



Course of
"Automatic Control Systems"
2022/23

Real Bode diagrams – Fourier analysis

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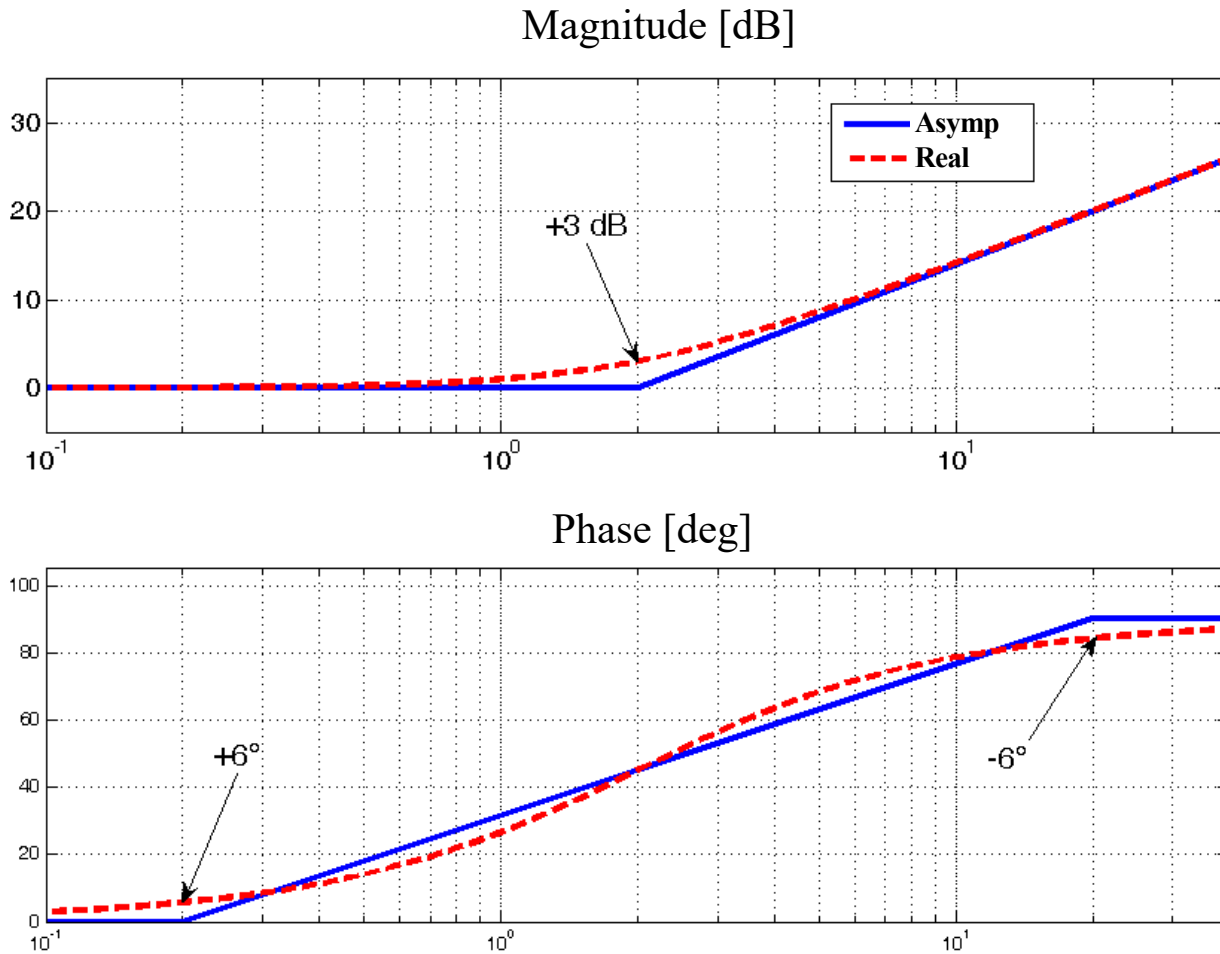
Team code: **uxbsz19**



Real Bode diagrams

- ✦ *In the real Bode diagrams the magnitude and phase of a transfer function $W(s)$ with $s = j\omega$ are drawn accurately* also the in two decades around the break points of the binomial and trinomial terms.
- ✦ The real Bode diagrams are usually traced applying some *corrections to the asymptotic Bode diagrams*
- ✦ The real Bode diagrams *can be drawn in MATLAB using the command 'bode'*

Zero of multiplicity one $W(s) = (1 + s\tau)$



This result can be easily generalized to a generic binomial term



Real Bode diagrams: trinomial term

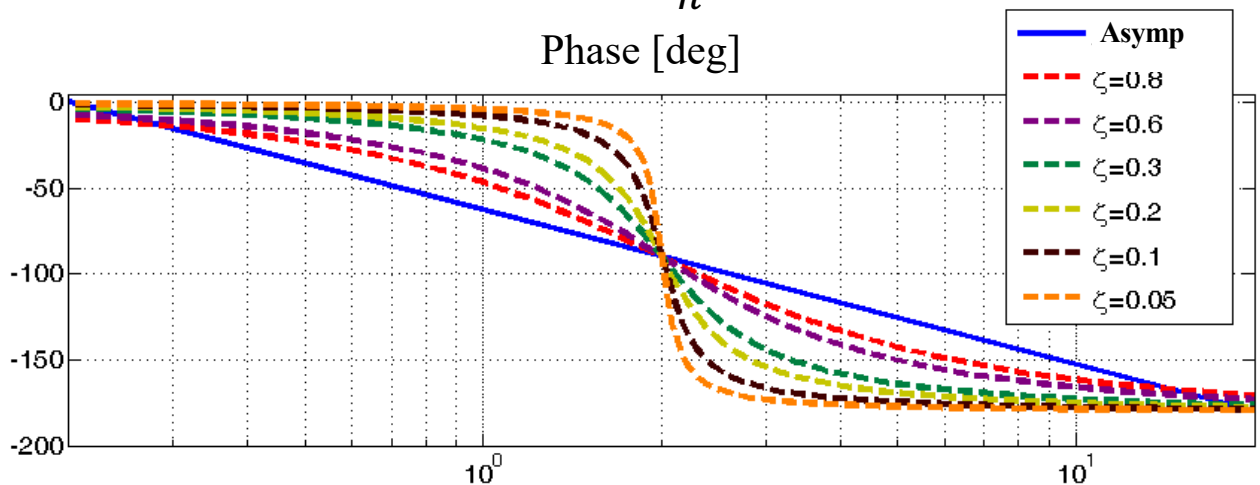
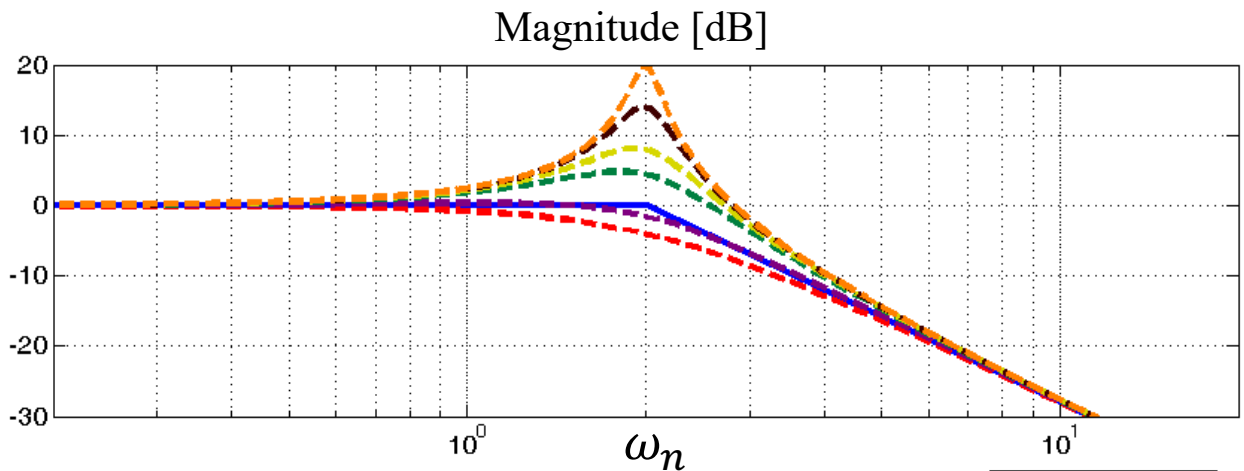
Complex conjugate poles of multiplicity one $W(s) = \left(1 + \frac{2\zeta s}{\omega_n} + \frac{s^2}{\omega_n^2}\right)^{-1}$

Peak module

$$M_p = \frac{1}{2\zeta\sqrt{1-2\zeta^2}}$$

Peak frequency

$$\omega_p = \omega_n\sqrt{1-2\zeta^2}$$



This result can be easily generalized to a generic trinomial term



Bode magnitude table

- ✦ Monomial terms of multiplicity 1. The slope is constant in $\omega \in [0 \infty[$

Zero in the origin	+20 dB/decade
Pole in the origin	-20 dB/decade

- ✦ Binomial and trinomial terms of multiplicity 1. The slope changes on the break point

	<u>Independent from the sign of the real part</u>
Real Zero	+20 dB/decade
Real Pole	-20 dB/decade
Comp. Conjug. zeros	+40 dB/decade
Comp. Conjug. poles	-40 dB/decade

- ✦ When the term has a multiplicity greater than one, the slopes should be multiplied by the multiplicity.



Bode phase table

- Constant and monomial terms of multiplicity 1. The slope is constant in $\omega \in [0, \infty[$

K < 0	-180° per $\omega \in [0, \infty)$
Zero in the origin	+90° per $\omega \in [0, \infty)$
Pole in the origin	-90° per $\omega \in [0, \infty)$

- Binomial and trinomial terms of multiplicity 1. The slope changes one decade before and after the breaking point.

	Negative real part	Positive real part
Real Zero	+90° +45° → -45° /decade	-90° -45° → +45° /decade
Real Pole	-90° -45° → +45° /decade	+90° +45° → -45° /decade
Comp. Conjug. zeros	+180° +90° → -90° /decade	-180° -90° → +90° /decade
Comp. Conjug. poles	-180° -90° → +90° /decade	+180° +90° → -90° /decade

- When the term has a multiplicity greater than one, the phase variation should be multiplied by the multiplicity.



Examples

✦ Trace the real Bode diagrams of the functions

$$W(s) = \frac{1000(s + 0.5)}{s(s^2 + 10s + 100)}$$

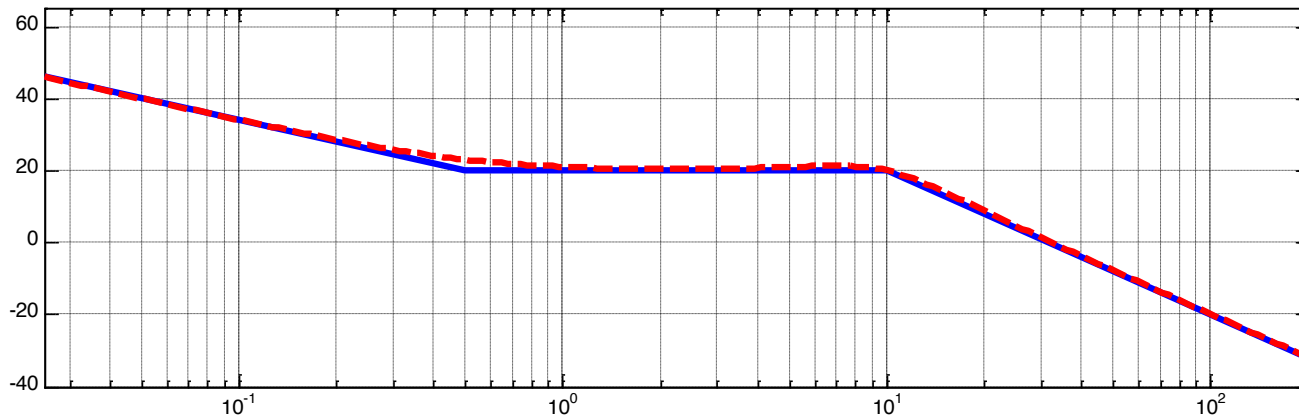
$$W(s) = \frac{s(s - 2)}{(s^2 + 5s + 25)}$$



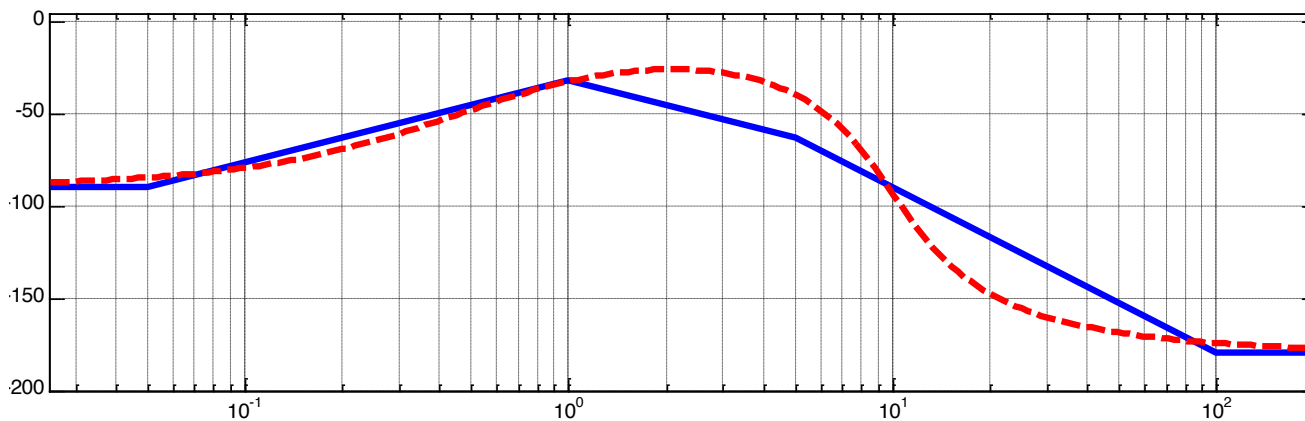
Example 1

$$W(s) = \frac{1000(s + 0.5)}{s(s^2 + 10s + 100)}$$

Magnitude [dB]



Phase [deg]

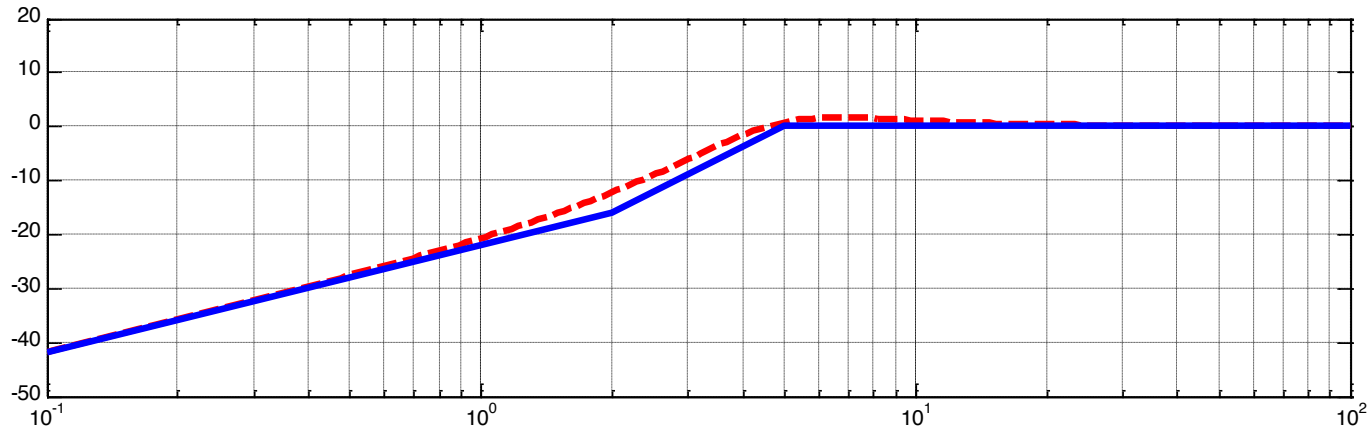




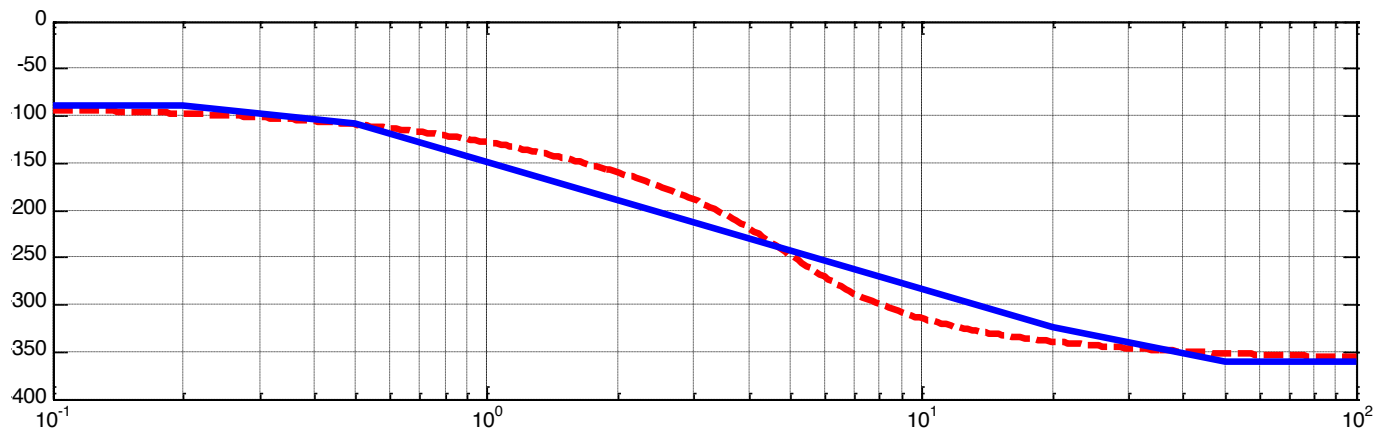
Exmple 2

$$W(s) = \frac{s(s-2)}{(s^2 + 5s + 25)}$$

Magnitude [dB]

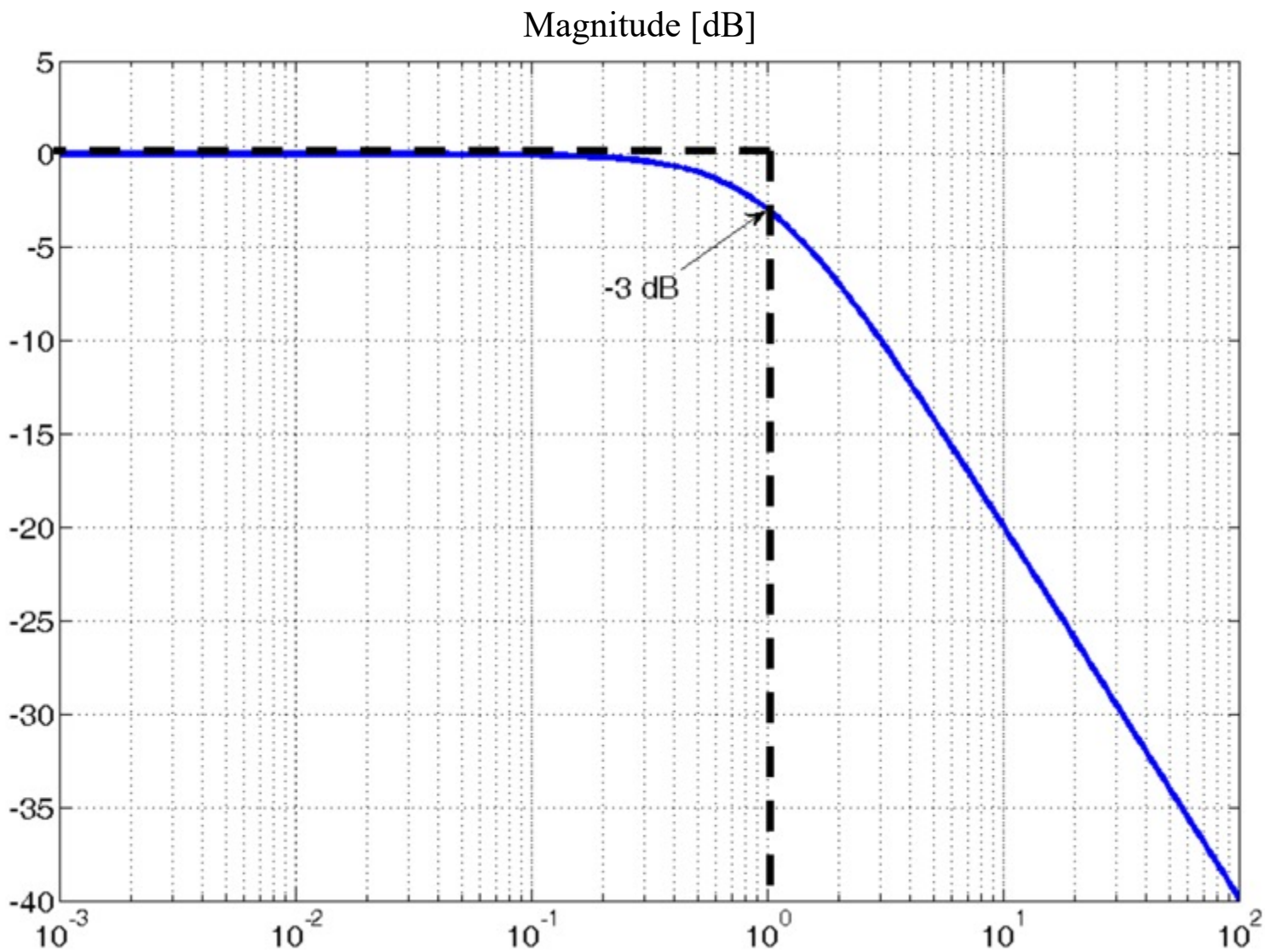


Phase [deg]





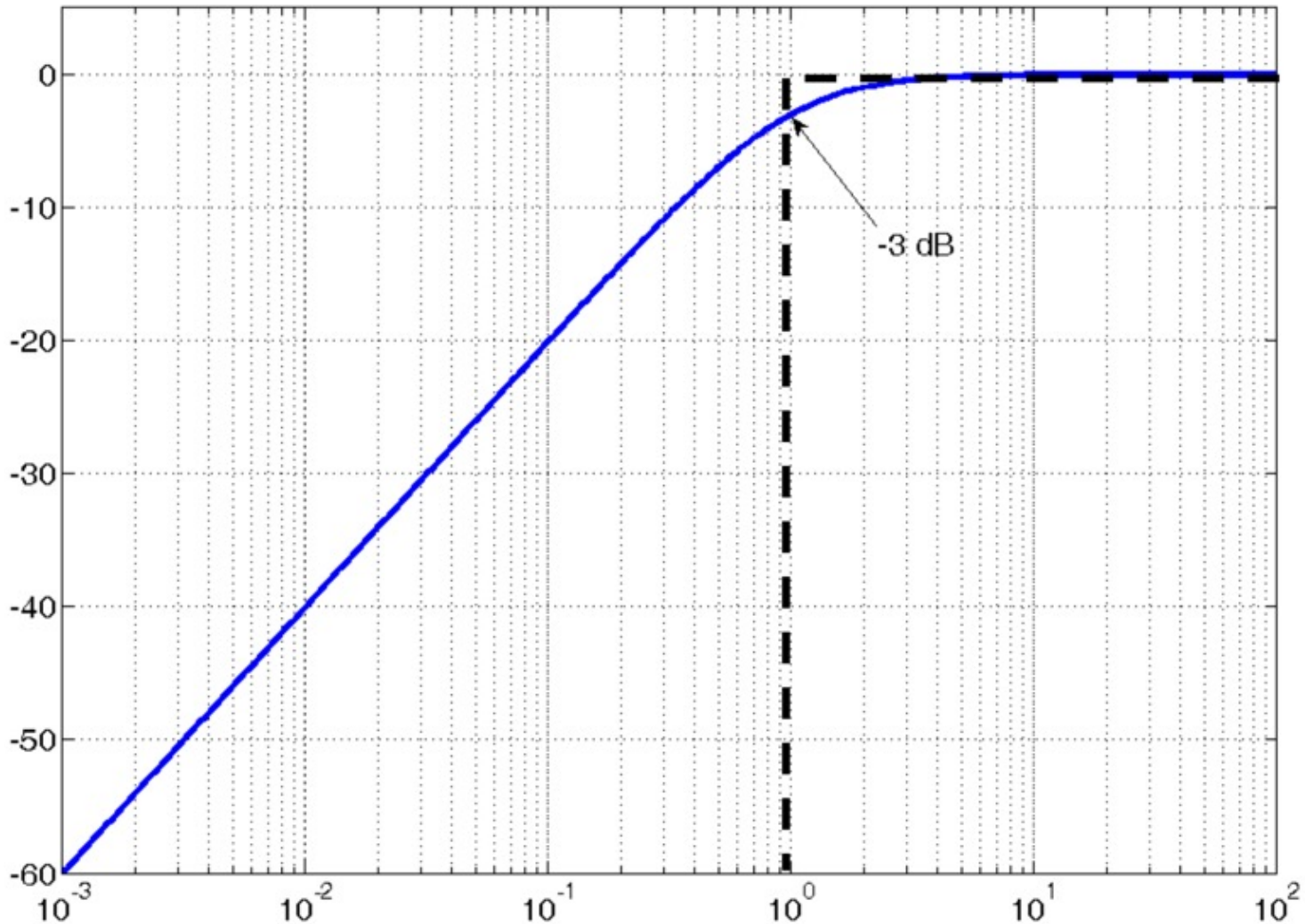
Low-pass filter





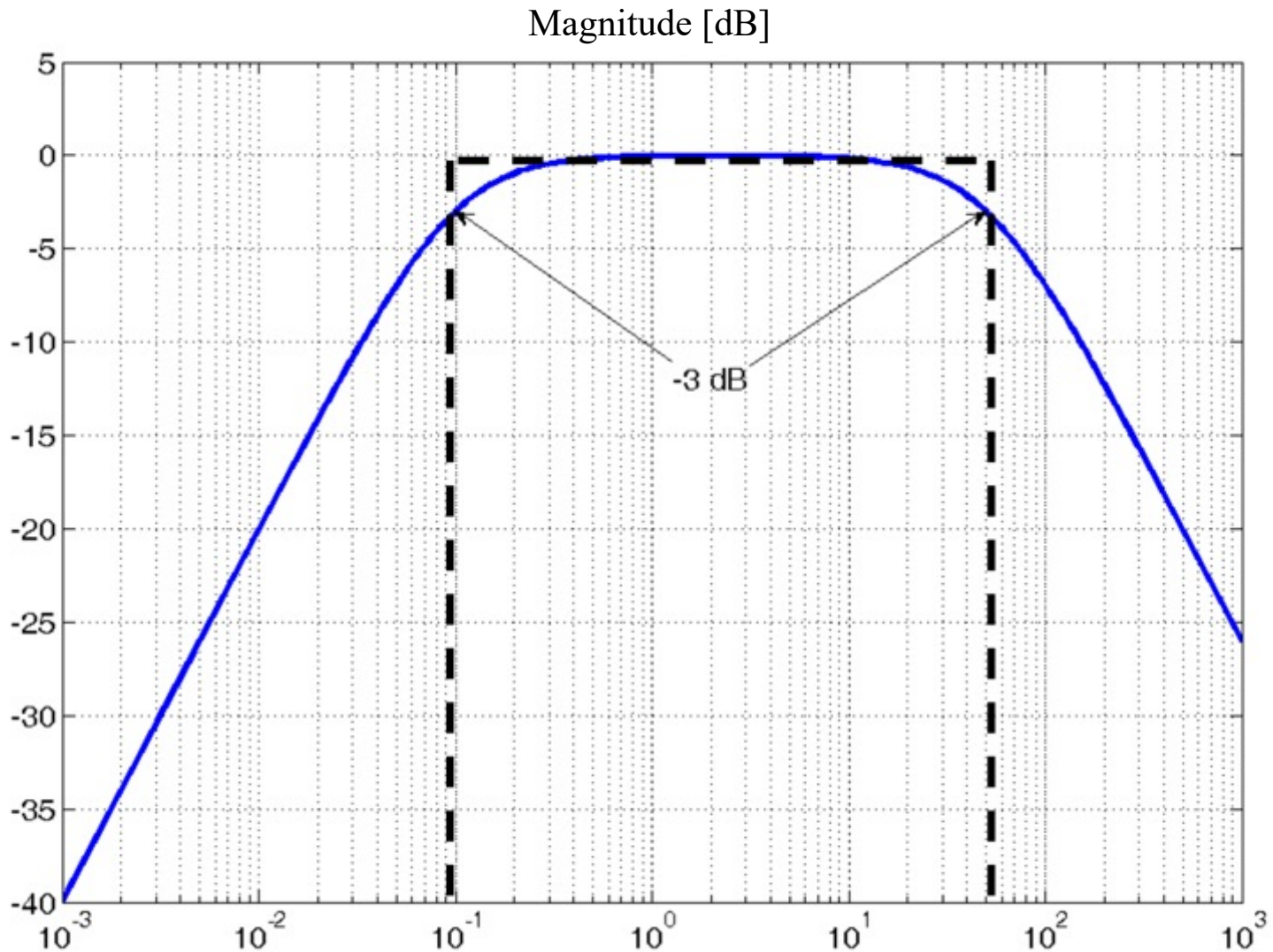
High-pass filter

Magnitude [dB]





Band-pass filter





Fourier analysis

- Any periodic function $f(t)$ with period T ,

$$f(t) = f(t + T),$$

can be written as

$$f(t) = F_0 + \sum_{n=1}^{\infty} [F_{cn} \cos(n\omega_0 t) + F_{sn} \sin(n\omega_0 t)]$$

where $\omega_0 = \frac{2\pi}{T}$,

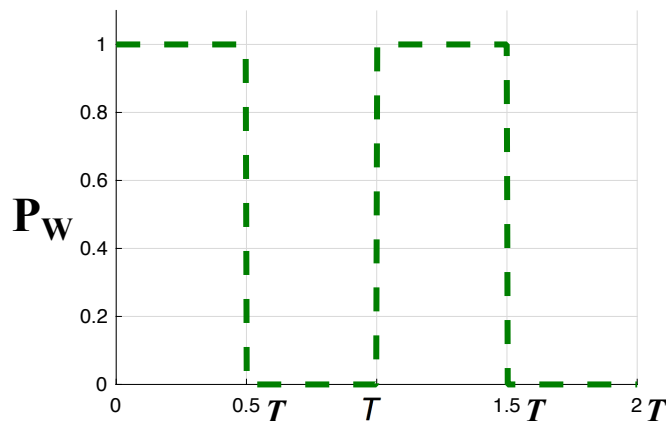
$$F_0 = \frac{1}{T} \int_T f(t) dt \quad F_{cn} = \frac{2}{T} \int_T f(t) \cos(n\omega_0 t) dt \quad F_{sn} = \frac{2}{T} \int_T f(t) \sin(n\omega_0 t) dt .$$

F_0 is the average value of f over a single period.

The component with ω_0 is the fundamental harmonic or 1st harmonic, that with $n\omega_0$ is n-th harmonic.



Example: square wave



$$P_w(t) = \begin{cases} 1 & \text{if } 0 < t \leq T/2 \\ 0 & \text{if } T/2 < t \leq T \end{cases}$$

Using Fourier analysis:

$$F_0 = \frac{1}{2}, \quad F_{cn} = 0 \quad \forall n \in \mathbb{N}, \quad F_{sn} = \begin{cases} \frac{2}{n\pi} & \text{if } n \text{ is odd} \\ 0 & \text{if } n \text{ is even} \end{cases}.$$

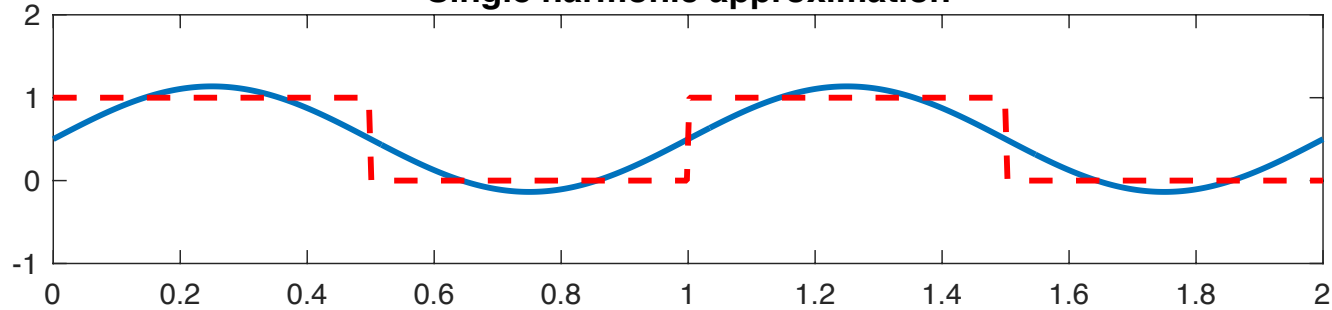
Therefore, the square wave can be written

$$P_w(t) = \frac{1}{2} + \frac{2}{\pi} \sin(\omega_0 t) + \frac{2}{3\pi} \sin(3\omega_0 t) + \frac{2}{5\pi} \sin(5\omega_0 t) + \dots$$



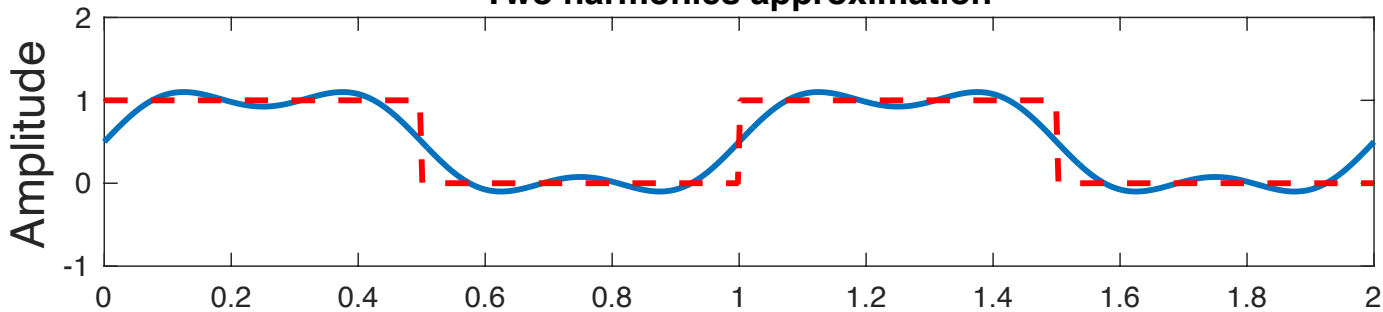
Example: approximation of a square wave

Single harmonic approximation



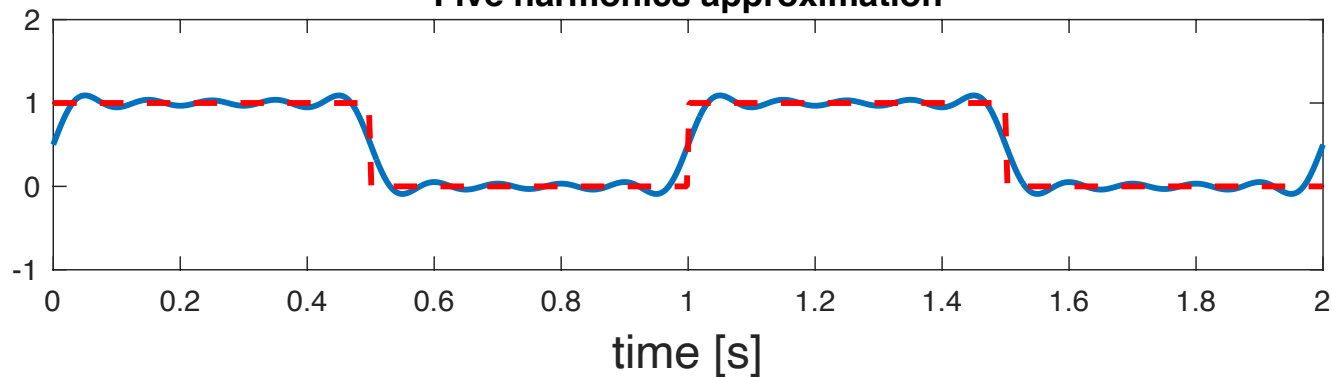
$$\omega_0 = 2\pi \text{ rad/s}$$

Two harmonics approximation



$$\omega_0, 3\omega_0$$

Five harmonics approximation



$$\omega_0, 3\omega_0, \\ 5\omega_0, 7\omega_0, \\ 9\omega_0,$$



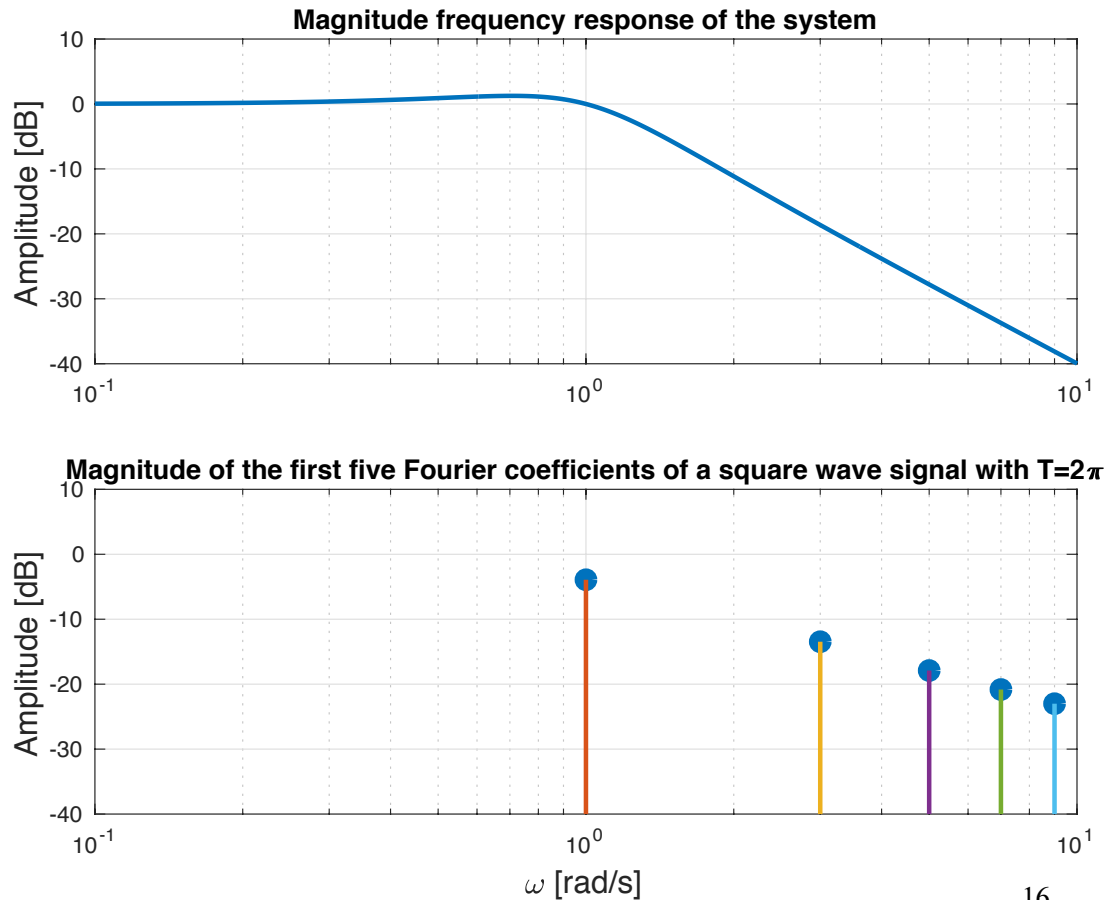
Example: steady state response to a square wave

Let us consider the system with transfer function:

$$G(s) = \frac{1}{s^2 + s + 1}$$

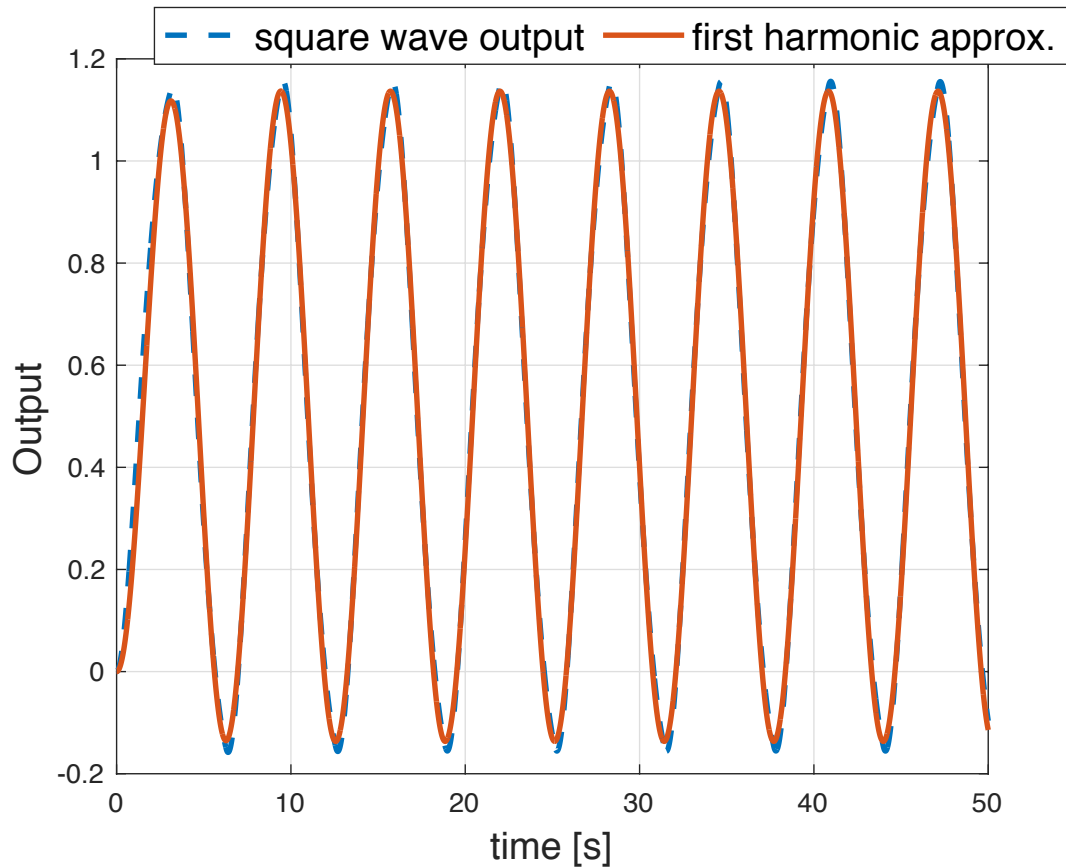
and assume we want to compute the steady-state response to the square wave with period $T=2\pi$.

$$\begin{aligned} \bullet \quad u(t) &= \frac{1}{2} + \frac{2}{\pi} \sin t \\ &\quad + \frac{2}{3\pi} \sin(3t) + \\ &\quad + \frac{2}{5\pi} \sin(5t) + \dots \end{aligned}$$





Example: steady state response to a square wave



The steady state response of the system with transfer function $G(s) = \frac{1}{s^2 + s + 1}$

is practically identical to the response assuming just the first two terms of the Fourier expansion (the average value plus the first harmonic)