

Course of "Automatic Control Systems" 2022/23

Control requirements: Transient performance

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▲ A SISO closed loop control system in the Laplace domain can be indicated as



- G(s) plant to be controlled
- K(s) controller
- **R(s) reference**
- Y(s) controlled output
- U(s) control variable
- E(s) tracking error
- **D(s) disturb**
- N(s) measurement noise

Closed loop function

 $T(s) = W(s) = \frac{G(s)K(s)}{1 + G(s)K(s)}$

Different notations T or W



- \checkmark The closed loop control requirements can be divided in four classes:

 - Robust stability (DONE)
 - Steady-state performances (DONE)
 - Transient performances



- A The *transient performance* are usually expressed in terms of tracking properties of the reference signal R(s).
- \land R(s) is usually assumed as a polynomial signal of order 0 (step)
- \checkmark The transient performance can be divided in
 - Dynamic precision (overshoot, oscillation period ...)
 - Time response (rise time, peak time, settling time...)

A The rejection of the disturbs is usually not included among the transient performance because the transfer functions $R(s) \rightarrow Y(s)$ and $D(s) \rightarrow Y(s)$ have the same poles (excluding poles-zeros cancellation).



- ▲ Let us define two quantities in the frequency domain related to the transient behavior
 - \Rightarrow Bandwidth B_3 of the closed loop system related to the settling time
 - * Resonant peak M_p of the closed loop system related to the overshoot

▲ For the typical magnitude plots encountered so far, we define the **bandwidth** B_3 as the first frequency such that for all frequencies greater than B_3 the magnitude is attenuated by a factor greater than $1/\sqrt{2}$ from its value in $\omega = 0$, that is

$$\frac{|T(jB_3)|}{|T(j0)|} = \frac{1}{\sqrt{2}} \quad \text{and} \quad \frac{|T(j\omega)|}{|T(j0)|} < \frac{1}{\sqrt{2}} \quad for \ all \ \omega > B_3$$

or equivalently in decibel

$$|T(jB_3)|_{db} = |T(j0)|_{db} - 3_{db}$$



A In order to evaluate the bandwidth B_3 and the resonant peak M_p of the complementary sensitivity function T(s), it is important to have an idea of the behavior of T(s) assuming a regular stable open loop function F(s).





Open and closed loop transfer function

▲ Taking into account that



where we consider

- * Low frequencies all the ω at least one decade below the crossing frequency $ω_c$
- * High frequencies all the ω at least one decade above the crossing frequency ω_c







- ▲ Taking into account that $|T(j0)|_{db} \cong 0$
 - * the bandwidth B_3 of T(s) is the first frequency such that for all frequencies greater than B_3 the magnitude is less than -3db
 - * the resonant peak M_p of T(s) is the maximum value assumed by the magnitude of T(s)

▲ In order to quantify the bandwidth B_3 and resonant peak M_p of T(s) we need to analyze the behavior of T(s) in the two decades with center ω_c

▲ To do aim, we will introduce the so called *Nichols chart* that relates the magnitude and phase of the open loop function F(s) to the the magnitude and phase of the closed loop function T(s)



Nichols plots

- A The Nichols plots are an alternative solution to the Bode diagrams for the representation of the transfer functions with $s = j\omega$.
- ▲ The Nichols plots are composed by:

✤ x-axis reporting the phase of the transfer function

♦ y-axis reporting the magnitude in decibel of the transfer function

A In a Nichols plot magnitude and phase of transfer function are represented by a curve parametrized in ω.





Nichols plot: example 1

▲ The Nichols plot can be obtained from the Bode plots of magnitude and phase.



A Note that a point on the Nichols plots can also indicate the value of $F(j\omega)$ in a finite interval of ω .



Nichols plot: example 2

\checkmark Let us consider the transfer function







Nichols plot: example 3

 \blacktriangle Frequency function with resonance







- ▲ Nichols charts consist of constant-magnitude loci and constant phase loci of the closed loop system T(s) as a function of the open loop system F(s)
- ▲ Let us indicate the magnitude and phase of $F(j\omega)$ and $T(j\omega)$ with

$$F(j\omega) = A(j\omega)e^{\alpha(j\omega)}$$
 and $T(j\omega) = M(j\omega)e^{\varphi(j\omega)}$

▲ Taking into account that

$$T(j\omega) = \frac{F(j\omega)}{1 + F(j\omega)}$$

we have that

$$M(j\omega) = \frac{A(j\omega)}{\sqrt{1 + A(j\omega)^2 + 2A(j\omega)\cos(\alpha(j\omega))}}$$
$$\varphi(j\omega) = \operatorname{arctg}\left(\frac{\sin(\alpha(j\omega))}{A + \cos(\alpha(j\omega))}\right)$$



- ▲ For a fixed value of $\overline{\omega}$, we can
 - \Rightarrow identify a point on the Nichols plot of the open loop function $F(j\omega)$
 - * associate to this point the corresponding value of magnitude and phase of the closed loop function $T(j\omega)$.



A Repeating this procedure on a grid on points of the Nichols plot of $F(j\omega)$ it will be possible to find the constant-magnitude loci and constant phase loci of the closed loop system T(s)







▲ Making use of the previous relations

$$M(j\omega) = \frac{A(j\omega)}{\sqrt{1 + A(j\omega)^2 + 2A(j\omega)\cos(\alpha(j\omega))}}$$

$$\varphi(j\omega) = \operatorname{arctg}\left(\frac{\sin(\alpha(j\omega))}{A + \cos(\alpha(j\omega))}\right)$$
we can notice that:

$$* A \gg 1 \rightarrow \begin{cases} M \approx 1\\ \varphi \approx 0 \end{cases}$$

$$* A \ll 1 \rightarrow \begin{cases} M \approx A\\ \varphi \approx \alpha \end{cases}$$



Nichols chart: example

\checkmark Let us consider the transfer function

 $F(s) = \frac{3}{(1+s)^3}$



Open loop function $F(j\omega)$

Closed loop function $T(j\omega)$