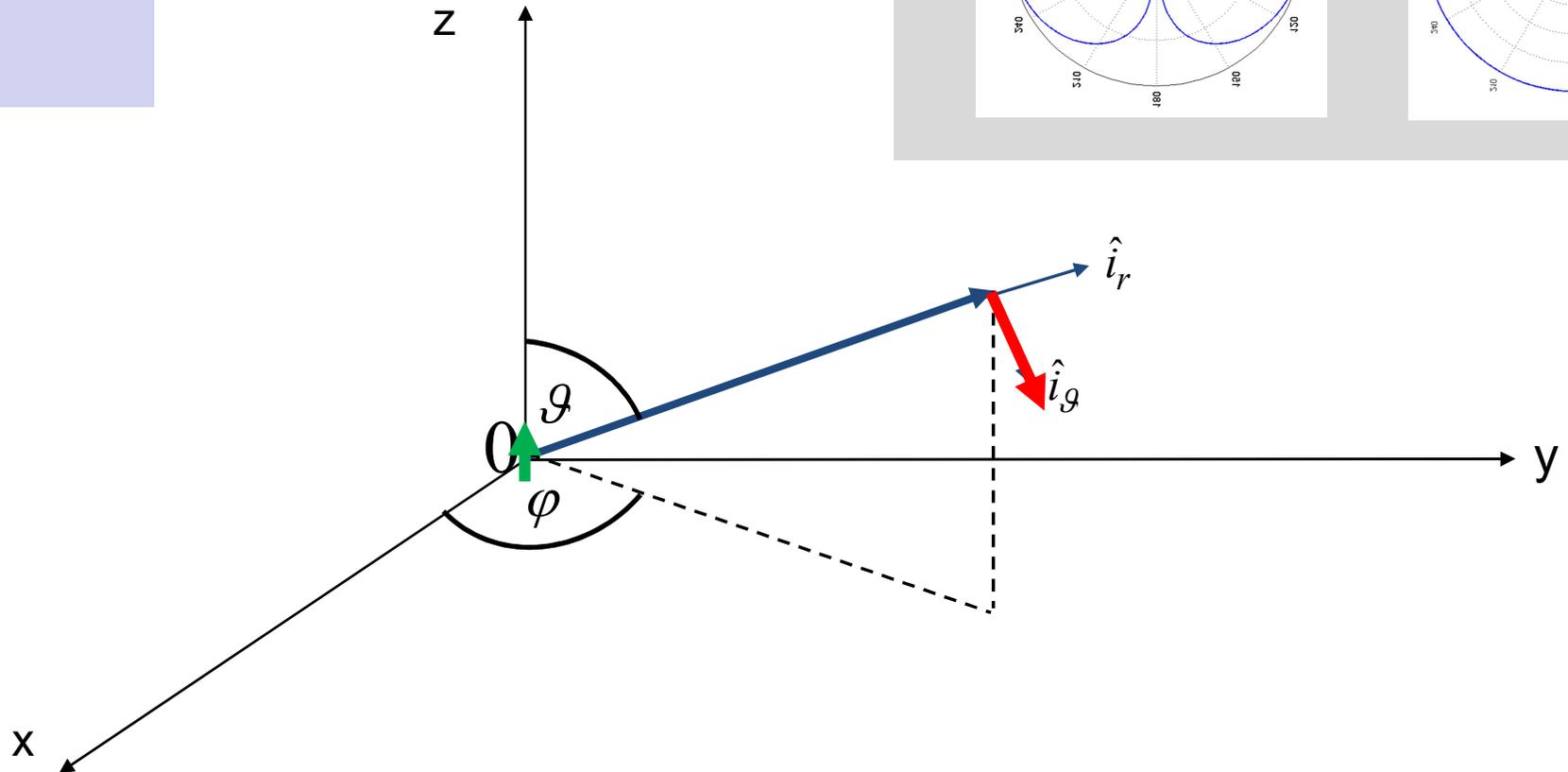
$V_0$  is the voltage induced at the antenna terminals, which are assumed open-circuited

# Rx effective length

Elementary electrical dipole

$$\mathbf{E}_i(\mathcal{R}, \varphi) = \Delta z \sin \mathcal{G} \hat{i}_{\mathcal{G}}$$

$$|E_{\text{eff}}| = |\mathbf{E}_i \cdot \mathbf{l}|$$



# Rx effective length

Elementary electrical dipole

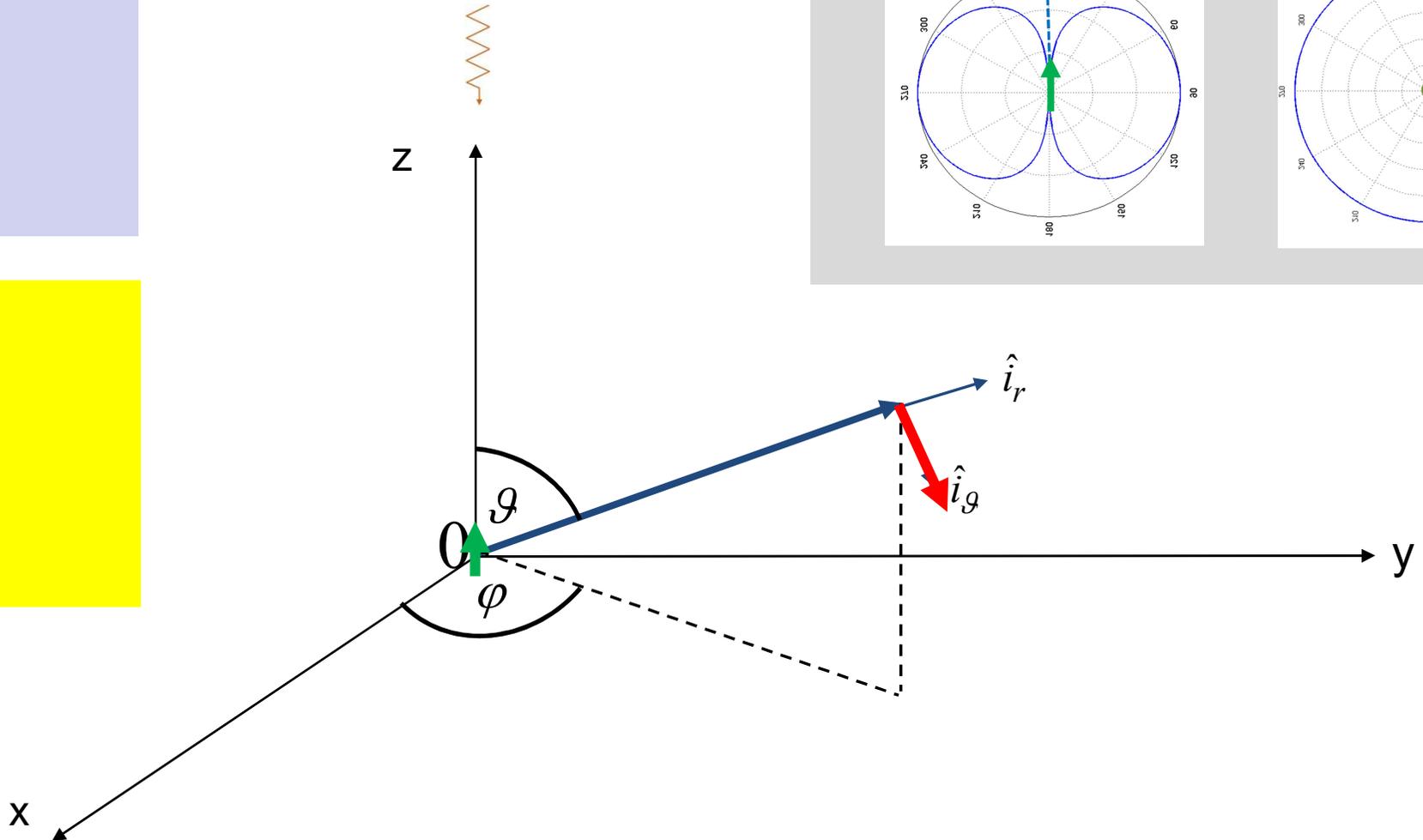
$$\mathbf{E}_i(\theta, \varphi) = \Delta z \sin \theta \hat{i}_\theta$$

$$|E_{eff}| = |\mathbf{E}_i \cdot \mathbf{l}|$$

First example

$$= 0$$

$$= 0$$



# Rx effective length

Elementary electrical dipole

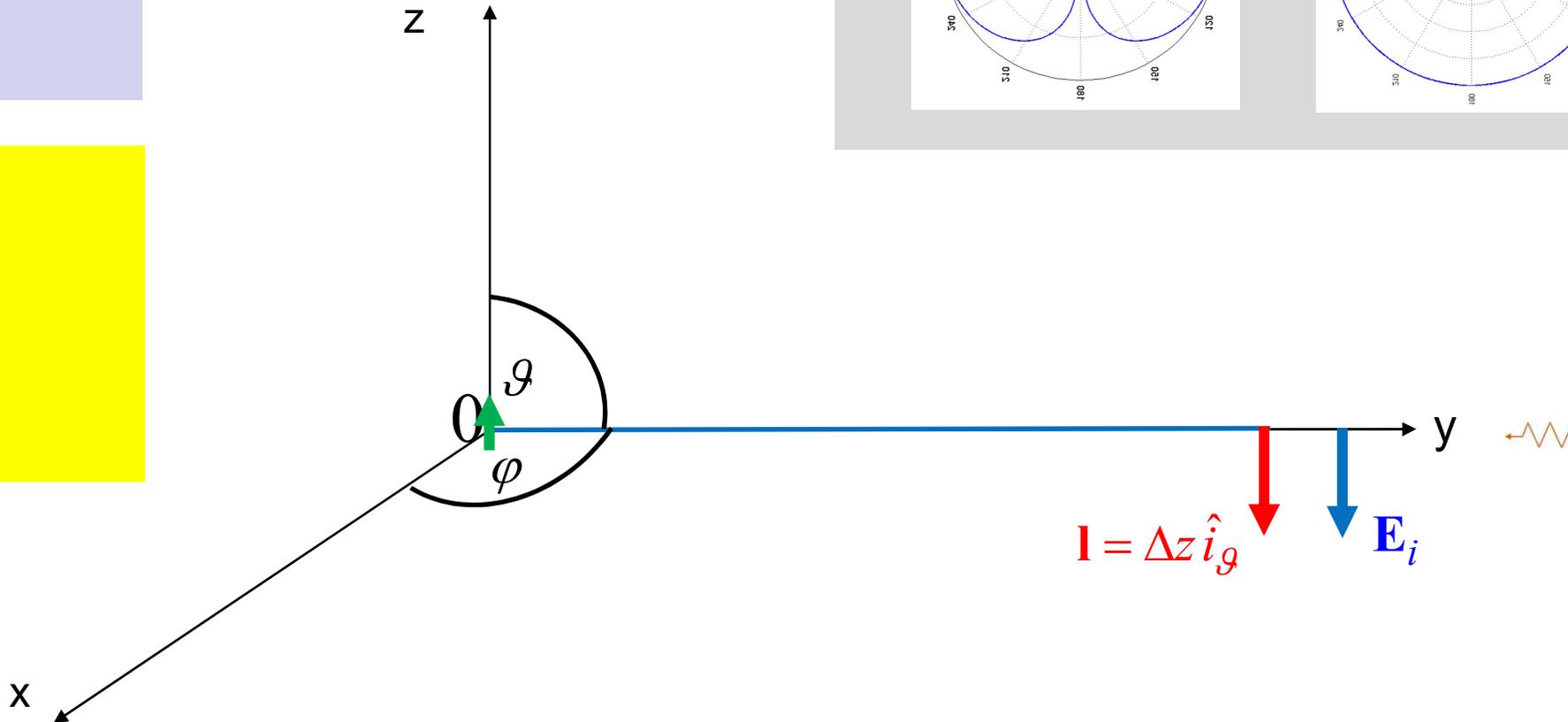
$$r(\theta, \varphi) = \Delta z \sin \theta \hat{i}_\theta$$

$$|r| = |\mathbf{E}_i \cdot \mathbf{l}|$$

Second example

$$\theta = \frac{\pi}{2}; \quad \varphi = \frac{\pi}{2}$$

$$|r| = |\Delta z \mathbf{E}_i|$$



# Rx effective length

Elementary electrical dipole

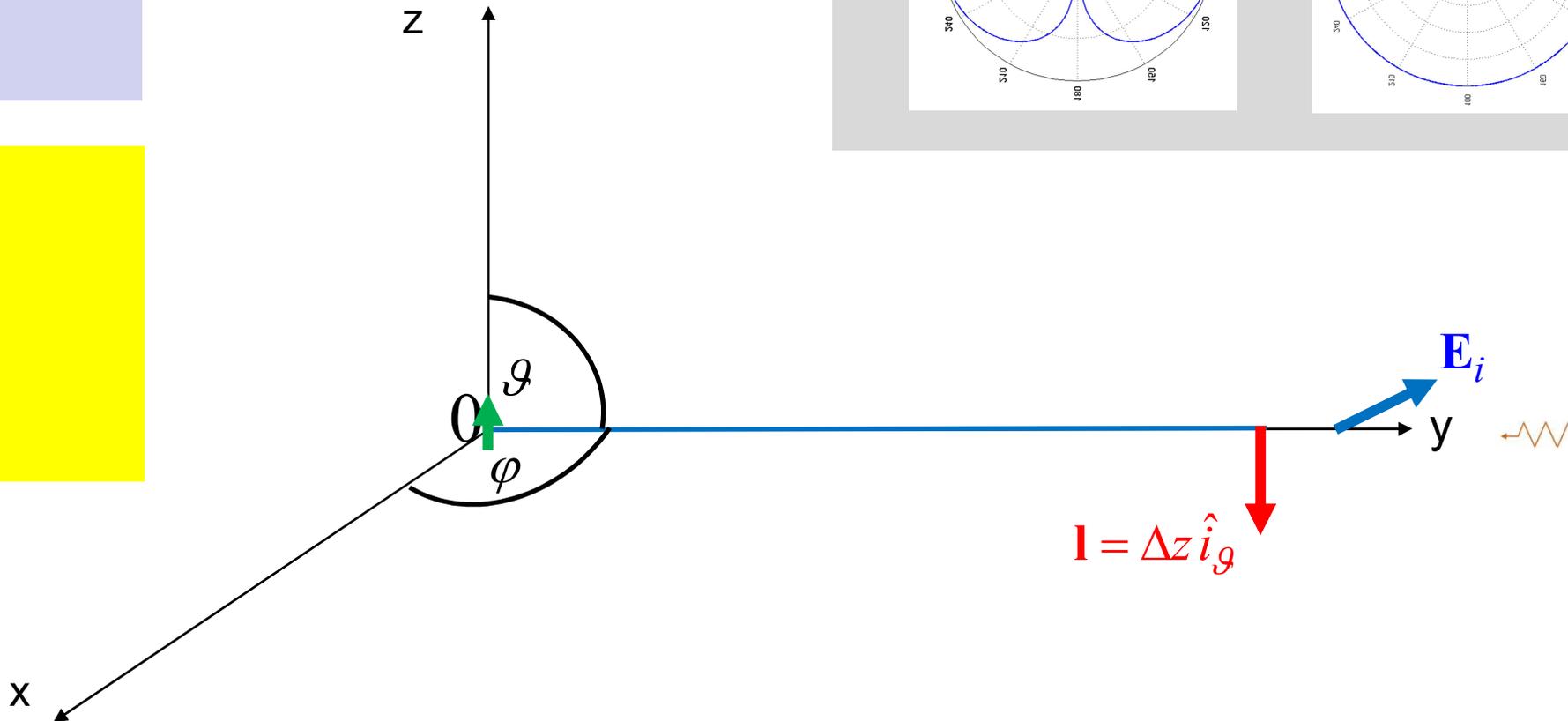
$$G(\theta, \varphi) = \Delta z \sin \theta \hat{i}_\theta$$

$$|G| = |\mathbf{E}_i \cdot \mathbf{l}|$$

Second example

$$\theta = \frac{\pi}{2}; \quad \varphi = \frac{\pi}{2}$$

$$|G| = 0$$



# Rx effective length

Elementary electrical dipole

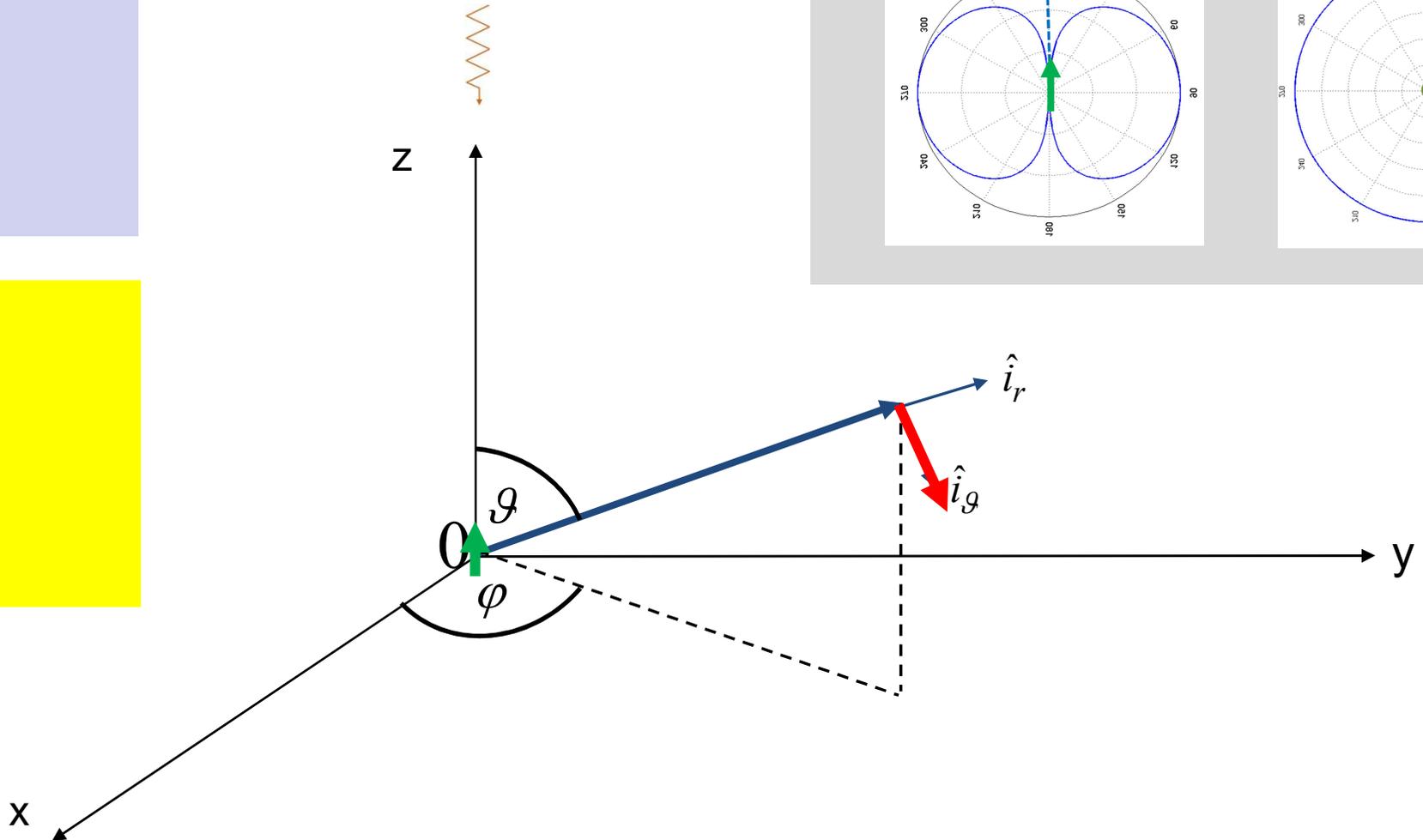
$$\mathbf{E}_i(\theta, \varphi) = \Delta z \sin \theta \hat{i}_\theta$$

$$|E_{eff}| = |\mathbf{E}_i \cdot \mathbf{l}|$$

First example

$$= 0$$

$$= 0$$



# Rx effective length

Interestingly, this result can be extended to **ALL** the antennas by applying the **RECIPROcity THEOR**

It can be shown that **for an elementary electrical dipole or for a small loop antenna**, the following property is valid:

$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

$\mathbf{l}$  is the tx antenna effective length

$\mathbf{E}_i$  is the incident, locally plane, field

$V_0$  is the voltage induced at the antenna terminals, which are assumed open-circuited

# Rx effective length

$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

Where

$\mathbf{E}_i$  is the incident, locally plane, field

$V_0$  is the voltage induced at the antenna terminals, which are assumed open-circuited

$\mathbf{l}(\vartheta, \varphi) = l_\vartheta(\vartheta, \varphi)\hat{i}_\vartheta + l_\varphi(\vartheta, \varphi)\hat{i}_\varphi$  can be referred to as receiving effective length of the antenna (and not only transmitting effective length)

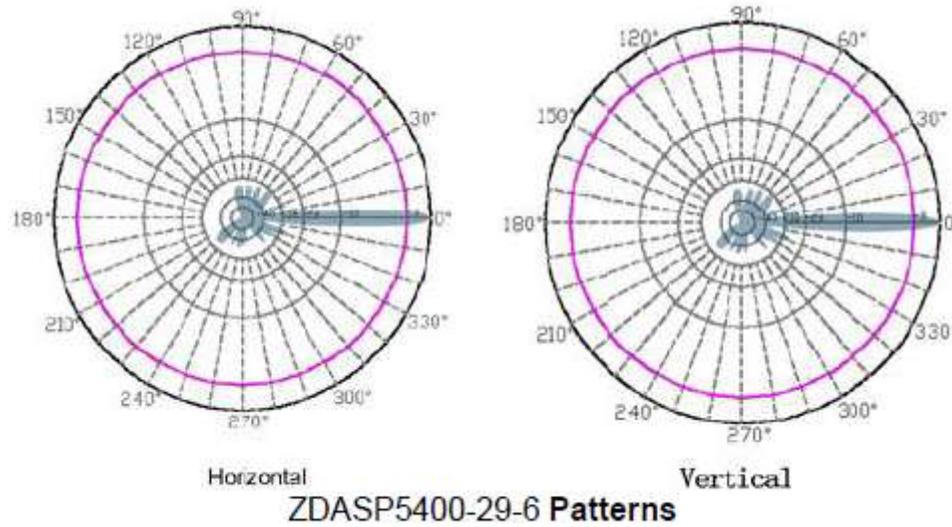
Note that this means that the behavior of an antenna when transmitting and when receiving are related.

# Rx effective length

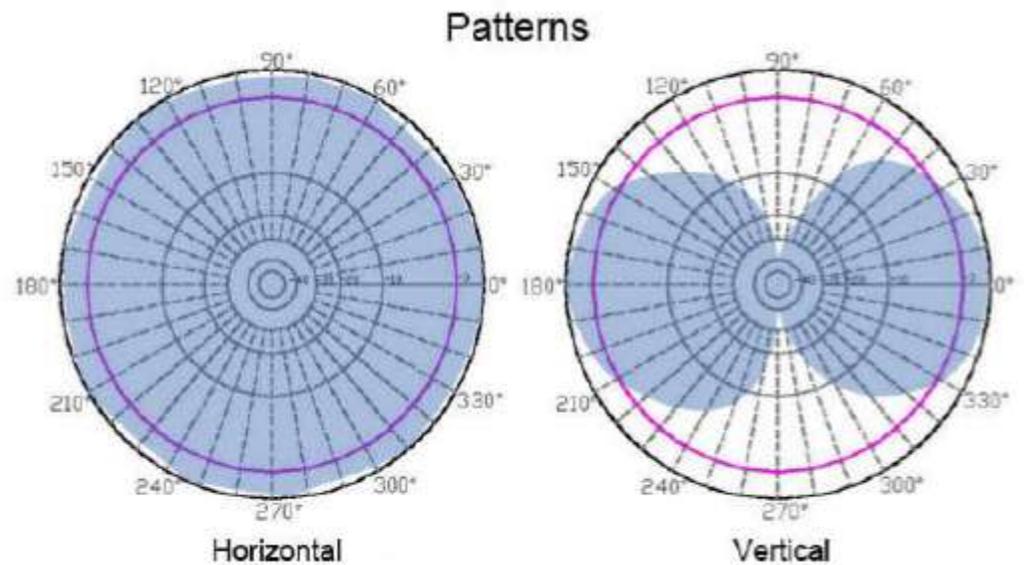
## three examples from the real life



# Rx effective length

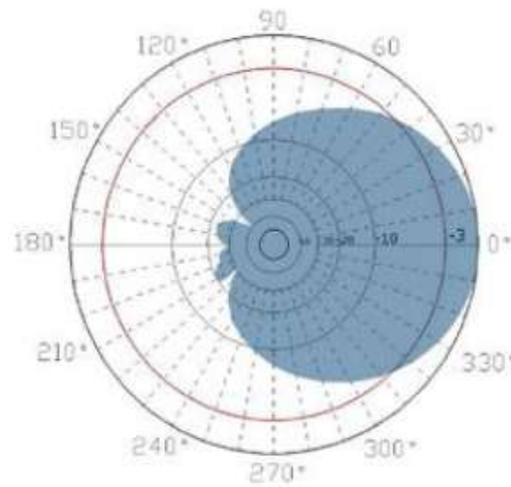


# Rx effective length

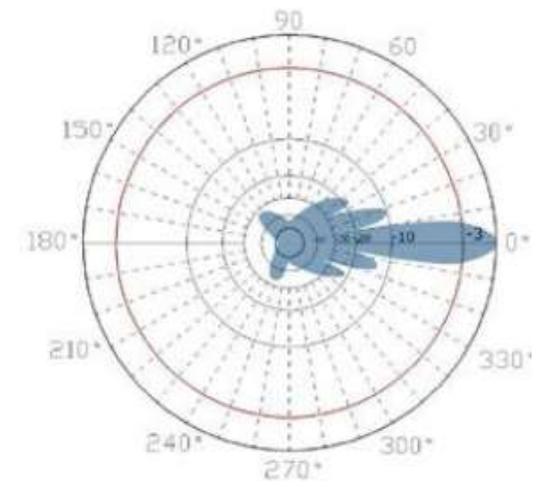


ZDAEW1900-3 Patterns

# Rx effective length



Horizontal

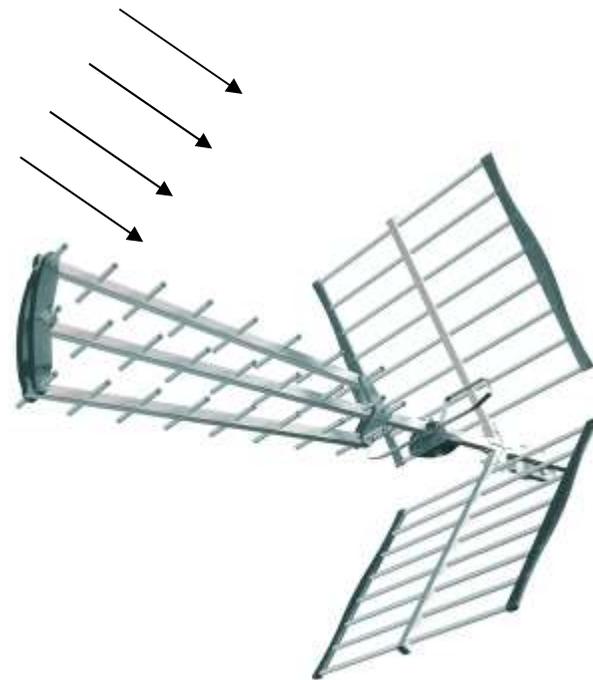


Vertical

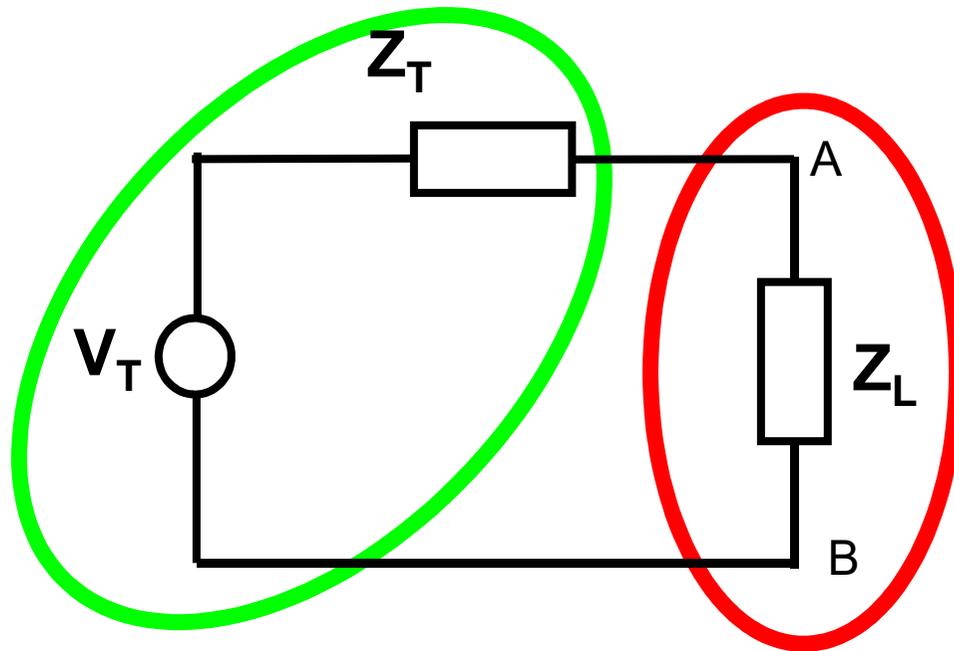
## ZDADJ800-13-90 Patterns

# Parameters of the Rx Antenna

- Rx effective length
- Equivalent circuit of the rx antenna
- Effective Area



# Equivalent circuit of the Rx antenna



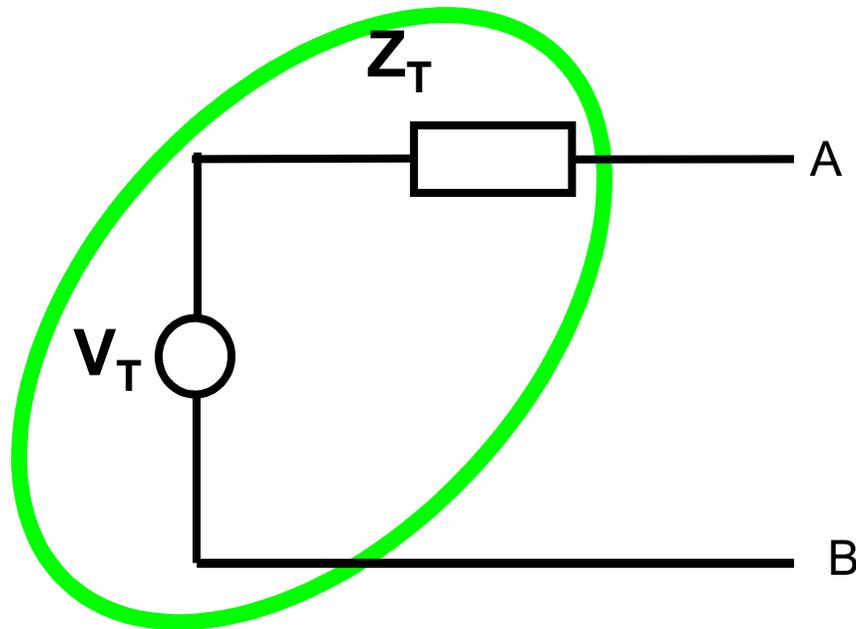
Rx antenna

Load

$$Z_L = R_L + jX_L$$

The incident electric field sets up currents on the antenna. Such currents may be represented by a Thevenin equivalent generator, which delivers power to any connected receiving load impedance.

# Equivalent circuit of the Rx antenna



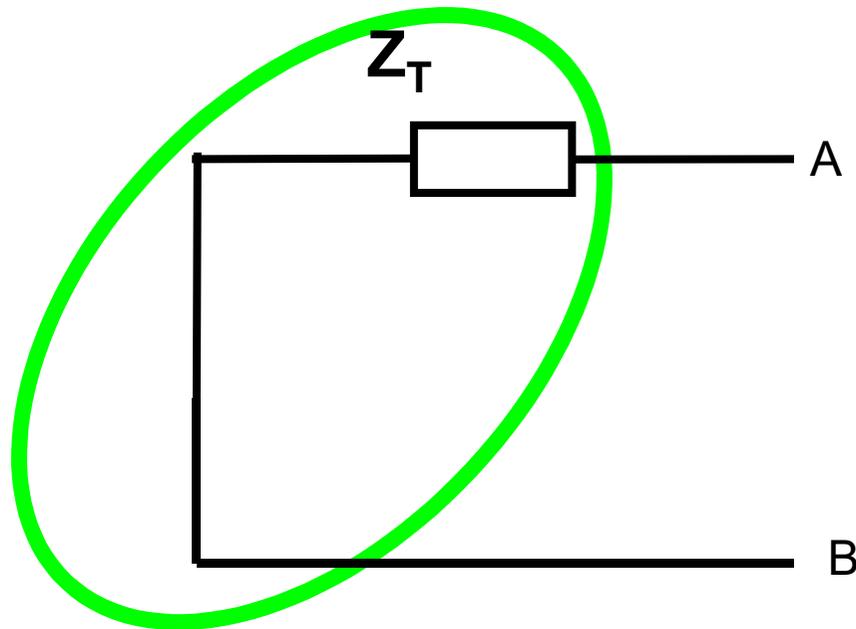
Rx antenna

$$|V_T| = |V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

$$Z_L = R_L + jX_L$$

The incident electric field sets up currents on the antenna. Such currents may be represented by a Thevenin equivalent generator, which delivers power to any connected receiving load impedance.

# Equivalent circuit of the Rx antenna



$$|V_T| = |V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

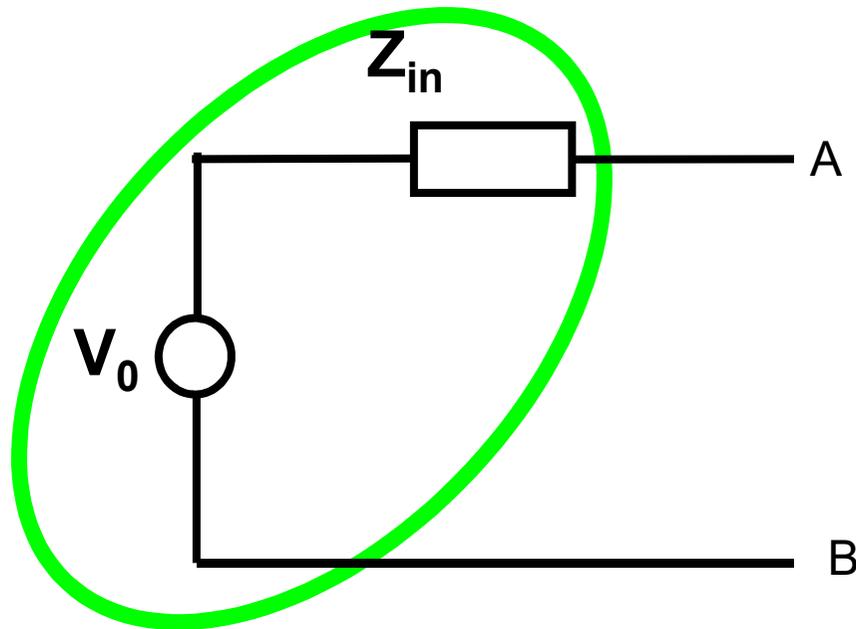
$$Z_T = Z_{in} = R_{in} + jX_{in}$$

$$Z_L = R_L + jX_L$$

Rx antenna

The incident electric field sets up currents on the antenna. Such currents may be represented by a Thevenin equivalent generator, which delivers power to any connected receiving load impedance.

# Equivalent circuit of the Rx antenna



Rx antenna

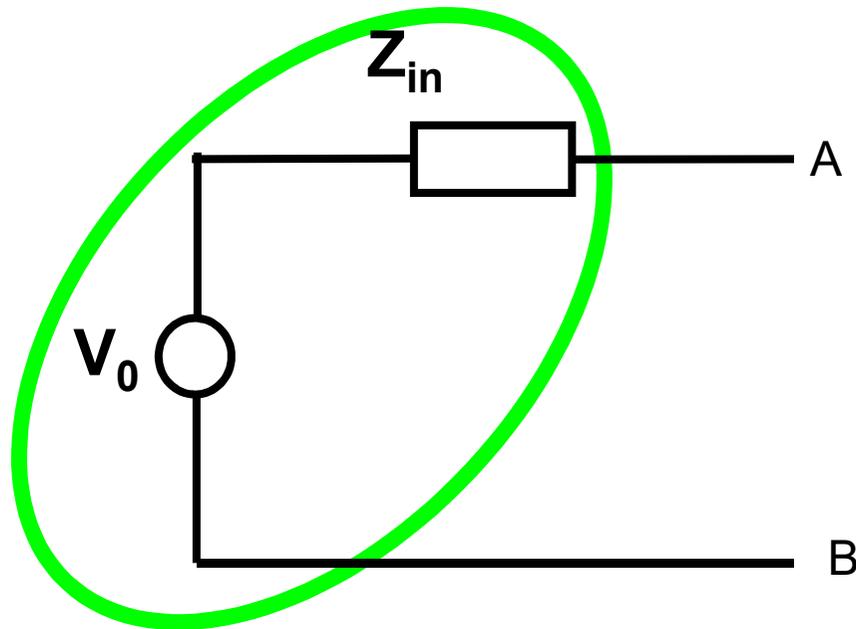
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# Equivalent circuit of the Rx antenna



Rx antenna

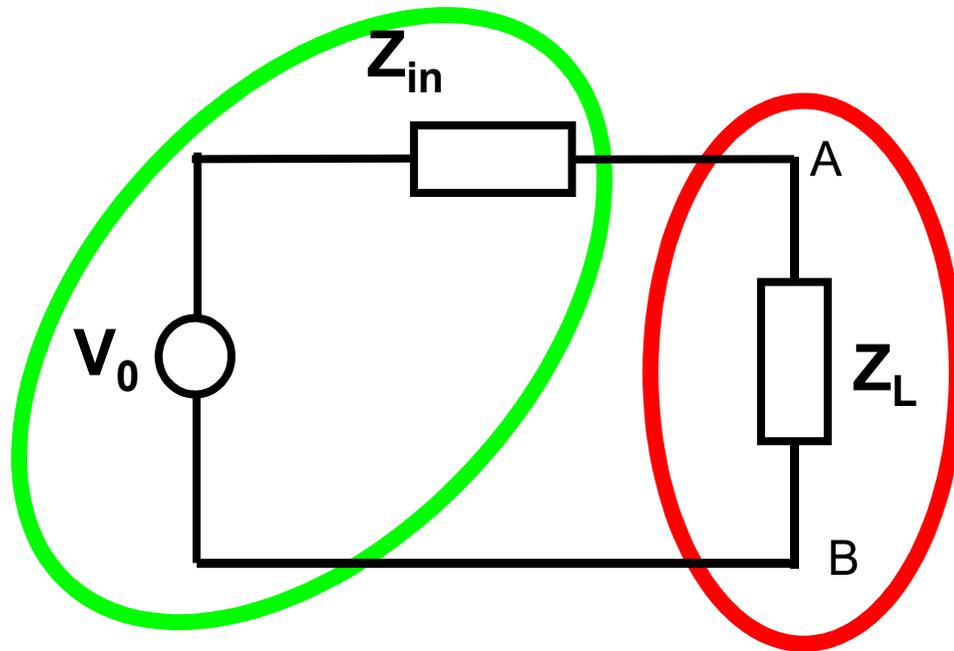
$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

$$Z_{in} = R_{in} + jX_{in}$$

$$Z_L = R_L + jX_L$$

The incident electric field sets up currents on the antenna. Such currents may be represented by a Thevenin equivalent generator, which delivers power to any connected receiving load impedance.

# Equivalent circuit of the Rx antenna



Rx antenna

Load

$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

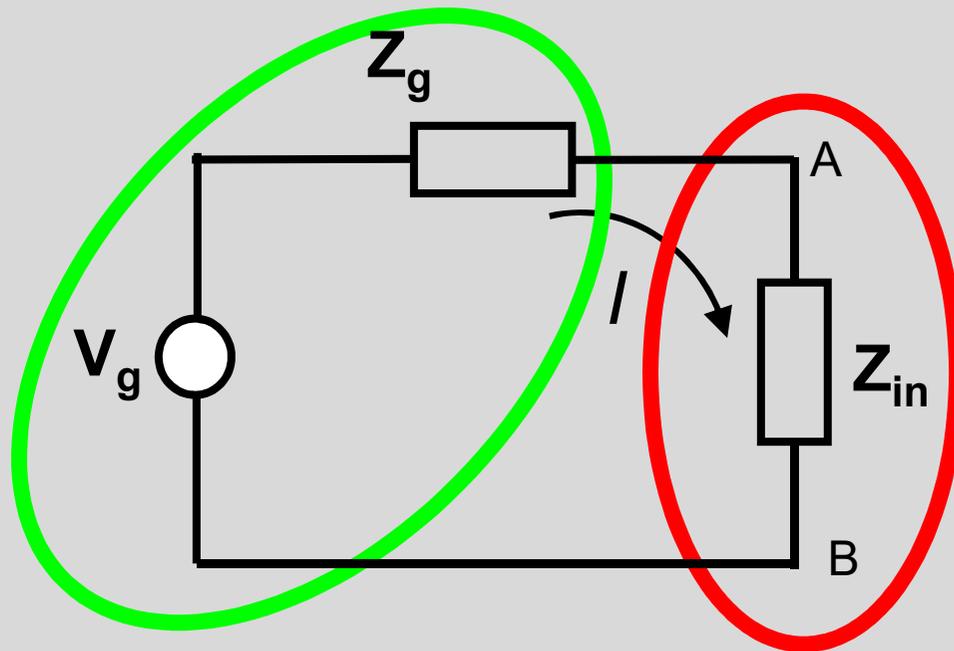
$$Z_{in} = R_{in} + jX_{in}$$

$$Z_L = R_L + jX_L$$

The incident electric field sets up currents on the antenna. Such currents may be represented by a Thevenin equivalent generator, which delivers power to any connected receiving load impedance.

....MEMO..

## Equivalent circuit of the Tx antenna

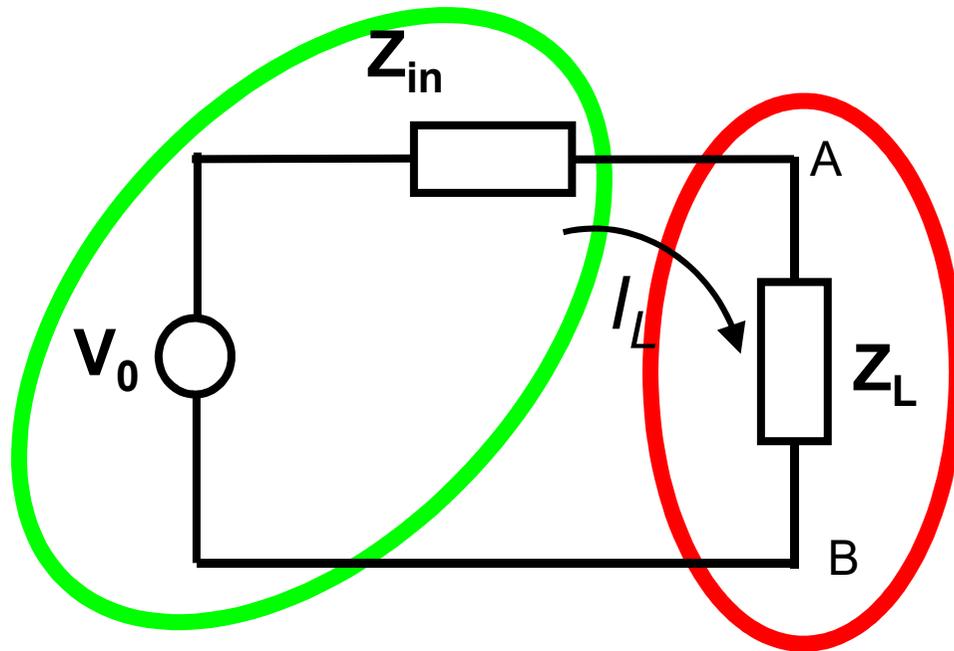


Tx circuit

Tx Antenna

$$Z_{in} = R_{in} + jX_{in}$$
$$P_{in} = \frac{1}{2} R_{in} |I|^2$$

# Equivalent circuit of the Rx antenna



Rx antenna

Load

$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

$$Z_{in} = R_{in} + jX_{in}$$

$$Z_L = R_L + jX_L$$

$$V_0 = (Z_{in} + Z_L)I_L \Rightarrow |I_L|^2 = \frac{|V_0|^2}{|Z_{in} + Z_L|^2}$$

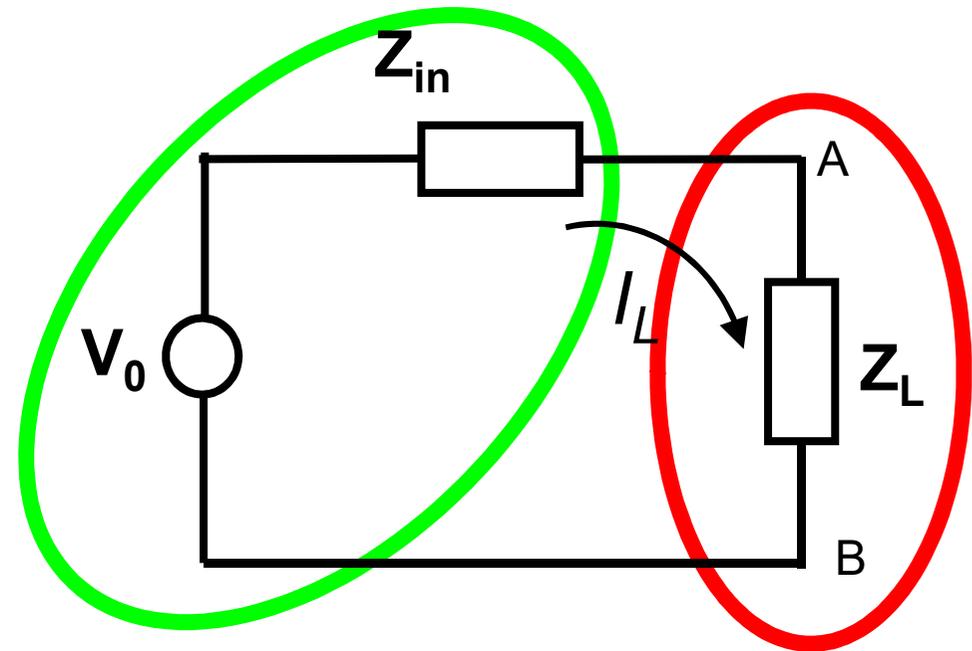
$$P_L = \frac{1}{2} R_L |I_L|^2 = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |V_0|^2 = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

# Equivalent circuit of the Rx antenna

$$|V_0| = |\mathbf{E}_i \cdot \mathbf{l}|$$

$$Z_{in} = R_{in} + jX_{in}$$

$$Z_L = R_L + jX_L$$



Rx antenna

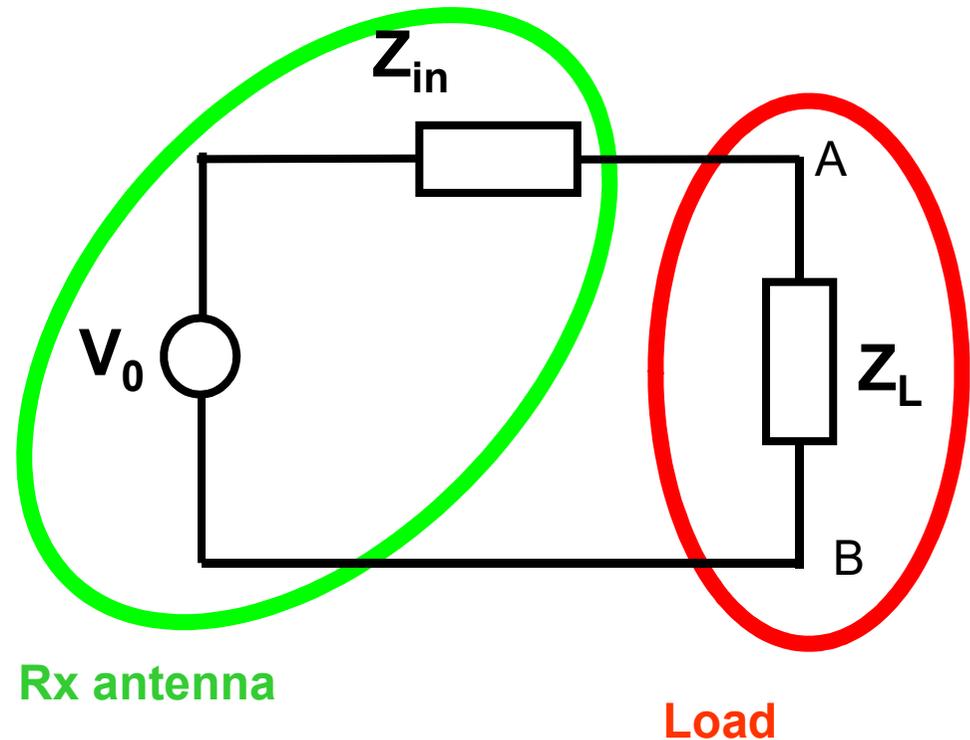
Load

$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

# Equivalent circuit of the Rx antenna

Polarization matching

$$|\mathbf{E}_i \cdot \mathbf{l}| = |\mathbf{E}_i| |\mathbf{l}|$$



$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

# Equivalent circuit of the Rx antenna

Polarization matching

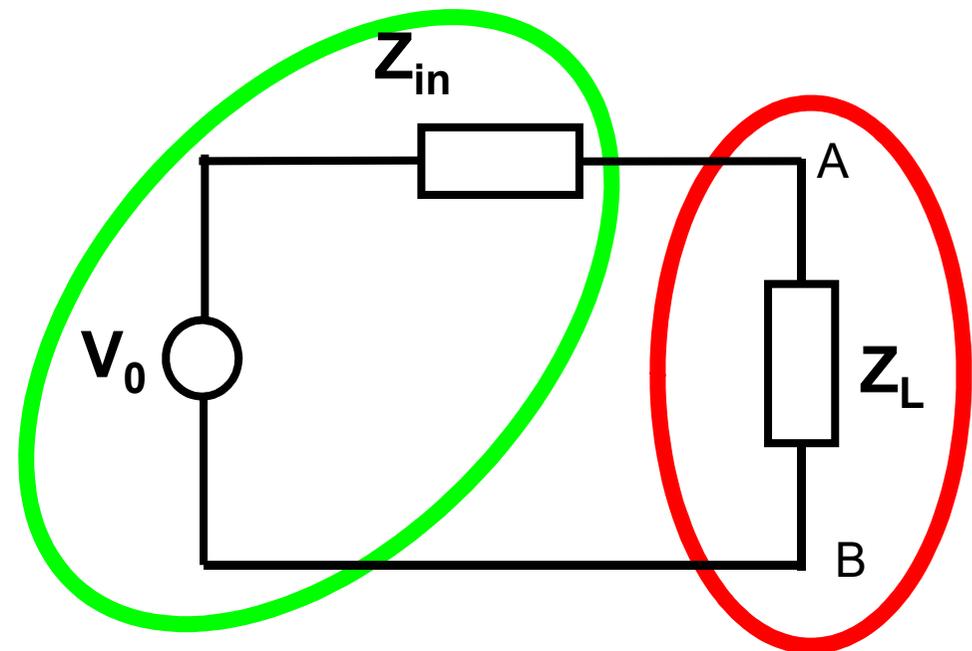
$$|\mathbf{E}_i \cdot \mathbf{l}| = |\mathbf{E}_i| |\mathbf{l}|$$

Power matching

$$Z_L = Z_{in}^* \Rightarrow \begin{cases} Z_{in} = R_{in} + jX_{in} \\ Z_L = R_{in} - jX_{in} \end{cases}$$

$$\rightarrow \frac{R_L}{|Z_{in} + Z_L|^2} = \frac{R_{in}}{(2R_{in})^2} = \frac{1}{4R_{in}}$$

$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$



Rx antenna

Load

# Equivalent circuit of the Rx antenna

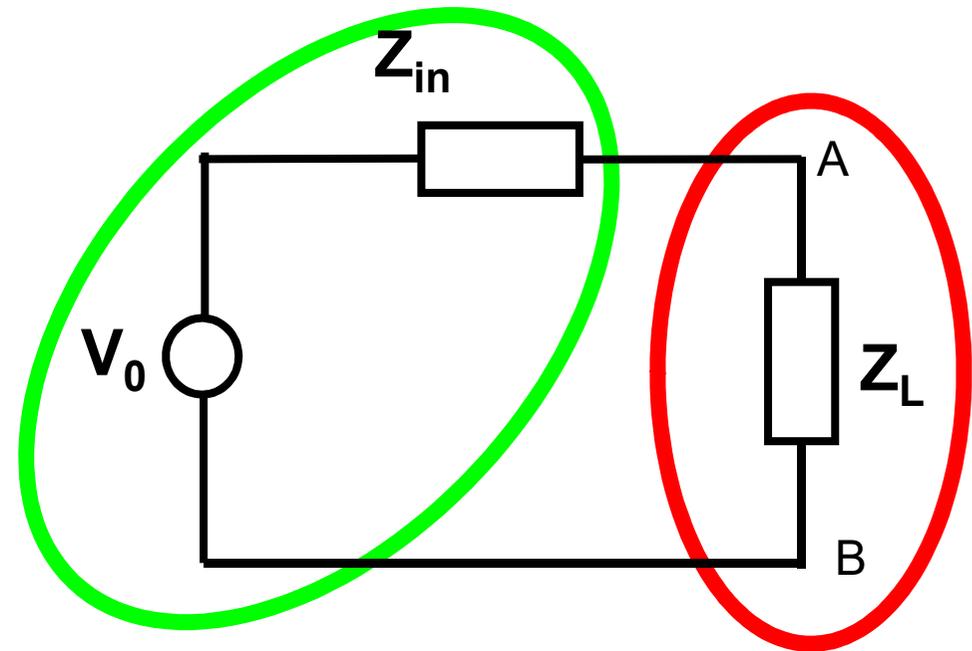
Polarization matching

$$|\mathbf{E}_i \cdot \mathbf{l}| = |\mathbf{E}_i| |\mathbf{l}|$$

Power matching

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$$\frac{R_L}{|Z_{in} + Z_L|^2} = \frac{R_{in}}{(2R_{in})^2} = \frac{1}{4R_{in}}$$



Rx antenna

Load

$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

$$\Rightarrow P_{Lmax} = \frac{1}{2} \frac{1}{4R_{in}} |\mathbf{E}_i|^2 |\mathbf{l}|^2 = \frac{\zeta}{\zeta} \frac{1}{2} \frac{1}{4R_{in}} |\mathbf{E}_i|^2 |\mathbf{l}|^2$$

# Equivalent circuit of the Rx antenna

Polarization matching

$$|\mathbf{E}_i \cdot \mathbf{l}| = |\mathbf{E}_i| |\mathbf{l}|$$

Power matching

$$Z_L = Z_{in}^* \Rightarrow \begin{cases} Z_{in} = R_{in} + jX_{in} \\ Z_L = R_{in} - jX_{in} \end{cases}$$

$$\frac{R_L}{|Z_{in} + Z_L|^2} = \frac{R_{in}}{(2R_{in})^2} = \frac{1}{4R_{in}}$$

$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

$$\Rightarrow P_{Lmax} = \frac{1}{2} \frac{1}{4R_{in}} |\mathbf{E}_i|^2 |\mathbf{l}|^2 = \frac{\zeta}{\zeta} \frac{1}{2} \frac{1}{4R_{in}} |\mathbf{E}_i|^2 |\mathbf{l}|^2$$

$$P_{Lmax} = \frac{1}{2} \frac{|\mathbf{E}_i|^2}{\zeta} \left[ \frac{\zeta |\mathbf{l}(\vartheta, \varphi)|^2}{4R_{in}} \right]$$

Effective Area

$$A_{eff}(\vartheta, \varphi)$$

# Effective area

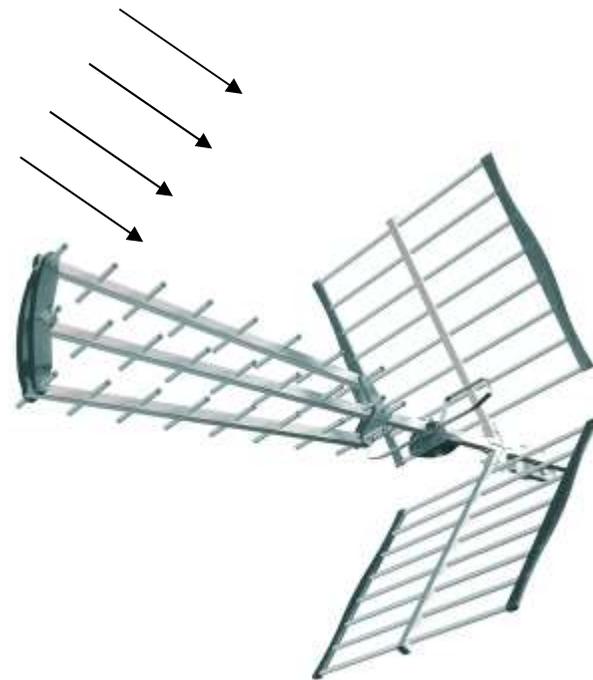
In general, the aperture of an antenna is not directly related to its physical size.

However some types of antennas, for example parabolic dishes and horns, have a physical aperture which collects the radio waves. In these aperture antennas, the physical aperture  $A_{\text{phys}}$  provides a good approximation of the effective area  $A$ .

In any case, the effective aperture  $A$  is always less than the area of the antenna's physical aperture  $A_{\text{phys}}$ .

# Parameters of the Rx Antenna

- Rx effective length
- Equivalent circuit of the rx antenna
- Effective Area



# Gain and Effective Area

Gain

$$G(\vartheta, \varphi) = \frac{\frac{1}{2} \frac{|\mathbf{E}|^2}{\zeta}}{\frac{1}{4\pi r^2} P_{in}}$$

$$P_{in} = \frac{1}{2} R_{in} |I|^2$$

$$\mathbf{E}(r, \vartheta, \varphi) = \frac{j\zeta l e^{-j\beta r}}{2\lambda r} \mathbf{1}(\vartheta, \varphi)$$

Effective Area

$$A_{eff}(\vartheta, \varphi) = \frac{\zeta |\mathbf{1}(\vartheta, \varphi)|^2}{4R_{in}}$$

$$G(\vartheta, \varphi) = \frac{4\pi r^2 \frac{|\mathbf{E}|^2}{\zeta}}{P_{in}} = \frac{4\pi r^2 \left( \frac{\zeta^2 |I|^2}{4\lambda^2} \frac{|\mathbf{1}(\vartheta, \varphi)|^2}{r^2} \right)}{\frac{1}{2} R_{in} |I|^2}$$

# Gain and Effective Area

Gain

$$G(\vartheta, \varphi) = \frac{\frac{1}{2} \frac{|\mathbf{E}|^2}{\zeta}}{\frac{1}{4\pi r^2} P_{in}}$$

$$P_{in} = \frac{1}{2} R_{in} |I|^2$$

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Effective Area

$$A_{eff}(\vartheta, \varphi) = \frac{\zeta |\mathbf{l}(\vartheta, \varphi)|^2}{4R_{in}}$$

$$G(\vartheta, \varphi) = \frac{4\pi r^2}{2\zeta} \frac{|\mathbf{E}|^2}{P_{in}} = \frac{4\pi \cancel{\lambda}^2}{\cancel{2\zeta}} \left( \frac{\cancel{\zeta} \cancel{|I|^2}}{4\lambda^2} \frac{|\mathbf{l}(\vartheta, \varphi)|^2}{\cancel{r^2}} \right) \frac{\cancel{\lambda}}{R_{in} \cancel{\lambda}^2} = \frac{4\pi}{\lambda^2} \left[ \frac{\zeta |\mathbf{l}(\vartheta, \varphi)|^2}{4R_{in}} \right]$$

# Gain and Effective Area

Gain

$$G(\vartheta, \varphi) = \frac{\frac{1}{2} \frac{|\mathbf{E}|^2}{\zeta}}{\frac{1}{4\pi r^2} P_{in}}$$

$$P_{in} = \frac{1}{2} R_{in} |I|^2$$

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Effective Area

$$A_{eff}(\vartheta, \varphi) = \frac{\zeta |\mathbf{1}(\vartheta, \varphi)|^2}{4R_{in}}$$

$$G(\vartheta, \varphi) = \frac{4\pi r^2}{2\zeta} \frac{|\mathbf{E}|^2}{P_{in}} = \frac{4\pi r^2}{2\zeta} \left( \frac{\zeta^2 |I|^2}{4\lambda^2} \frac{|\mathbf{1}(\vartheta, \varphi)|^2}{r^2} \right) \frac{2}{R_{in} |I|^2} = \frac{4\pi}{\lambda^2} \left[ \frac{\zeta |\mathbf{1}(\vartheta, \varphi)|^2}{4R_{in}} \right]$$

$$G(\vartheta, \varphi) = \frac{4\pi}{\lambda^2} A_{eff}(\vartheta, \varphi)$$

# Effective Area

three examples from the real life



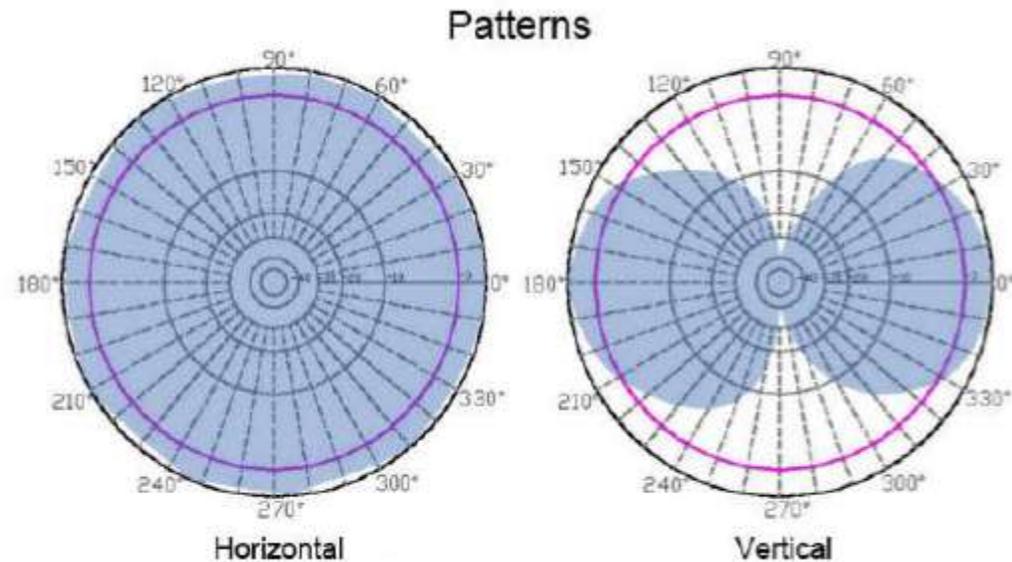
# Effective Area

three examples from the real life



(maximum) Gain = 3 dB

$$G(\vartheta, \varphi) = \frac{4\pi}{\lambda^2} A_{eff}(\vartheta, \varphi)$$



ZDAEW1900-3 Patterns

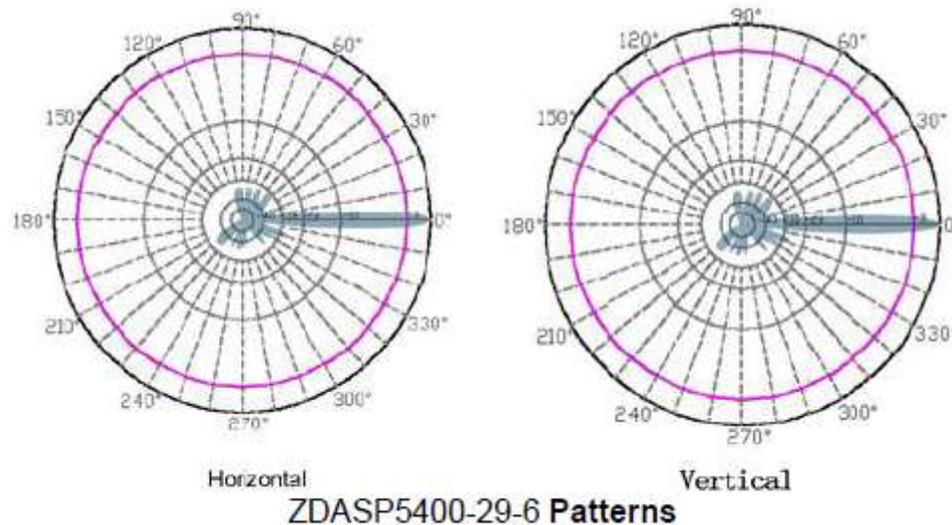
# Effective Area

three examples from the real life



(maximum) Gain = 29 dB

$$G(\vartheta, \varphi) = \frac{4\pi}{\lambda^2} A_{eff}(\vartheta, \varphi)$$



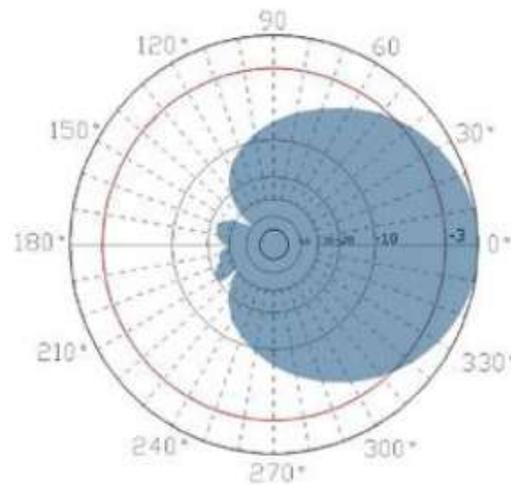
# Effective Area

three examples from the real life

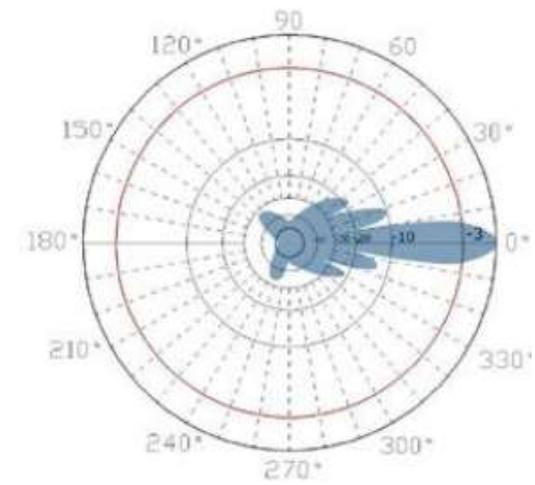


(maximum) Gain = 13 dB

$$G(\vartheta, \varphi) = \frac{4\pi}{\lambda^2} A_{eff}(\vartheta, \varphi)$$



Horizontal



Vertical

ZDADJ800-13-90 Patterns

# Equivalent circuit of the Rx antenna

$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

$$P_{Lmax} = \frac{1}{2} \frac{|\mathbf{E}_i|^2 |\mathbf{l}|^2}{4R_{in}}$$

imum power transfer to the load is achieved when the two following conditions are simultaneously verified

## 1) Polarization matching

$$\Rightarrow |\mathbf{E}_i \cdot \mathbf{l}| = |\mathbf{E}_i| |\mathbf{l}|$$

## 2) Power matching

$$\Rightarrow \frac{R_L}{|Z_{in} + Z_L|^2} = \frac{1}{4R_{in}}$$

$$P_{Lmax} = \frac{1}{2} \frac{|\mathbf{E}_i|^2}{\zeta} A_{eff}(\vartheta, \varphi)$$

# Equivalent circuit of the Rx antenna

$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2$$

$$P_{Lmax} = \frac{1}{2} \frac{|\mathbf{E}_i|^2 |\mathbf{l}|^2}{4R_{in}}$$

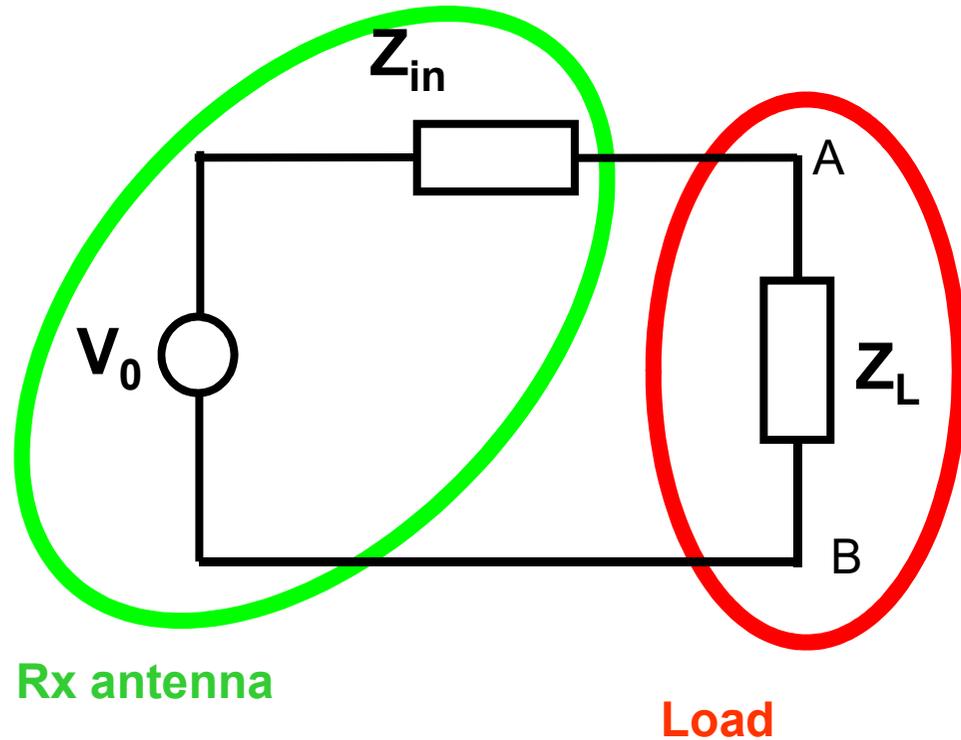
$$P_L = \frac{1}{2} \frac{R_L}{|Z_{in} + Z_L|^2} |\mathbf{E}_i \cdot \mathbf{l}|^2 = \underbrace{\frac{1}{2} \frac{|\mathbf{E}_i|^2 |\mathbf{l}|^2}{4R_{in}}}_{P_{Lmax}} \underbrace{\frac{|\mathbf{E}_i \cdot \mathbf{l}|^2}{|\mathbf{E}_i|^2 |\mathbf{l}|^2}}_{\eta_A} \underbrace{\frac{4R_{in} R_L}{|Z_{in} + Z_L|^2}}_{\eta_B}$$

$$P_L = \eta_A \eta_B P_{Lmax}$$

# Equivalent circuit of the Rx antenna

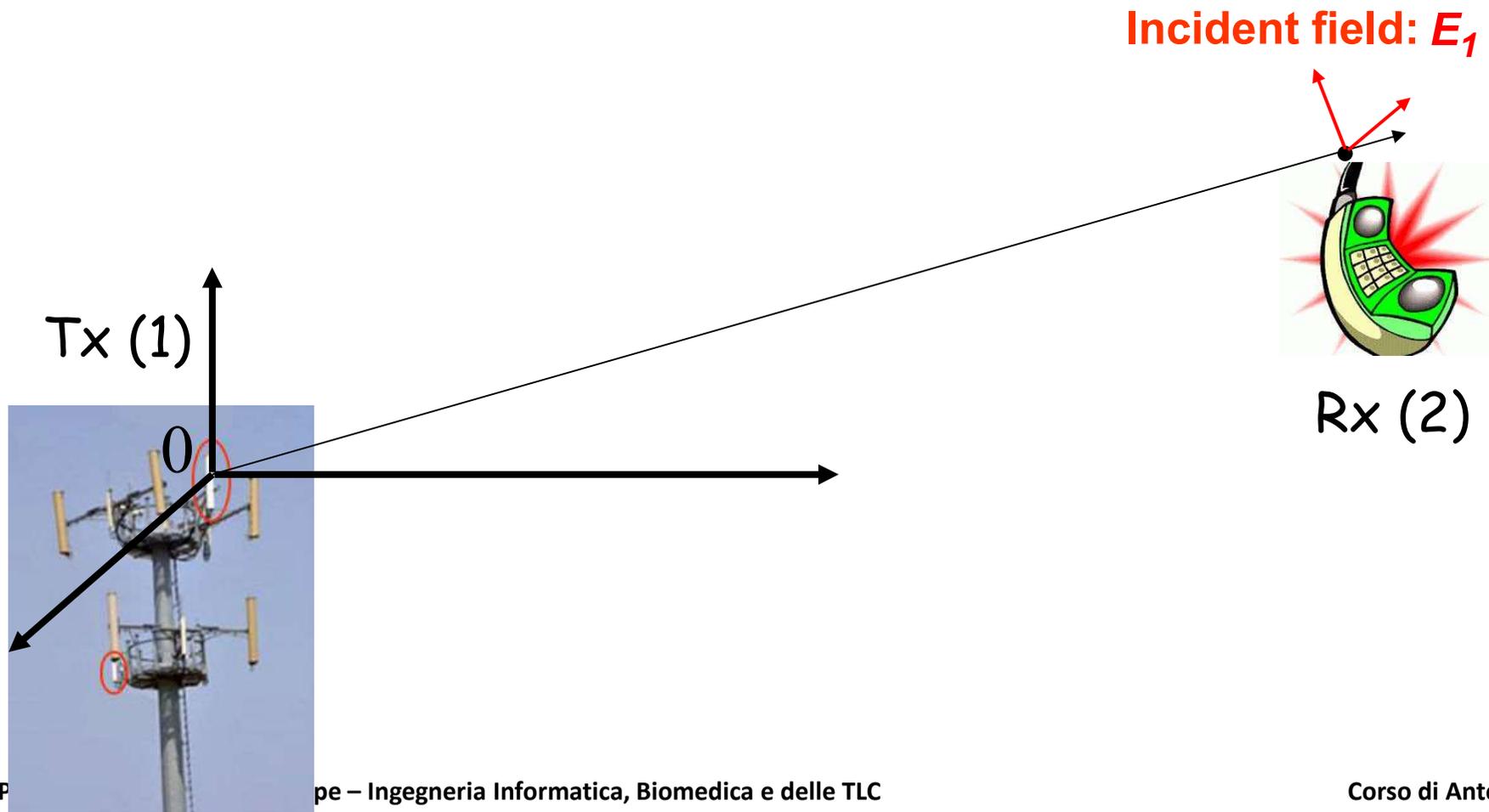
$$P_L = \eta_A \eta_B P_{Lmax}$$

$$P_{Lmax} = \frac{1}{2} \frac{|\mathbf{E}_i|^2}{\zeta} A_{eff}(\vartheta, \varphi)$$



$$P_L = \frac{1}{2} \frac{|\mathbf{E}_i|^2}{\zeta} A_{eff}(\vartheta, \varphi) \eta_A \eta_B$$

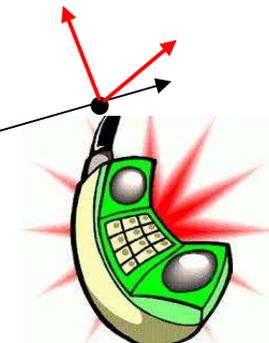
# Radio link equation



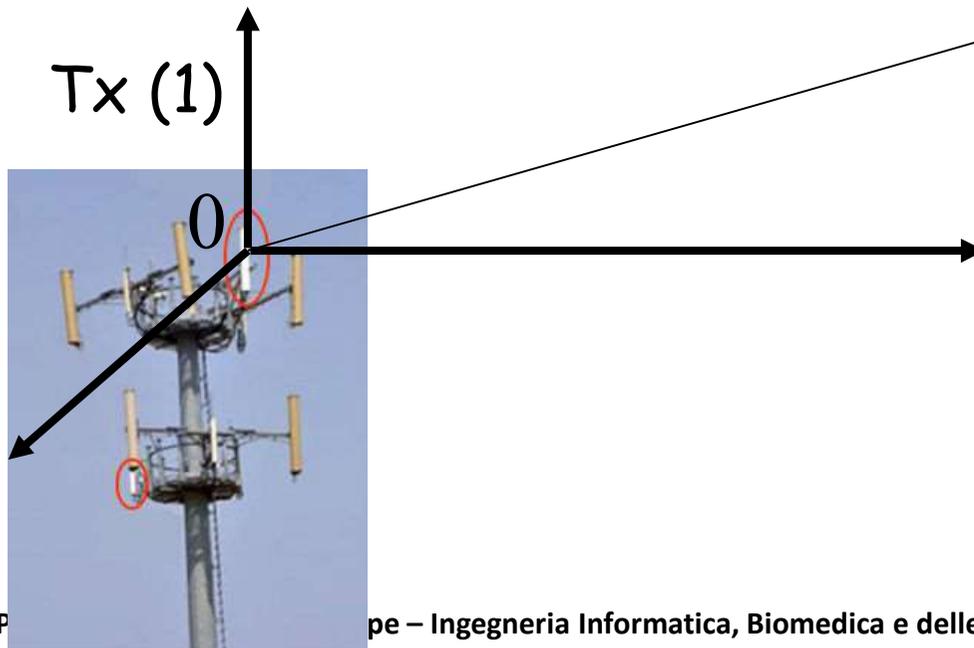
# Radio link equation

$$P_2 = \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} A_{eff} \eta_A \eta_B$$

Incident field:  $E_1$



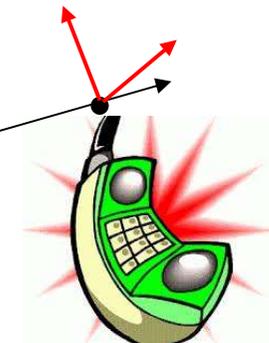
Rx (2)



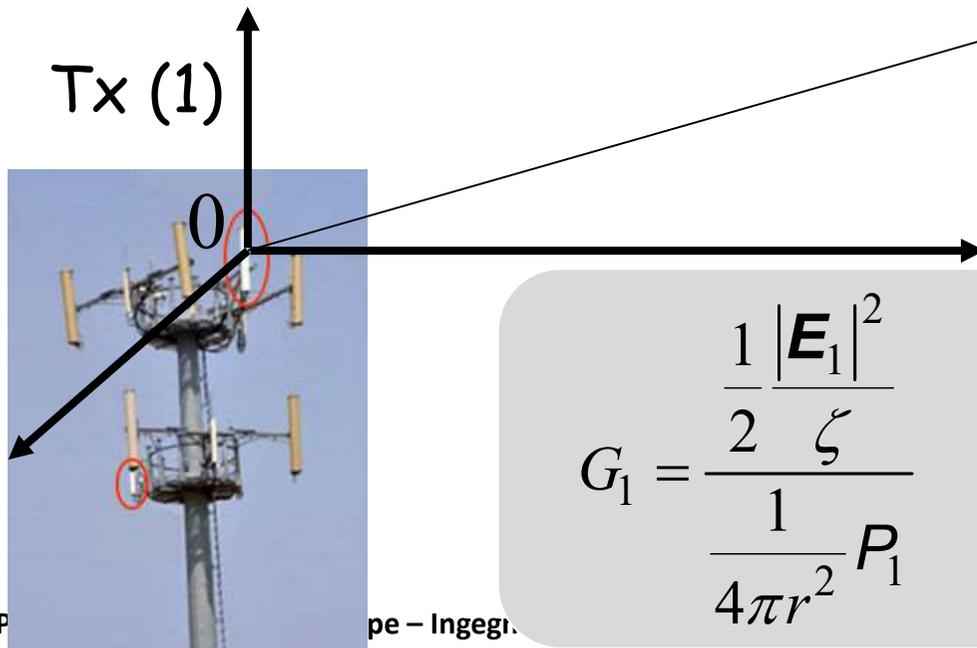
# Radio link equation

$$P_2 = \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} A_{eff} \eta_A \eta_B$$

Incident field:  $\mathbf{E}_1$



Rx (2)

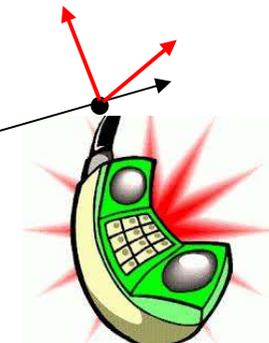


$$G_1 = \frac{\frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta}}{\frac{1}{4\pi r^2} P_1}$$

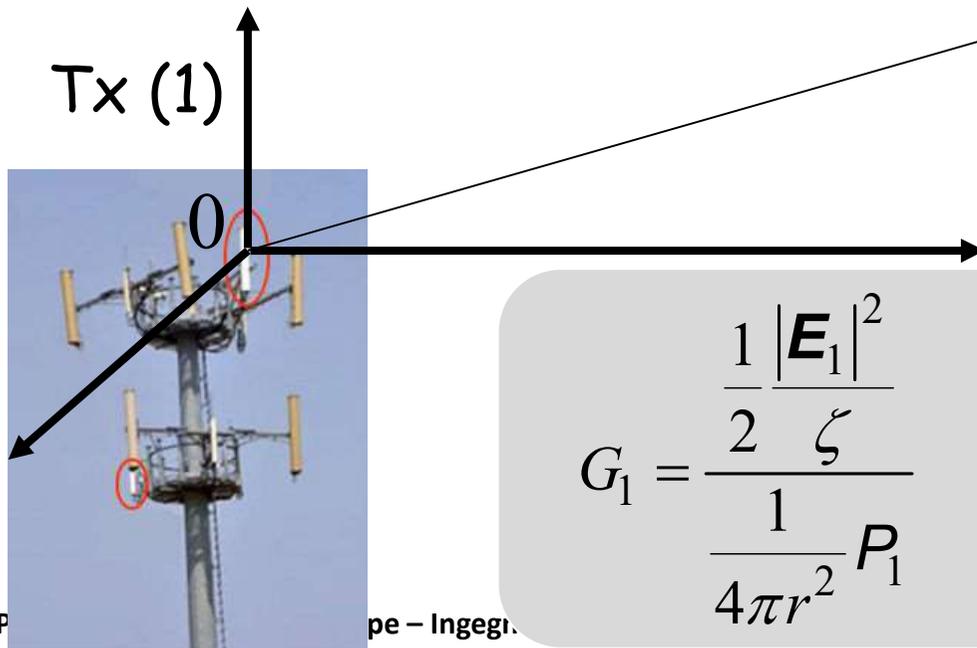
# Radio link equation

$$P_2 = \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} A_{eff} \eta_A \eta_B$$

Incident field:  $\mathbf{E}_1$



Rx (2)



$$G_1 = \frac{\frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta}}{\frac{1}{4\pi r^2} P_1} \quad \longrightarrow \quad \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} = \frac{P_1 G_1}{4\pi r^2}$$

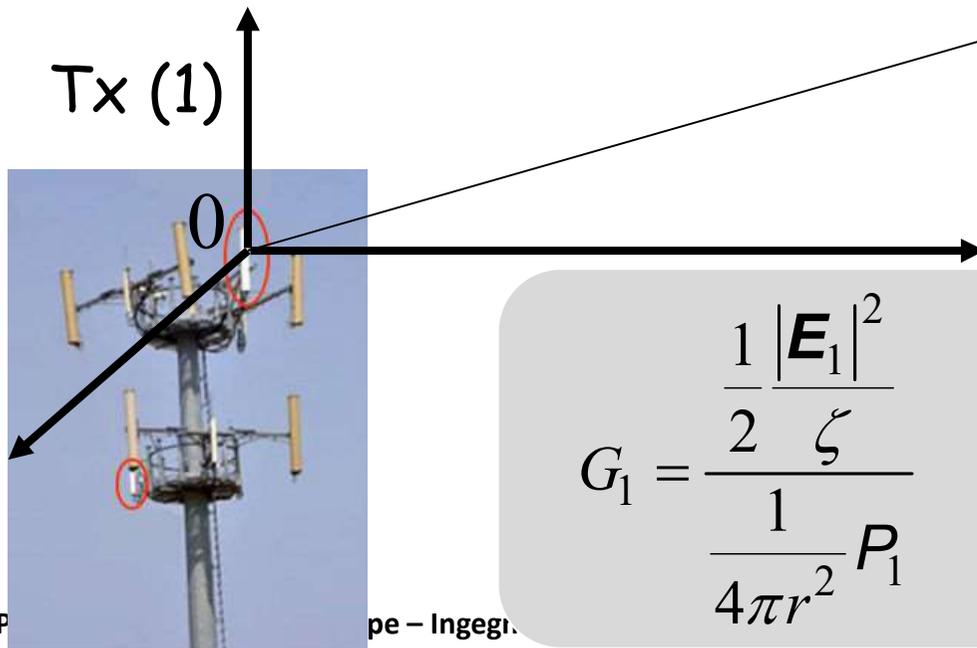
# Radio link equation

$$P_2 = \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} A_{eff2} \eta_A \eta_B = P_1 G_1 A_{eff2} \left( \frac{1}{4\pi r^2} \right) \eta_A \eta_B$$

Incident field:  $\mathbf{E}_1$



Rx (2)



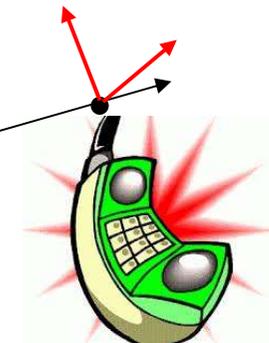
$$G_1 = \frac{\frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta}}{\frac{1}{4\pi r^2} P_1} \quad \rightarrow \quad \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} = \frac{P_1 G_1}{4\pi r^2}$$

# Radio link equation

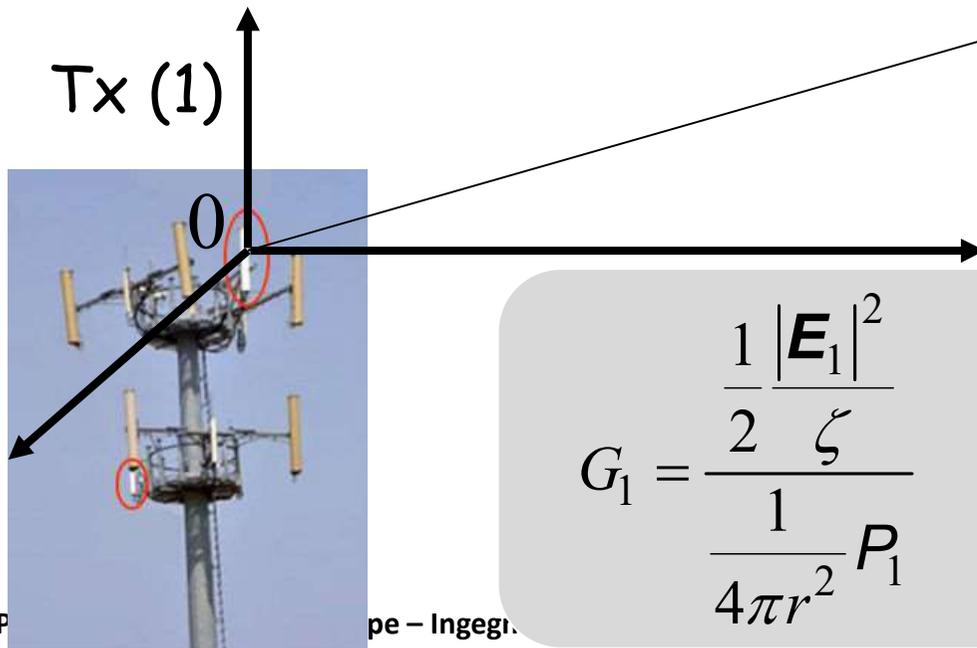
$$P_2 = \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} A_{eff2} \eta_A \eta_B = P_1 G_1 A_{eff2} \left( \frac{1}{4\pi r^2} \right) \eta_A \eta_B$$

$$P_2 = P_1 G_1 A_{eff2} \left( \frac{1}{4\pi r^2} \right) \eta_A \eta_B$$

Incident field:  $\mathbf{E}_1$



Rx (2)

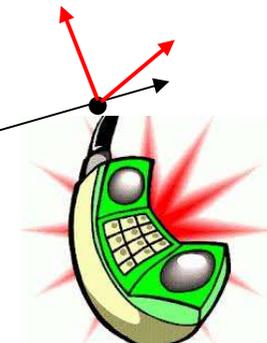


$$G_1 = \frac{\frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta}}{\frac{1}{4\pi r^2} P_1} \rightarrow \frac{1}{2} \frac{|\mathbf{E}_1|^2}{\zeta} = \frac{P_1 G_1}{4\pi r^2}$$

# Radio link equation (Friis equation)

$$P_2 = P_1 G_1 A_{eff2} \left( \frac{1}{4\pi r^2} \right) \eta_A \eta_B$$

Incident field:  $E_1$



Rx (2)

Tx (1)

0

