

Campi Elettromagnetici

**Corso di Laurea in Ingegneria Informatica,
Biomedica e delle Telecomunicazioni**

a.a. 2019-2020 - Laurea “Triennale” – Secondo semestre - Secondo anno

Università degli Studi di Napoli “Parthenope”

Stefano Perna

Color legend

New formulas, important considerations,
important formulas, important concepts

Very important for the discussion

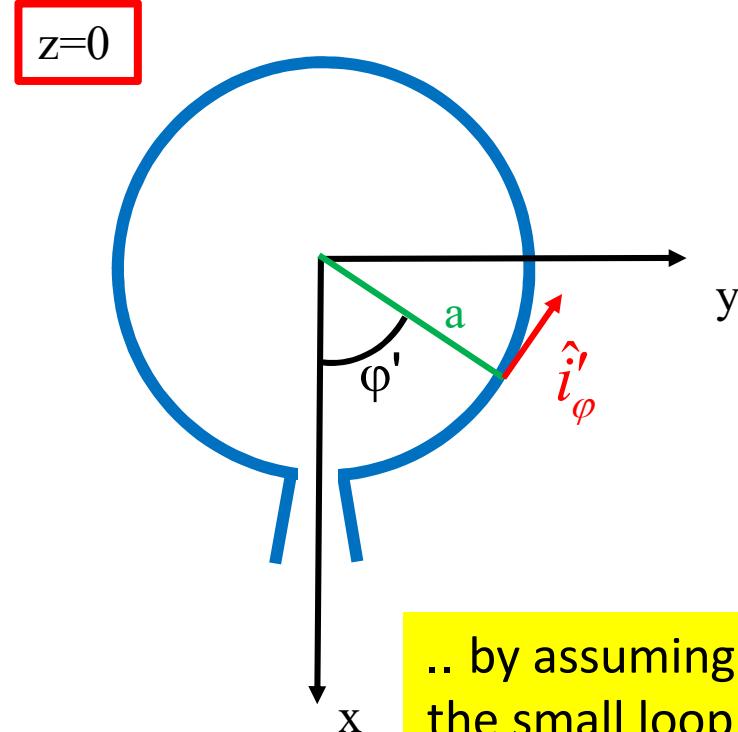
Memo

Mathematical tools to be exploited

Mathematics

Small loop antenna

$$\mathbf{J} = I\delta(z)\delta(r-a)\hat{i}'_\phi$$



$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J} \quad \text{Vector Helmholtz equation}$$

J

$$\mathbf{A} \approx \frac{j\beta I \mu A S}{4\pi} \frac{e^{-j\beta r}}{r} \left[1 + \frac{1}{j\beta r} \right] \sin \theta \hat{i}_\phi$$

$$\mathbf{H} = \frac{1}{\mu} \nabla \times \mathbf{A} \quad \text{and} \quad \mathbf{E} = -j\omega \mathbf{A} + \frac{\nabla \times \mathbf{A}}{j\omega \epsilon \mu}$$

E

H

.. by assuming that the current I in the small loop is constant and that the radius of the loop $a \ll \lambda$

Small loop antenna

The E.M. field radiated by the small loop antenna

$$\begin{aligned}\vec{\mathbf{H}}(\vec{r}) &= H_r(r, \vartheta) \hat{i}_r + H_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{E}}(\vec{r}) &= E_\varphi(r, \vartheta) \hat{i}_\varphi\end{aligned}\quad \begin{cases} H_r = \frac{I\Delta S}{2\pi} \frac{(j\beta)^2}{r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta = \frac{I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\varphi = -\frac{\zeta I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

Small loop antenna: far field

The E.M. field radiated by the small loop antenna

$$\begin{aligned}\vec{\mathbf{H}}(\vec{r}) &= H_r(r, \vartheta) \hat{i}_r + H_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{E}}(\vec{r}) &= E_\phi(r, \vartheta) \hat{i}_\phi\end{aligned}\quad \begin{cases} H_r = \frac{I\Delta S}{2\pi} \frac{(j\beta)^2}{r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta = \frac{I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\phi = -\frac{\zeta I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

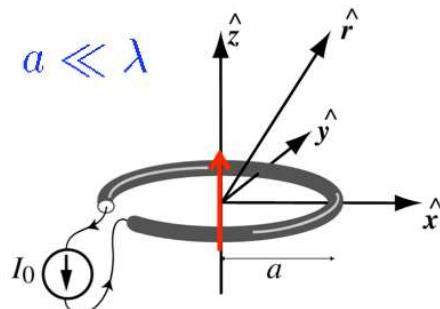
... for $r \gg \lambda$ ($\beta r \gg 1$) simplifies as

$$\begin{cases} H_r = 0 \\ H_\vartheta = -\frac{\beta \Delta s I}{2\lambda r} \sin \vartheta \exp(-j\beta r) = \frac{-E_\phi}{\zeta} \\ E_\phi = \frac{\zeta \beta \Delta s I}{2\lambda r} \sin \vartheta \exp(-j\beta r) \end{cases}$$

Small loop antenna: far field

In the far-field case ($r \gg \lambda$) the small loop antenna behaves as follows

$$\begin{aligned}\vec{\mathbf{E}}(\vec{\mathbf{r}}) &= E_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{H}}(\vec{\mathbf{r}}) &= H_\vartheta(r, \vartheta) \hat{i}_\vartheta\end{aligned}\quad \left\{ \begin{array}{l} E_\vartheta = \frac{\zeta \beta \Delta s I}{2\lambda r} \sin \vartheta \exp(-j\beta r) \\ H_\vartheta = -\frac{\beta \Delta s I}{2\lambda r} \sin \vartheta \exp(-j\beta r) = \frac{-E_\vartheta}{\zeta} \end{array} \right.$$



$$\Delta S = \pi a^2$$

- the e.m. field propagates along \hat{i}_r
- the e.m. field lies on the plane orthogonal to the propagation direction
- $|E|$ and $|H|$ exhibit the decaying factor $1/r$
- $|E|$ and $|H|$ are proportional through ζ

$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$

$$\vec{\mathbf{S}} = \frac{1}{2\zeta} |\vec{\mathbf{E}}|^2 \hat{i}_r = \frac{\zeta}{2} |\vec{\mathbf{H}}|^2 \hat{i}_r$$

Small loop antenna: power flux

$$\begin{cases} H_r = \frac{I\Delta S}{2\pi} \frac{(j\beta)^2}{r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta = \frac{I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\varphi = -\frac{\zeta I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

$$P = \frac{1}{2} \oint_S \left[\mathbf{E} \times \mathbf{H}^* \right] \cdot \hat{i}_r dS = -\frac{1}{2} \int_0^{2\pi} d\varphi \int_0^\pi d\vartheta r^2 \sin \vartheta E_\varphi H_\vartheta^*$$

$$dS = r^2 \sin \vartheta d\vartheta d\varphi$$

$$\left[\mathbf{E} \times \mathbf{H}^* \right] \cdot \hat{i}_r = \left[\left(E_\varphi \hat{i}_\varphi \right) \times \left(H_\vartheta \hat{i}_\vartheta + H_r \hat{i}_r \right)^* \right] \cdot \hat{i}_r = -E_\varphi H_\vartheta^* \hat{i}_r \cdot \hat{i}_r = -E_\varphi H_\vartheta^*$$

$$\begin{aligned} \hat{i}_\varphi &= \hat{i}_r \times \hat{i}_\vartheta \\ \hat{i}_\vartheta &= \hat{i}_\varphi \times \hat{i}_r \\ \hat{i}_r &= \hat{i}_\vartheta \times \hat{i}_\varphi \end{aligned}$$

Small loop antenna: power flux

$$\begin{cases} H_r = \frac{I\Delta S}{2\pi} \frac{(j\beta)^2}{r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta = \frac{I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\varphi = -\frac{\zeta I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

$$P = P_1 + jP_2$$

$$P_1 = \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\beta \Delta S}{\lambda} \right)^2 |I|^2$$

$$P_2 = \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\beta \Delta S}{\lambda} \right)^2 \frac{1}{(\beta r)^3} |I|^2$$

$$P = \frac{1}{2} \oint_S \left[\mathbf{E} \times \mathbf{H}^* \right] \cdot \hat{i}_r dS = -\frac{1}{2} \int_0^{2\pi} d\varphi \int_0^\pi d\vartheta r^2 \sin \vartheta E_\varphi H_\vartheta^* = \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\beta \Delta S}{\lambda} \right)^2 \left[1 + j \frac{1}{(\beta r)^3} \right] |I|^2$$

Small loop antenna: power flux

$$\begin{cases} H_r = \frac{I\Delta S}{2\pi} \frac{(j\beta)^2}{r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta = \frac{I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\varphi = -\frac{\zeta I\Delta S}{4\pi} \frac{(j\beta)^2}{r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

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- Note that in the far-field case only the first active power term exists and it does not depend on r
- Note that the real part of the power, in lossless medium, is independent of r , therefore if one consider two different spherical surfaces one gets the same result. Only the so-called radiative terms contribute.
- The reactive part depends on r . The reactive part depends on r . Its sign is positive showing that there is an excess of stored **magnetic** energy in the neighbor of the magnetic dipole (see Poynting's theorem)

Elementary electrical dipole vs. small loop antenna

Elementary electrical dipole

$$P = P_1 + jP_2$$

$$P_1 = \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\Delta z}{\lambda} \right)^2 |I|^2$$

$$P_2 = -\frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\Delta z}{\lambda} \right)^2 \frac{1}{(\beta r)^3} |I|^2$$

Small loop antenna

$$P = P_1 + jP_2$$

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Small loop antenna

WHY?

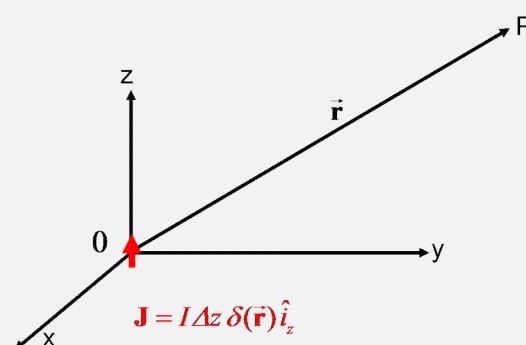


Small loop antenna

Elementary electrical dipole

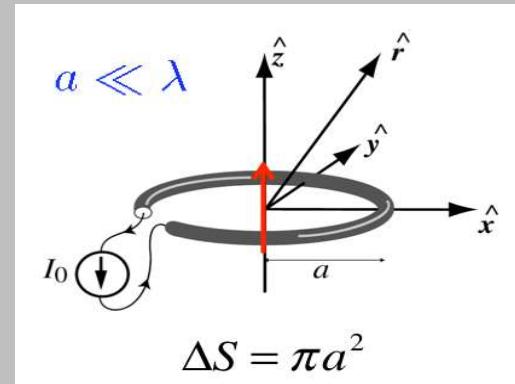
$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z$$

- Why are we interested in such a radiating element?
- Why is such a radiating element referred to as elementary electrical dipole?
- How can we physically approximate an elementary electrical dipole?



Small loop antenna

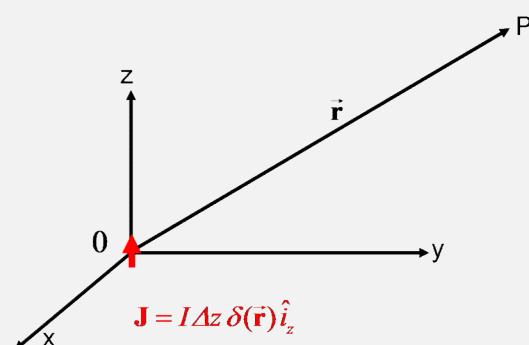
$$\mathbf{J} = I \delta(z) \delta(r - a) \hat{i}'_\phi$$



Small loop antenna

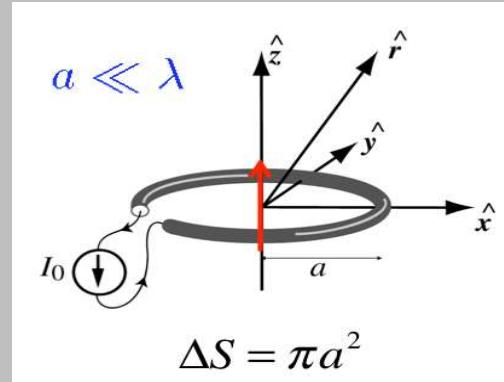
Elementary electrical dipole

$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z$$



Small loop antenna

$$\mathbf{J} = I \delta(z) \delta(r - a) \hat{i}'_\varphi$$

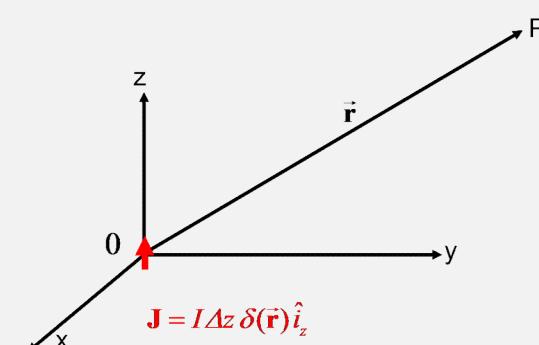


Elementary electrical dipole vs. small loop antenna

Elementary electrical dipole

$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z$$

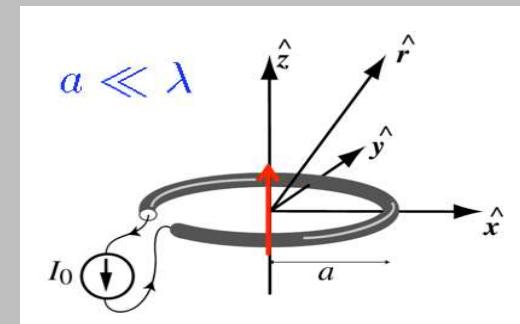
$$\begin{cases} E_r = \zeta \frac{I \Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\theta = \zeta \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\phi = \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$



Small loop antenna

$$\mathbf{J} = I \delta(z) \delta(r-a) \hat{i}'_\phi$$

$$\begin{cases} H_r = \frac{I \Delta S}{2\pi} j\beta \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ H_\theta = \frac{I \Delta S}{4\pi} j\beta \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ E_\phi = -\frac{\zeta I \Delta S}{4\pi} j\beta \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$



Elementary electrical dipole vs. small loop antenna

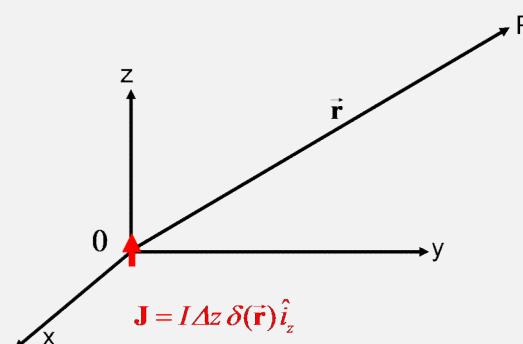
Elementary electrical dipole

$$\mathbf{J} = I \Delta z \delta(\vec{r}) \hat{i}_z$$

for $r \gg \lambda$

$$\mathbf{E} = \frac{j\zeta I}{2\lambda} \frac{\exp(-j\beta r)}{r} \Delta z \sin \vartheta \hat{i}_\vartheta$$

$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$



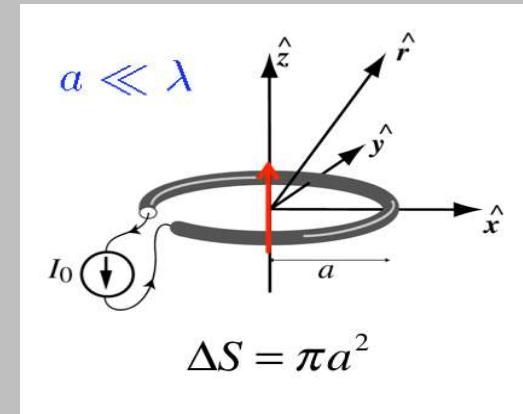
Small loop antenna

$$\mathbf{J} = I \delta(z) \delta(r - a) \hat{i}'_\varphi$$

for $r \gg \lambda$

$$\mathbf{E} = \frac{\zeta \beta \Delta S I}{2\lambda} \frac{\exp(-j\beta r)}{r} \sin \vartheta \hat{i}_\varphi$$

$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$



Elementary electrical dipole vs. small loop antenna

$$P = \frac{1}{2} \oint_S [\mathbf{E} \times \mathbf{H}^*] \cdot \hat{i}_r dS$$

Elementary electrical dipole

$$P = P_1 + jP_2$$

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Small loop antenna

WHY?



Magnetic Sources



$$\begin{cases} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} + \mathbf{J} \\ \nabla \cdot \epsilon\mathbf{E} = \rho \\ \nabla \cdot \mu\mathbf{H} = 0 \end{cases}$$

Magnetic Sources



$$\begin{cases} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} - \mathbf{J}_m \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} + \mathbf{J} \\ \nabla \cdot \epsilon\mathbf{E} = \rho \\ \nabla \cdot \mu\mathbf{H} = \rho_m \end{cases}$$

Magnetic Sources

What is the relation between sources and fields in this case?

$$\begin{cases} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} - \mathbf{J}_m \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} + \mathbf{J} \\ \nabla \cdot \epsilon\mathbf{E} = \rho \\ \nabla \cdot \mu\mathbf{H} = \rho_m \end{cases}$$

Magnetic Sources

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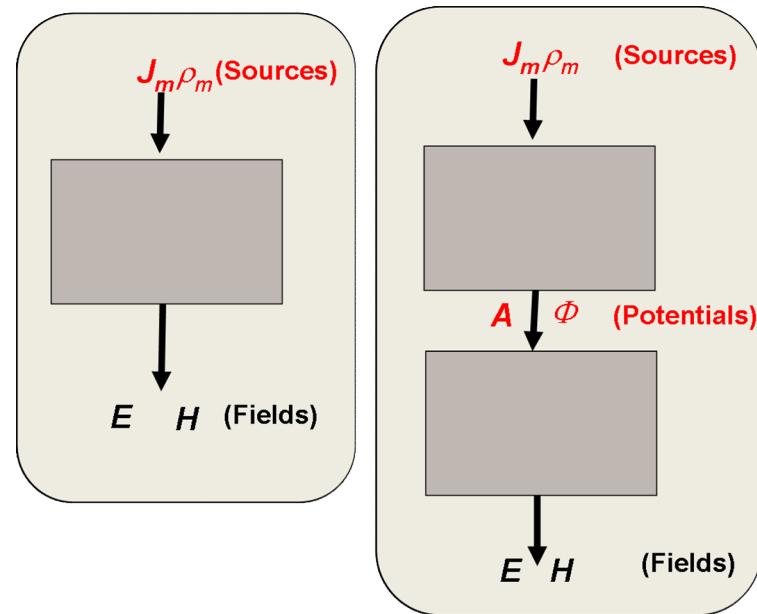
Let's simplify the question. What is the relation between sources and fields in this case?

$$\begin{cases} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} - \mathbf{J}_m \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} \\ \nabla \cdot \epsilon\mathbf{E} = 0 \\ \nabla \cdot \mu\mathbf{H} = \rho_m \end{cases}$$

Magnetic Sources

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In principle, we could replace the same approach as that exploited for the electric sources

Magnetic Sources

What is the relation between sources and fields in this case?

Let's simplify the question. What is the relation between sources and fields in this case?

$$\begin{cases} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} - \mathbf{J}_m \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} \\ \nabla \cdot \epsilon\mathbf{E} = 0 \\ \nabla \cdot \mu\mathbf{H} = \rho_m \end{cases}$$

In practice, we follow an easier way, provided by the duality theorem

Magnetic Sources

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In practice, we follow an easier way, provided by the duality theorem

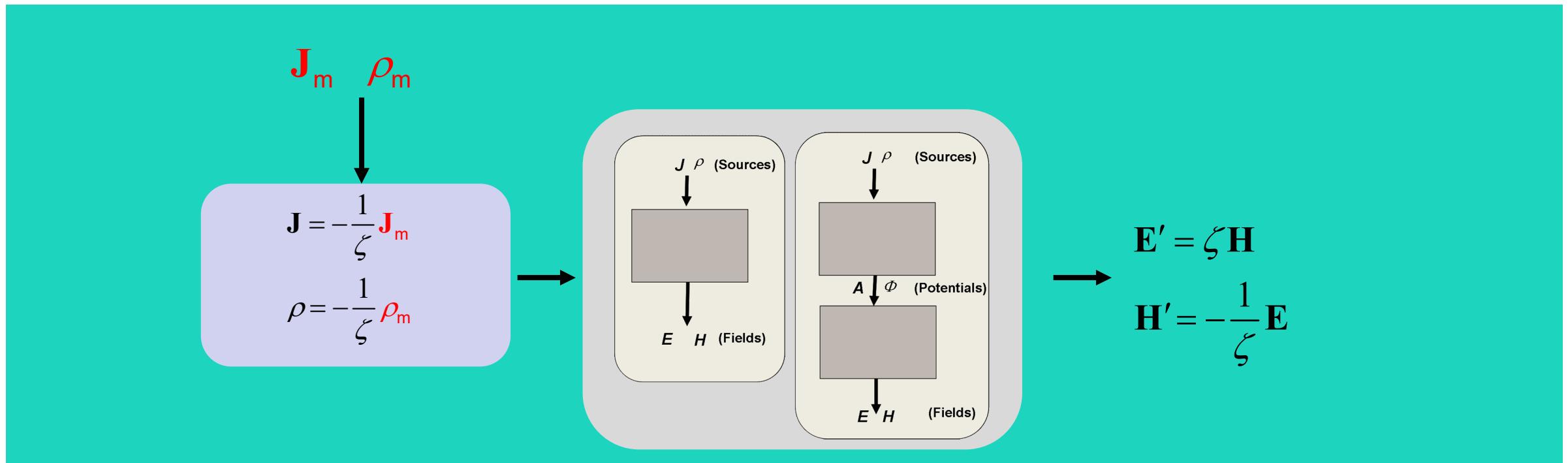
$$\begin{matrix} \mathbf{J} & \rho \\ \downarrow & \\ \mathbf{E}, \mathbf{H} & \end{matrix}$$



$$\begin{matrix} \mathbf{J}_m = -\zeta \mathbf{J} & \rho_m = -\zeta \rho \\ \downarrow & \\ \mathbf{E}' = \zeta \mathbf{H} & , \mathbf{H}' = -\frac{1}{\zeta} \mathbf{E} \end{matrix}$$

Duality Theorem

The diagram illustrates the Duality Theorem. On the left, a purple box contains the source terms \mathbf{J} , ρ pointing down to \mathbf{E}, \mathbf{H} . A large black arrow points to the right, leading to two green-bordered boxes. The top box contains the equations $\mathbf{J}_m = -\zeta \mathbf{J}$ and $\rho_m = -\zeta \rho$. The bottom box contains the equations $\mathbf{E}' = \zeta \mathbf{H}$ and $\mathbf{H}' = -\frac{1}{\zeta} \mathbf{E}$.



Elementary electrical and magnetic dipoles

Elementary electrical dipole

$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z = I \Delta z \delta(x) \delta(y) \delta(z) \hat{i}_z$$

$$I \Delta z = j \omega Q \Delta z = j \omega U$$

Elementary magnetic dipole

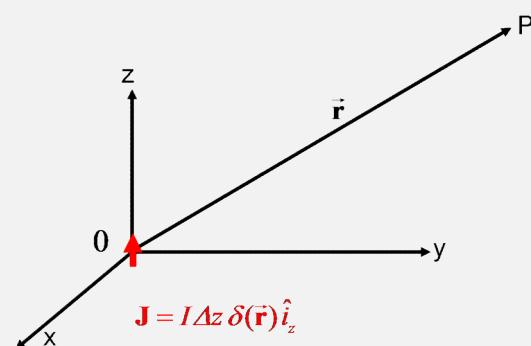
$$\mathbf{J}_m = I_m \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z = I_m \Delta z \delta(x) \delta(y) \delta(z) \hat{i}_z$$

$$I_m \Delta z = j \omega U_m$$

Elementary electrical and magnetic dipoles

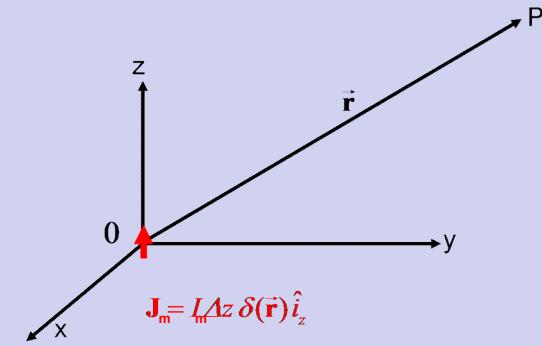
Elementary electrical dipole

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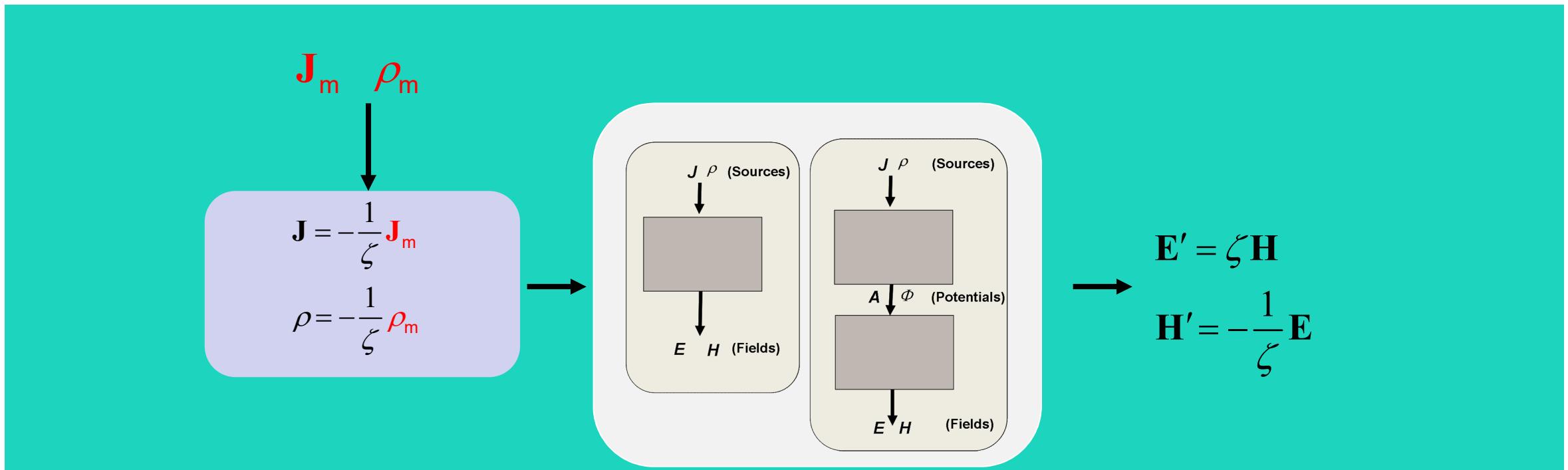
Elementary magnetic dipole

$$\mathbf{J}_m = I_m \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z$$



Duality Theorem

$$\begin{array}{ccc} \mathbf{J} & \rho \\ \downarrow & \\ \mathbf{E}, \mathbf{H} & \end{array} \longrightarrow \begin{array}{ccc} \mathbf{J}_m = -\zeta \mathbf{J} & \rho_m = -\zeta \rho \\ \downarrow & \\ \mathbf{E}' = \zeta \mathbf{H} & , \mathbf{H}' = -\frac{1}{\zeta} \mathbf{E} \end{array}$$



Elementary electrical and magnetic dipoles

Ampere equivalence theorem

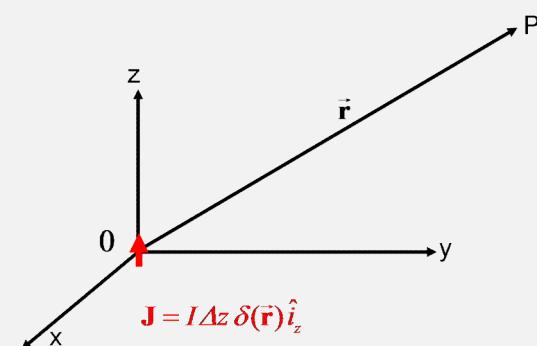
By applying the Duality theorem it turns out that the small loop antenna is equivalent to an elementary magnetic dipole, provided that:

$$U_m = \mu I \Delta S$$

Elementary electrical and magnetic dipoles

Elementary electrical dipole

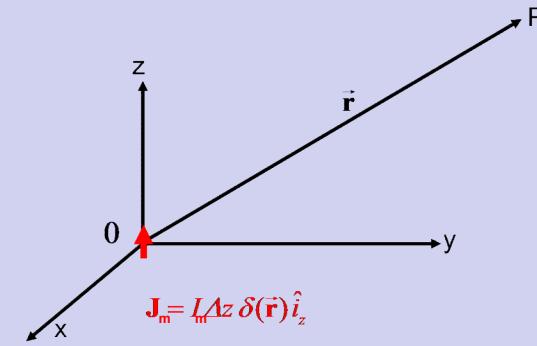
$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z$$



- Why are we interested in such a radiating element?
- Why is such a radiating element referred to as **elementary electrical dipole**?
- How can we physically approximate an **elementary electrical dipole**?

Elementary magnetic dipole

$$\mathbf{J}_m = I_m \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z$$



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- How can we physically approximate an **elementary magnetic dipole**?

Color legend

New formulas, important considerations,
important formulas, important concepts

Very important for the discussion

Memo

Mathematical tools to be exploited

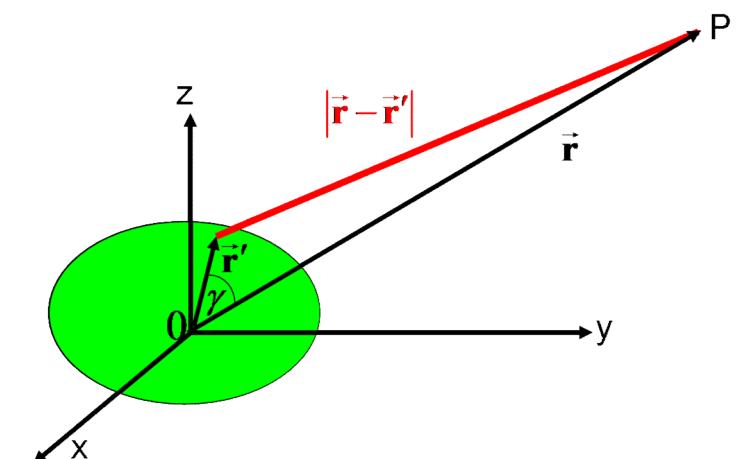
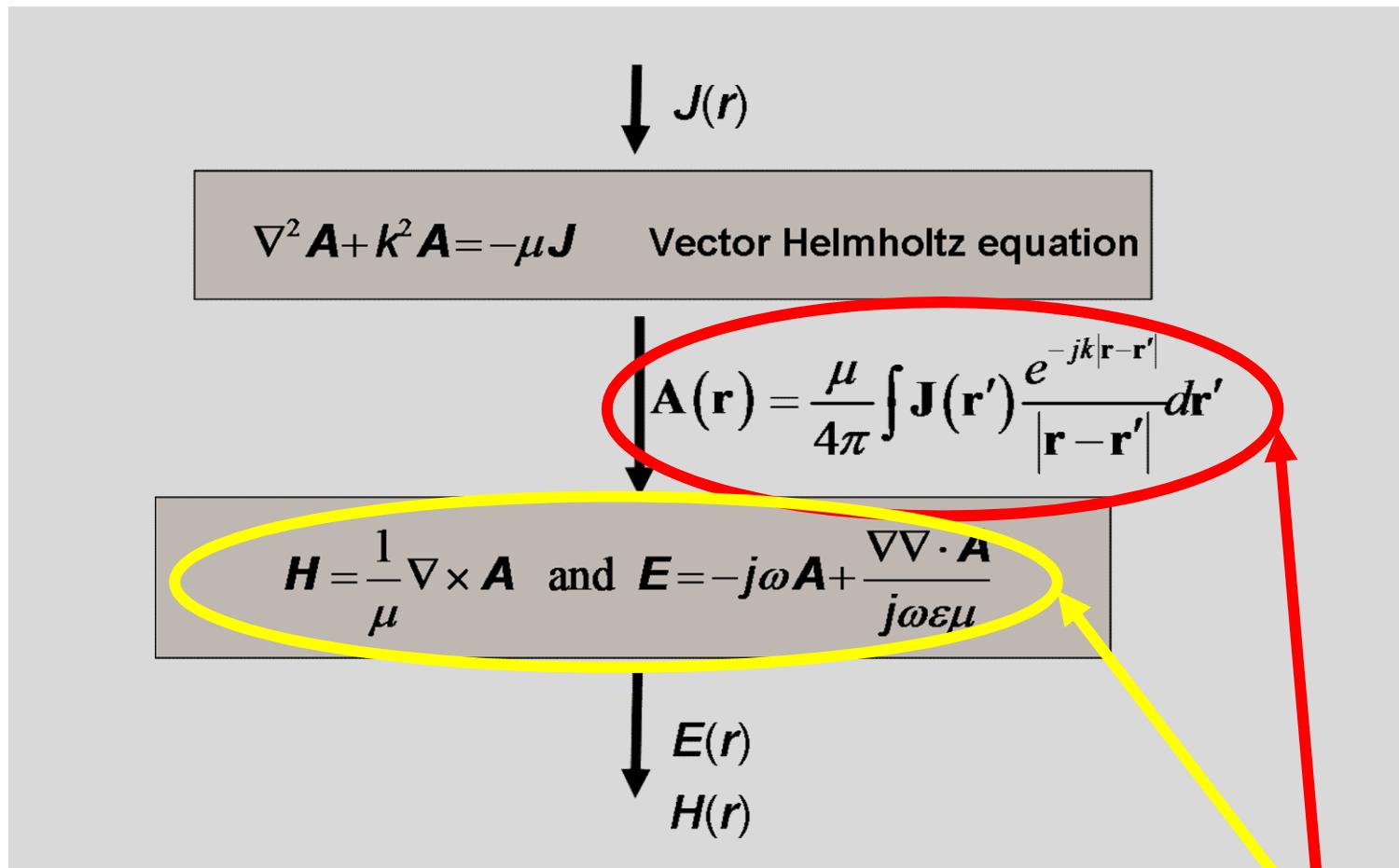
Mathematics

Outline

- Radiation problem for extended antennas
- Field regions



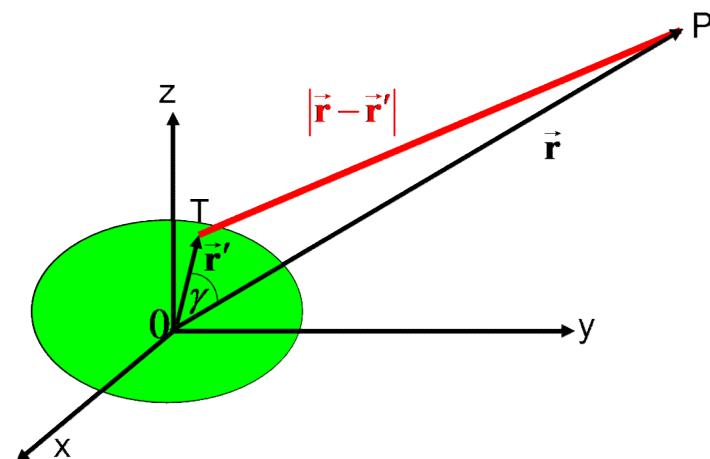
Extended antennas



Is it possible to simplify the expressions of the fields, possibly via proper approximation of the vector potential \mathbf{A} ?

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

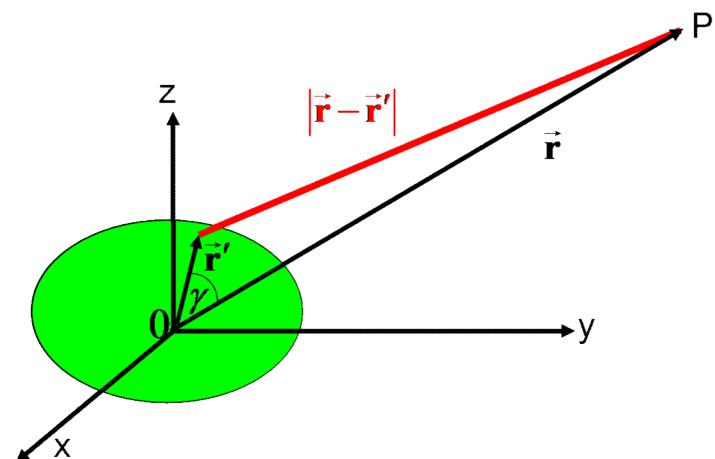


$$\begin{aligned}
 |\vec{r}| &= r & |\vec{r}'| &= r' \\
 |\vec{r} - \vec{r}'| &= \sqrt{r^2 + (r')^2 - 2rr' \cos \gamma} & &= \sqrt{r^2 \left[1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma \right]} \\
 &= r \sqrt{1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma}
 \end{aligned}$$

$$c = \sqrt{a^2 + b^2 - 2ab \cos \gamma}$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$



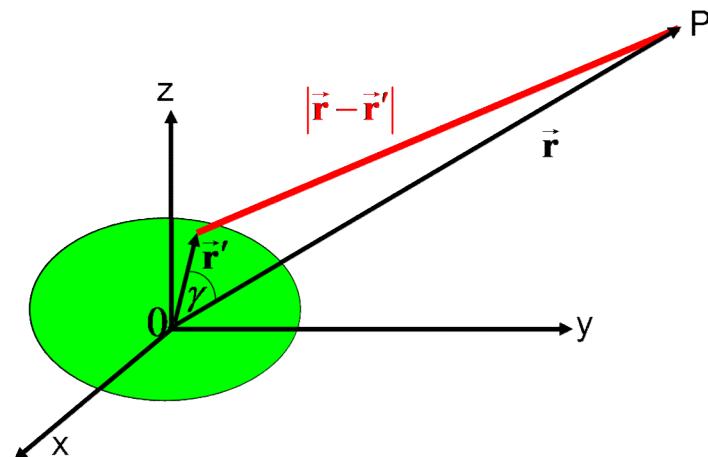
$$\begin{aligned}
 |\vec{r}| &= r & |\vec{r}'| &= r' \\
 |\vec{r} - \vec{r}'| &= \sqrt{r^2 + (r')^2 - 2rr' \cos \gamma} & &= \sqrt{r^2 \left[1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma \right]} \\
 &= r \sqrt{1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma} \\
 \sqrt{1+x} &= 1 + \frac{1}{2}x - \frac{1}{8}x^2 + \dots
 \end{aligned}$$

$$\begin{aligned}
 \sqrt{1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma} &= 1 + \frac{1}{2} \left[\left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma \right] - \frac{1}{8} \left[\left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma \right]^2 + \dots = 1 + \frac{1}{2} \left(\frac{r'}{r} \right)^2 - \frac{r'}{r} \cos \gamma - \frac{1}{8} \left[\cancel{\left(\frac{r'}{r} \right)^4} + 4 \left(\frac{r'}{r} \right)^2 \cos^2 \gamma - \cancel{4 \left(\frac{r'}{r} \right)^3 \cos \gamma} \right] + \dots \\
 &= 1 + \frac{1}{2} \left(\frac{r'}{r} \right)^2 - \frac{r'}{r} \cos \gamma - \frac{1}{2} \left(\frac{r'}{r} \right)^2 \cos^2 \gamma + \dots = 1 - \frac{r'}{r} \cos \gamma + \frac{1}{2} \left(\frac{r'}{r} \right)^2 (1 - \cos^2 \gamma) + \dots = 1 - \frac{r'}{r} \cos \gamma + \frac{1}{2} \left(\frac{r'}{r} \right)^2 \sin^2 \gamma + \dots
 \end{aligned}$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$|\vec{r}| = r \quad |\vec{r}'| = r'$$

$$|\vec{r} - \vec{r}'| = \sqrt{r^2 + (r')^2 - 2rr' \cos \gamma} = \sqrt{r^2 \left[1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma \right]}$$

$$= r \sqrt{1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma} = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$

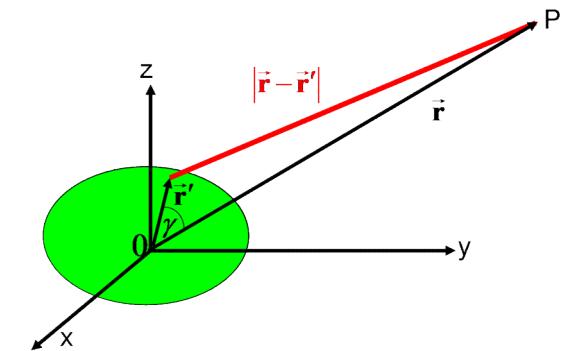
$$\sqrt{1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos \gamma}$$

$$= 1 - \frac{r'}{r} \cos \gamma + \frac{1}{2} \left(\frac{r'}{r} \right)^2 \sin^2 \gamma + \dots$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



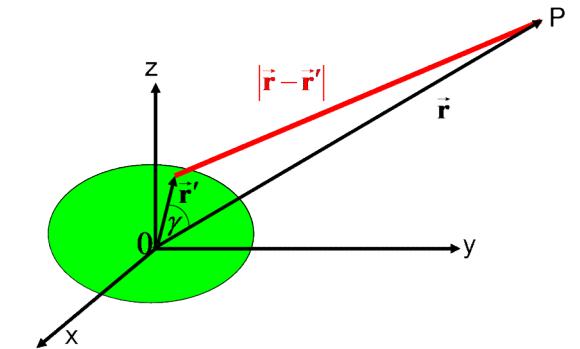
$$\frac{1}{|\vec{r}-\vec{r}'|} =$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} =$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



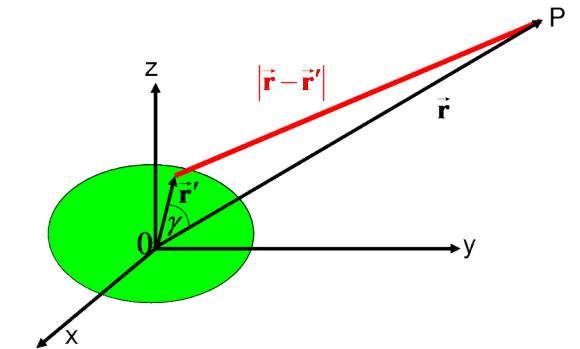
$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r}$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

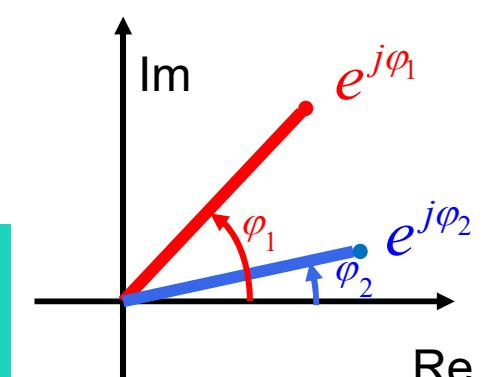
$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } r' \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{\cancel{j\beta r' \cos \gamma}} e^{\cancel{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma}} \dots \approx e^{-j\beta r} \quad \text{if } r' \ll \lambda$$

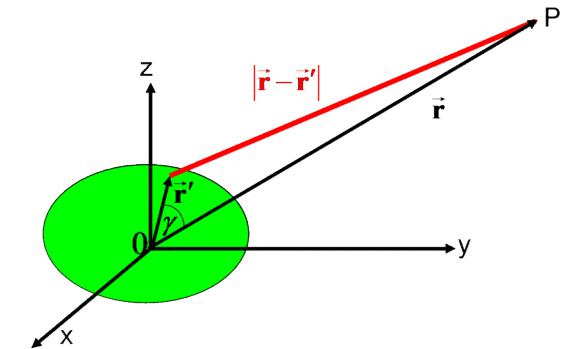
$$e^{j\beta r' \cos \gamma} \approx 1 \rightarrow \frac{2\pi}{\lambda} r' \ll 2\pi \rightarrow r' \ll \lambda$$



Extended antennas

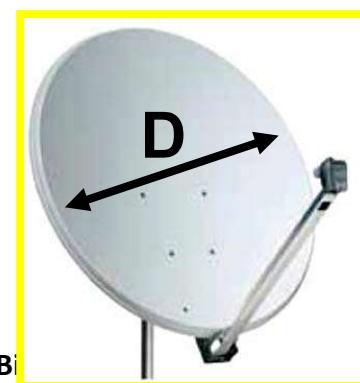
$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } r' \ll r$$

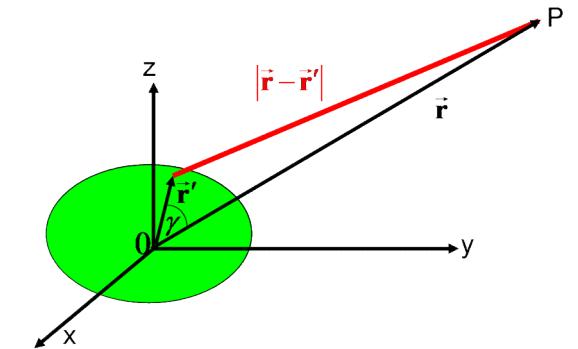
$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} \quad \text{if } r' \ll \lambda$$



Extended antennas

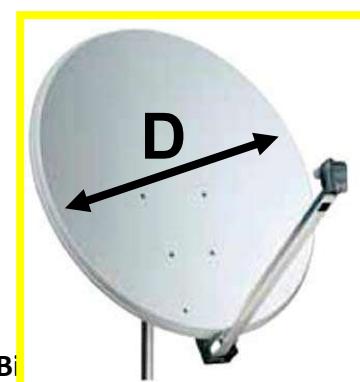
$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

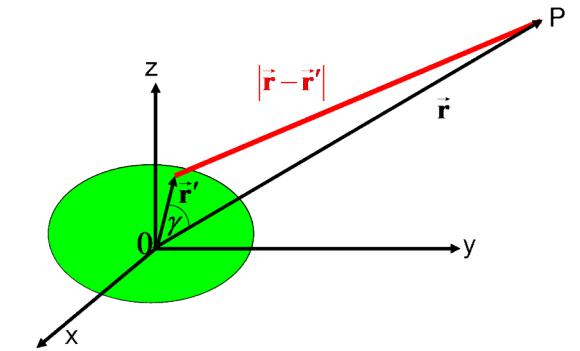
$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} \quad \text{if } D \ll \lambda$$



Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cancel{\cos \gamma} + \frac{(r')^2}{2r} \cancel{\sin^2 \gamma} + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cancel{\cos \gamma} + \frac{(r')^2}{2r} \cancel{\sin^2 \gamma} + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{\cancel{j\beta r'} \cancel{\cos \gamma}} e^{\cancel{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma}} \dots \approx e^{-j\beta r} \quad \text{if } D \ll \lambda$$

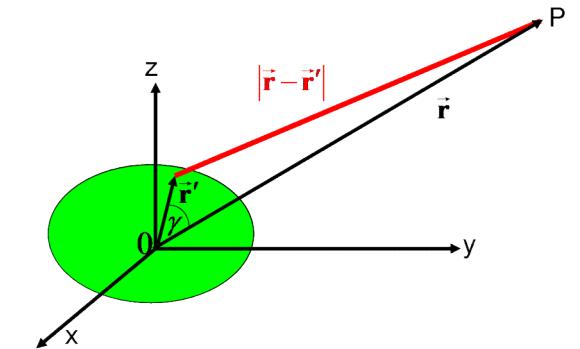
When the antennas are small with respect to the wavelength and to the distance from the observation point

$$\frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} \approx \frac{e^{-j\beta r}}{r} \quad \rightarrow \quad \mathbf{A}(\vec{r}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \int \mathbf{J}(\vec{r}') d\vec{r}'$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



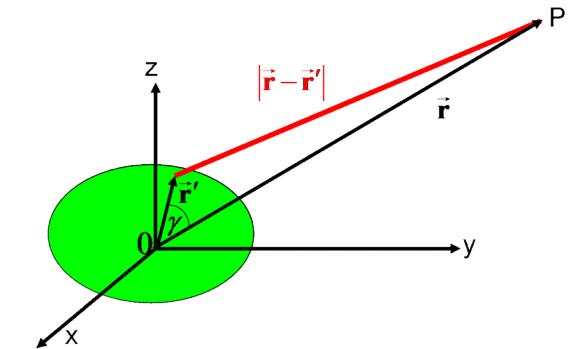
$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{\cancel{j\beta r' \cos \gamma}} e^{\cancel{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma}} \dots \approx e^{-j\beta r} \quad \text{if } D \ll \lambda$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

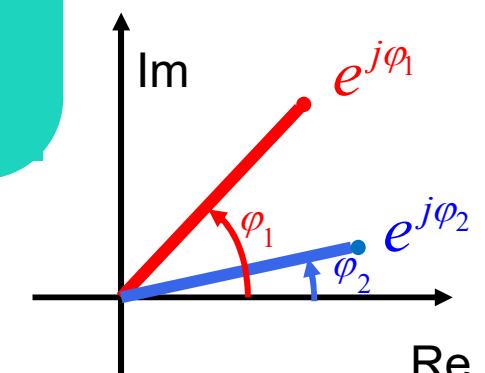
$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta r' \cos \gamma}$$

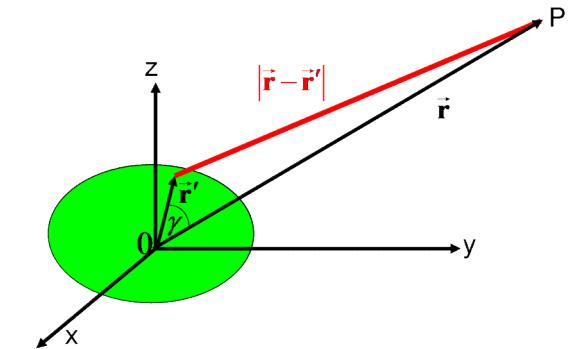
$$e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \approx 1 \rightarrow \beta \frac{(r')^2}{2r} \sin^2 \gamma \ll 2\pi \rightarrow \frac{2\pi}{\lambda} \left(\frac{D}{2} \right)^2 \frac{1}{2r} < \frac{\pi}{8} \rightarrow r > \frac{2D^2}{\lambda}$$



Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

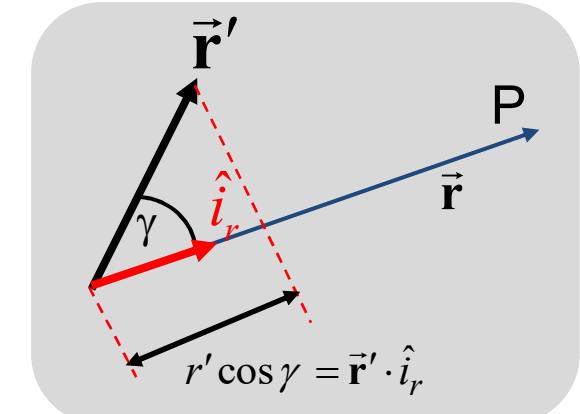
$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta r' \cos \gamma} \quad \text{if } r > \frac{2D^2}{\lambda}$$

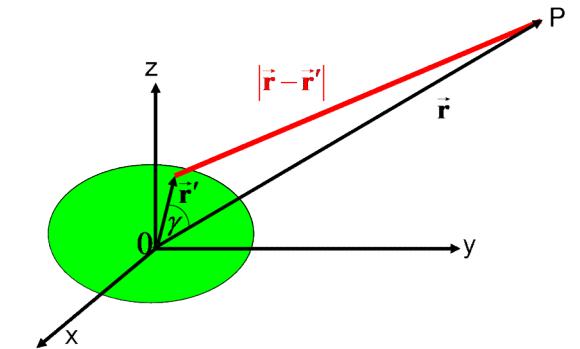
$$e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \approx 1 \quad \Rightarrow \quad \beta \frac{(r')^2}{2r} \sin^2 \gamma \ll 2\pi \quad \Rightarrow \quad \frac{2\pi}{\lambda} \left(\frac{D}{2} \right)^2 \frac{1}{2r} < \frac{\pi}{8} \quad \Rightarrow \quad r > \frac{2D^2}{\lambda}$$



Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

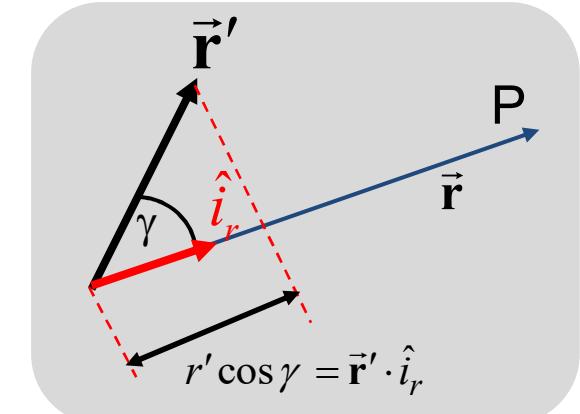
$$|\vec{r}-\vec{r}'| = r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta \vec{r} \cdot \hat{i}_r} \quad \text{if } r > \frac{2D^2}{\lambda}$$

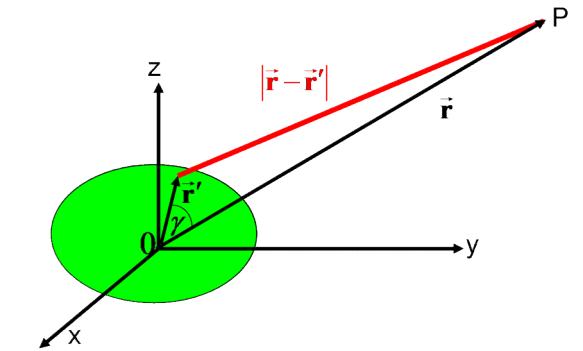
$$e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \approx 1 \rightarrow \beta \frac{(r')^2}{2r} \sin^2 \gamma \ll 2\pi \rightarrow \frac{2\pi}{\lambda} \left(\frac{D}{2} \right)^2 \frac{1}{2r} < \frac{\pi}{8} \rightarrow r > \frac{2D^2}{\lambda}$$



Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

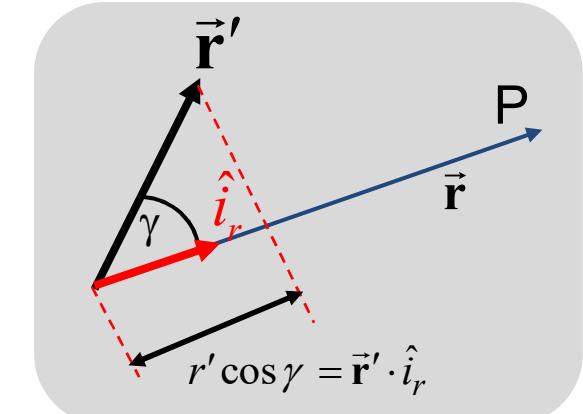
$$|\vec{r}-\vec{r}'| = r - \vec{r}' \cdot \hat{i}_r + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta \vec{r}' \cdot \hat{i}_r} \quad \text{if } r > \frac{2D^2}{\lambda}$$

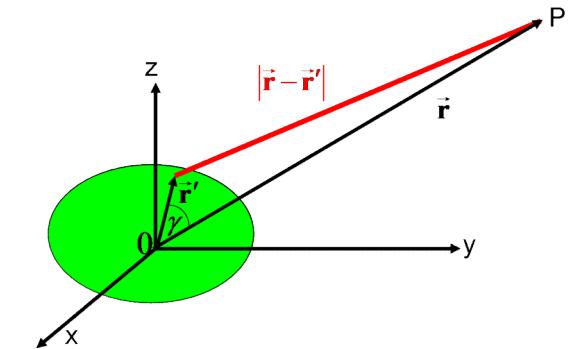
$$e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \approx 1 \rightarrow \beta \frac{(r')^2}{2r} \sin^2 \gamma \ll 2\pi \rightarrow \frac{2\pi}{\lambda} \left(\frac{D}{2} \right)^2 \frac{1}{2r} < \frac{\pi}{8} \rightarrow r > \frac{2D^2}{\lambda}$$



Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - \vec{r}' \cdot \hat{i}_r + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta \vec{r}' \cdot \hat{i}_r} \quad \text{if } r > \frac{2D^2}{\lambda}$$

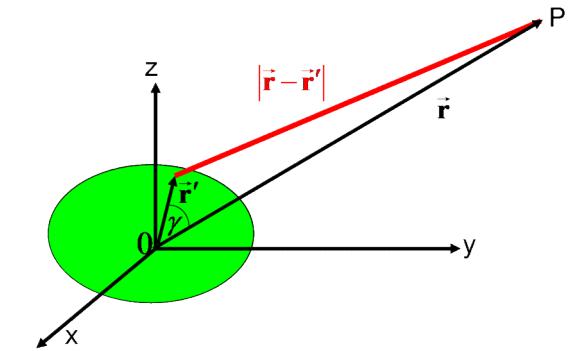
For all the antennas, if the distance from the observation point is sufficiently large

$$\frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} \approx \frac{e^{-j\beta r} e^{j\beta \vec{r}' \cdot \hat{i}_r}}{r} \quad \rightarrow \quad \mathbf{A}(\vec{r}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \int \mathbf{J}(\vec{r}') e^{j\beta \vec{r}' \cdot \hat{i}_r} d\vec{r}'$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

$$|\vec{r}-\vec{r}'| = r - \vec{r}' \cdot \hat{i}_r + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



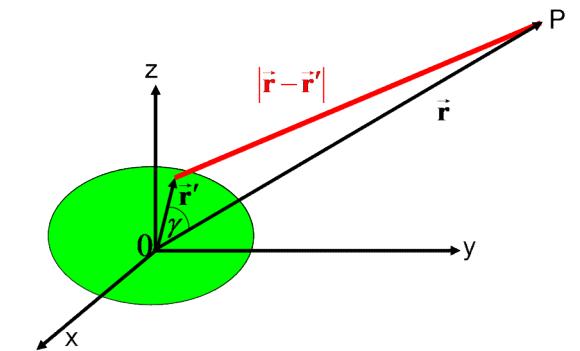
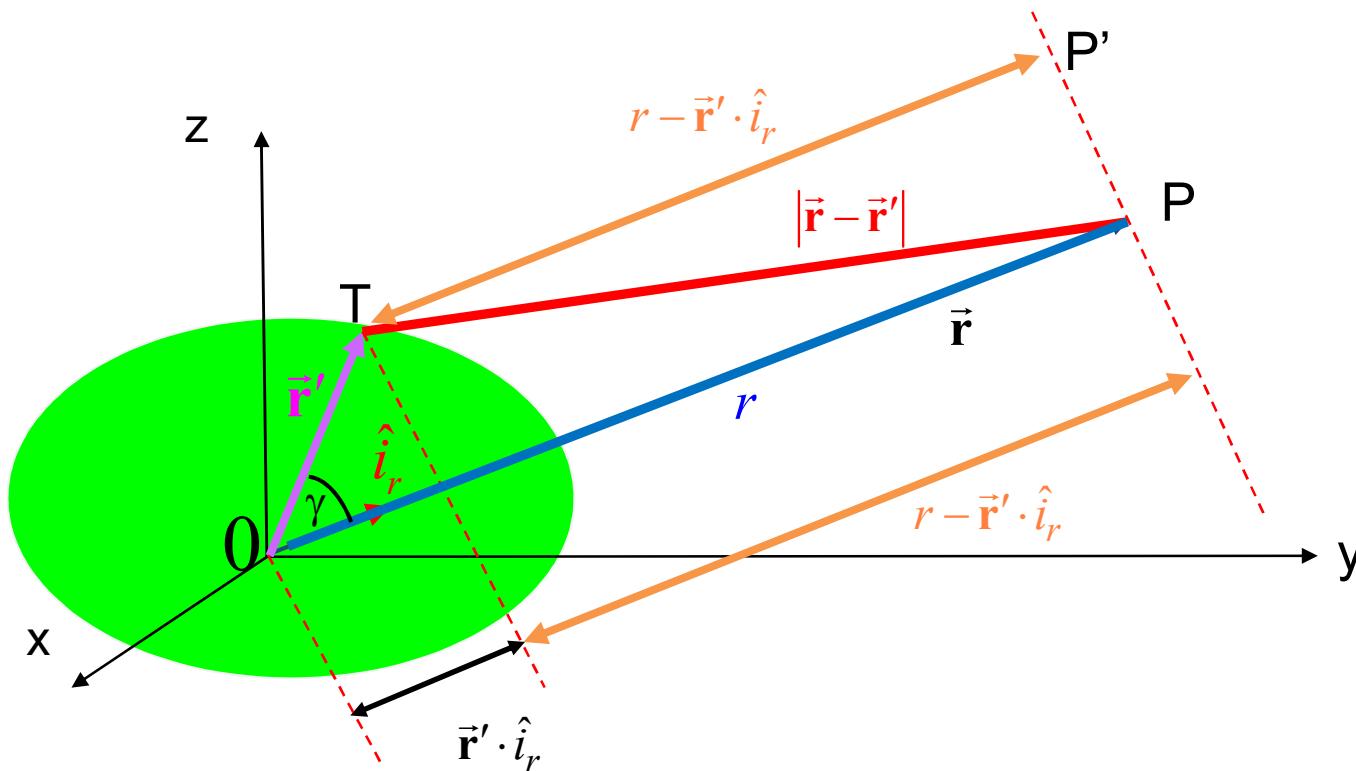
$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta \vec{r}' \cdot \hat{i}_r} \quad \text{if } r > \frac{2D^2}{\lambda}$$

Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

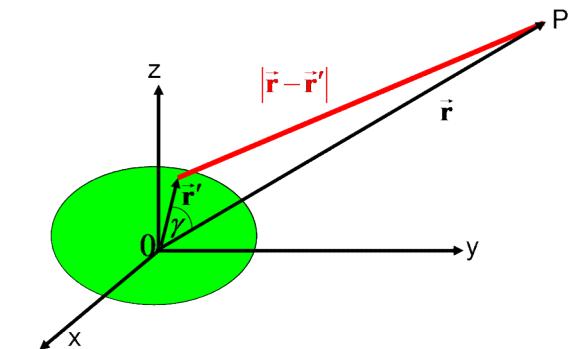
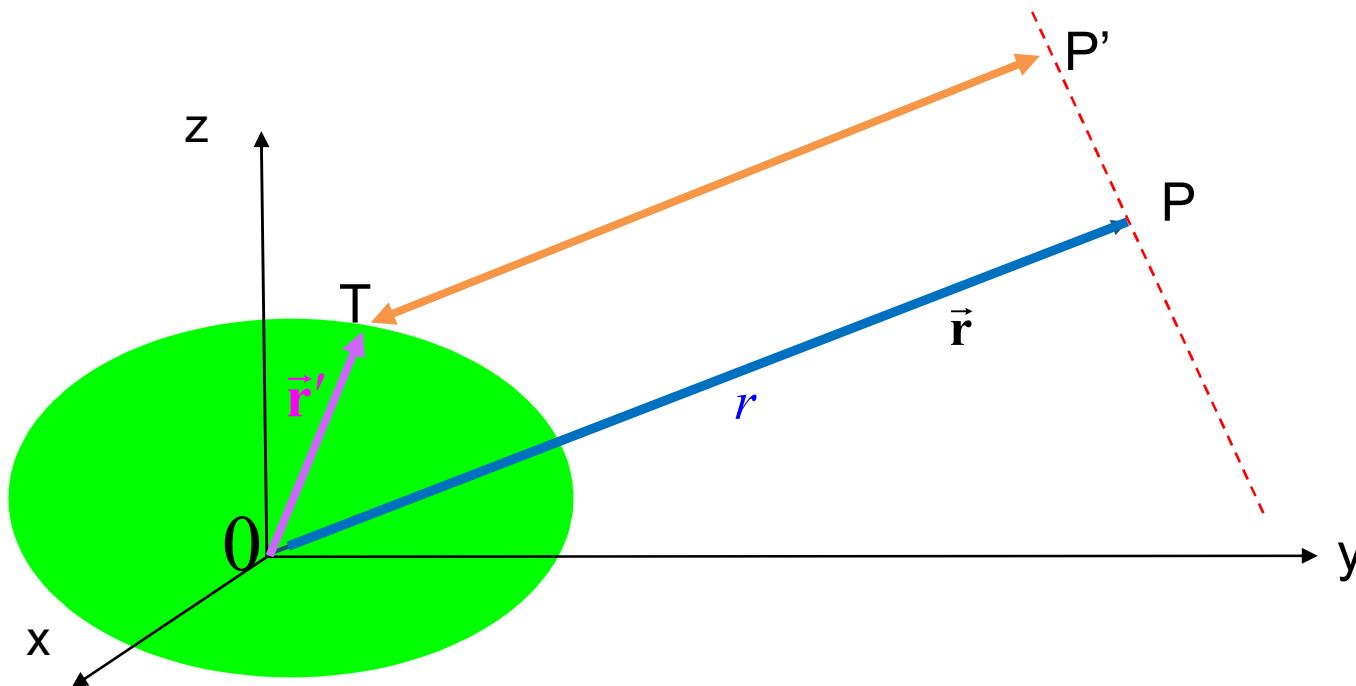
$$|\vec{r}-\vec{r}'| = r - \vec{r}' \cdot \hat{i}_r + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



Extended antennas

$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$

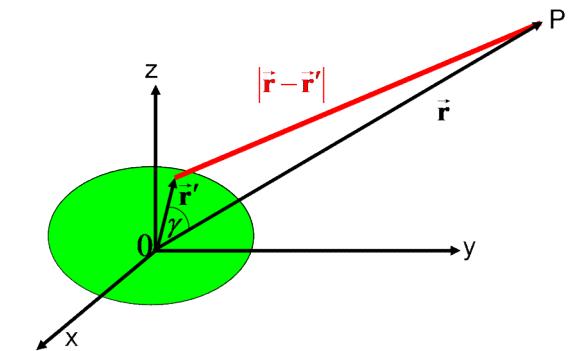
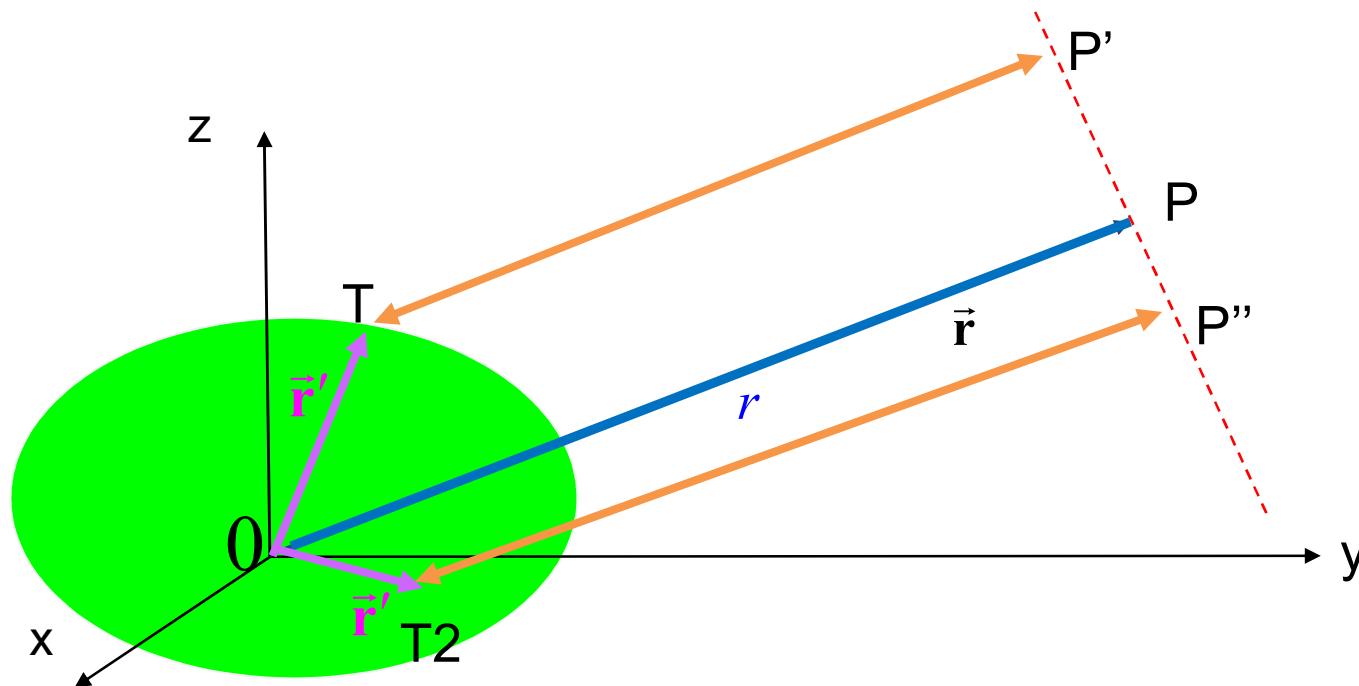
$$|\vec{r}-\vec{r}'| = r - \vec{r}' \cdot \hat{i}_r + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



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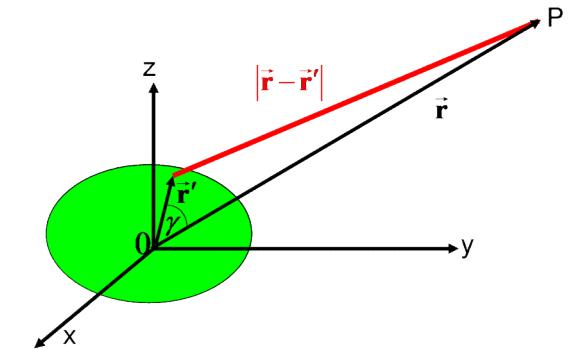
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$$|\vec{r}-\vec{r}'| = r - \vec{r}' \cdot \hat{i}_r + \frac{(r')^2}{2r} \sin^2 \gamma + \dots$$



$$\frac{1}{|\vec{r}-\vec{r}'|} = \frac{1}{r - r' \cos \gamma + \frac{(r')^2}{2r} \sin^2 \gamma + \dots} \approx \frac{1}{r} \quad \text{if } D \ll r$$

$$e^{-j\beta|\vec{r}-\vec{r}'|} = e^{-j\beta r} e^{j\beta r' \cos \gamma} e^{-j\beta \frac{(r')^2}{2r} \sin^2 \gamma} \dots \approx e^{-j\beta r} e^{j\beta \vec{r}' \cdot \hat{i}_r} \quad \text{if } r > \frac{2D^2}{\lambda}$$

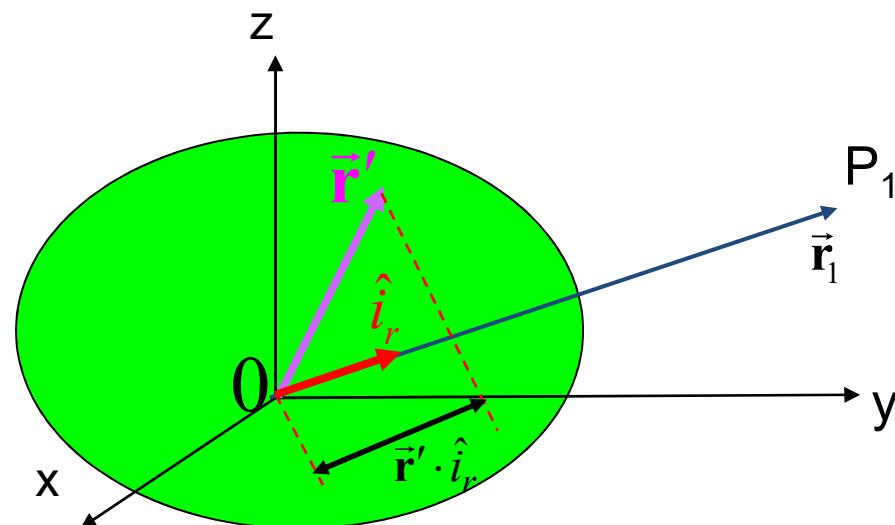
For all the antennas, if the distance from the observation point is sufficiently large

$$\frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} \approx \frac{e^{-j\beta r} e^{j\beta \vec{r}' \cdot \hat{i}_r}}{r} \quad \rightarrow \quad \mathbf{A}(\vec{r}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \int \mathbf{J}(\vec{r}') e^{j\beta \vec{r}' \cdot \hat{i}_r} d\vec{r}'$$

Extended antennas

$$r \gg D$$
$$r > \frac{2D^2}{\lambda}$$

$$\int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{\mathbf{i}}_r} d\vec{\mathbf{r}}' =$$



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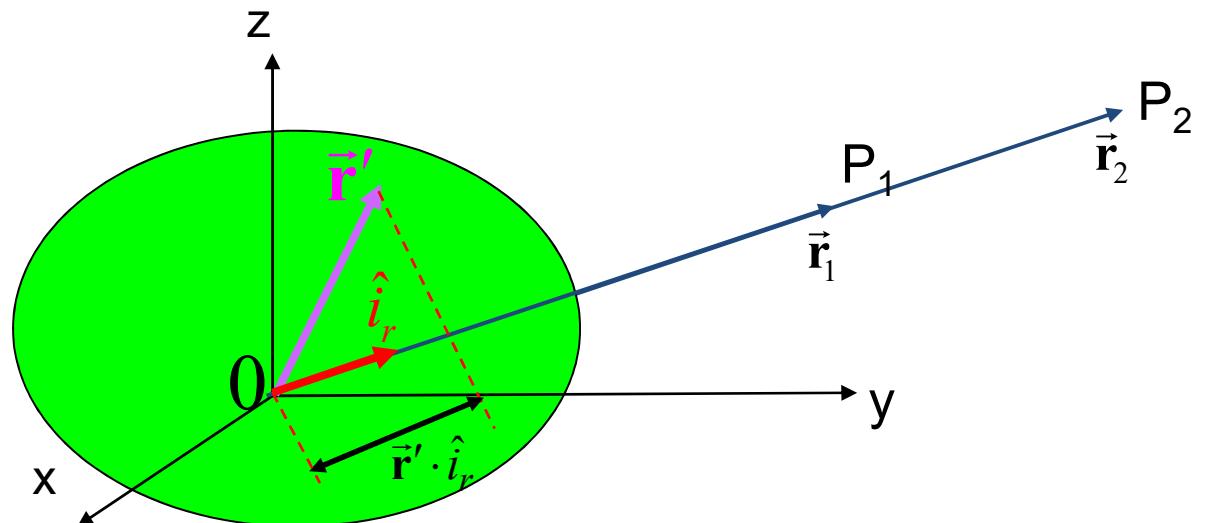
$$\frac{e^{-j\beta|\vec{\mathbf{r}}-\vec{\mathbf{r}}'|}}{|\vec{\mathbf{r}}-\vec{\mathbf{r}}'|} \approx \frac{e^{-j\beta r} e^{j\beta \vec{\mathbf{r}}' \cdot \hat{\mathbf{i}}_r}}{r} \quad \rightarrow \quad \mathbf{A}(\vec{\mathbf{r}}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{\mathbf{i}}_r} d\vec{\mathbf{r}}'$$

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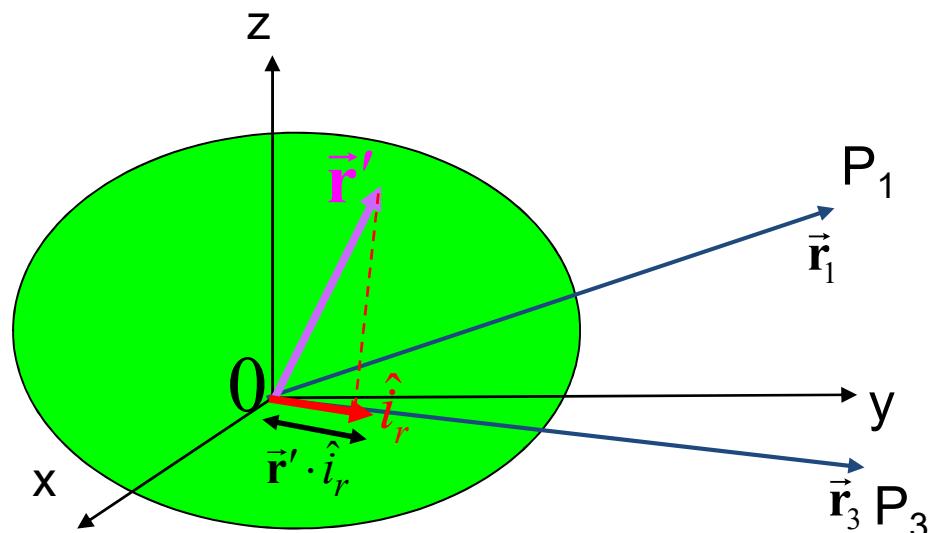
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Extended antennas

$$r \gg D$$

$$r > \frac{2D^2}{\lambda}$$

$$\int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{\mathbf{i}}_r} d\vec{\mathbf{r}}' = \mathbf{M}(\vartheta, \varphi) \rightarrow \mathbf{A}(\vec{\mathbf{r}}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \mathbf{M}(\vartheta, \varphi)$$



For all the antennas, if the distance from the observation point is sufficiently large

$$\frac{e^{-j\beta|\vec{\mathbf{r}}-\vec{\mathbf{r}}'|}}{|\vec{\mathbf{r}}-\vec{\mathbf{r}}'|} \approx \frac{e^{-j\beta r} e^{j\beta \vec{\mathbf{r}}' \cdot \hat{\mathbf{i}}_r}}{r}$$

$$\mathbf{A}(\vec{\mathbf{r}}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{\mathbf{i}}_r} d\vec{\mathbf{r}}'$$

Extended antennas

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Fraunhofer region

$r \gg D$

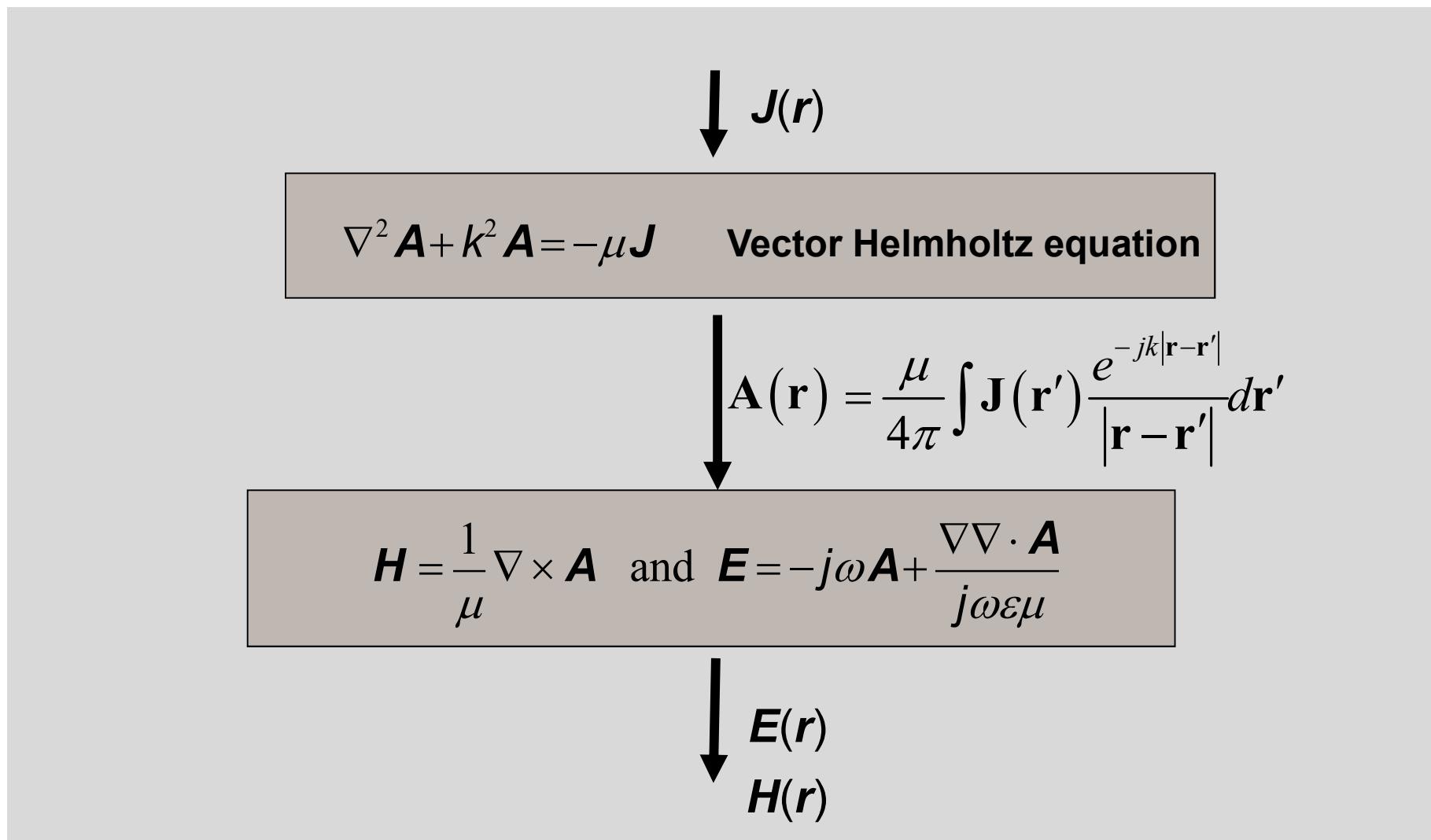
$r > \frac{2D^2}{\lambda}$

$r \gg \lambda$

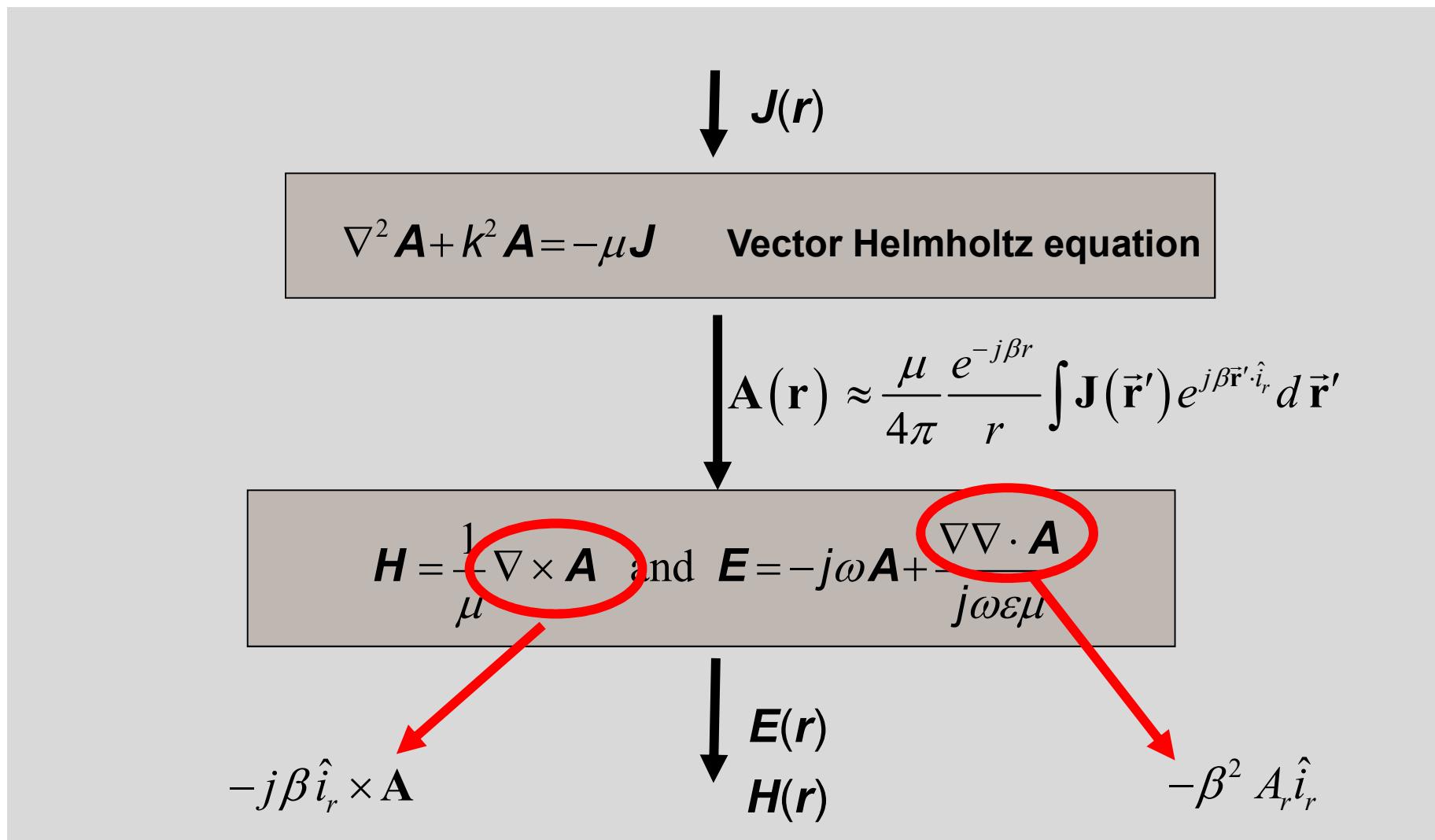
$$\int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{i}_r} d\vec{\mathbf{r}}' = \mathbf{M}(\vartheta, \varphi) \implies \mathbf{A}(\vec{\mathbf{r}}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \mathbf{M}(\vartheta, \varphi)$$
$$\left\{ \begin{array}{l} \nabla \nabla \cdot \mathbf{A}(\vec{\mathbf{r}}) \approx -\beta^2 A_r(\vec{\mathbf{r}}) \hat{i}_r \\ \nabla \times \mathbf{A}(\vec{\mathbf{r}}) \approx -j\beta \hat{i}_r \times \mathbf{A}(\vec{\mathbf{r}}) \end{array} \right.$$

Fraunhofer region

Radiation problem for extended antennas



Radiation problem for extended antennas



Fraunhofer region

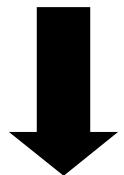
$r \gg D$
 $r > \frac{2D^2}{\lambda}$
 $r \gg \lambda$



$\int \mathbf{J}(\vec{r}') e^{j\beta \vec{r}' \cdot \hat{i}_r} d\vec{r}' = \mathbf{M}(\vartheta, \varphi) \implies \boxed{\mathbf{A}(\vec{r}) \approx \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \mathbf{M}(\vartheta, \varphi)}$

$$\left\{ \begin{array}{l} \nabla \nabla \cdot \mathbf{A}(\vec{r}) \approx -\beta^2 A_r(\vec{r}) \hat{i}_r \\ \nabla \times \mathbf{A}(\vec{r}) \approx -j\beta \hat{i}_r \times \mathbf{A}(\vec{r}) \end{array} \right.$$

Fraunhofer region



$\mathbf{A}(\vec{r}) = A_r(\vec{r}) \hat{i}_r + A_\vartheta(\vec{r}) \hat{i}_\vartheta + A_\varphi(\vec{r}) \hat{i}_\varphi$

$\mathbf{E}(\vec{r}) = -j\omega [\mathbf{A}(\vec{r}) - A_r(\vec{r}) \hat{i}_r] = -j\omega \left[\frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} \mathbf{M}(\vartheta, \varphi) - \frac{\mu}{4\pi} \frac{e^{-j\beta r}}{r} M_r(\vartheta, \varphi) \hat{i}_r \right] = -\frac{j\omega \mu}{4\pi} \frac{e^{-j\beta r}}{r} [\mathbf{M}(\vartheta, \varphi) - M_r(\vartheta, \varphi) \hat{i}_r]$

$\zeta \mathbf{H}(\vec{r}) = \hat{i}_r \times \mathbf{E}(\vec{r})$

Fraunhofer region

$$r \gg D$$

$$r > \frac{2D^2}{\lambda}$$

$$r \gg \lambda$$

$$\int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{i}_r} d\vec{\mathbf{r}}' = \mathbf{M}(\vartheta, \varphi)$$

Fraunhofer region

$$\mathbf{E}(\vec{\mathbf{r}}) =$$

$$= -\frac{j\omega\mu}{4\pi} \frac{e^{-j\beta r}}{r} \left[\mathbf{M}(\vartheta, \varphi) - M_r(\vartheta, \varphi) \hat{i}_r \right]$$

$$\zeta \mathbf{H}(\vec{\mathbf{r}}) = \hat{i}_r \times \mathbf{E}(\vec{\mathbf{r}})$$

Fraunhofer region

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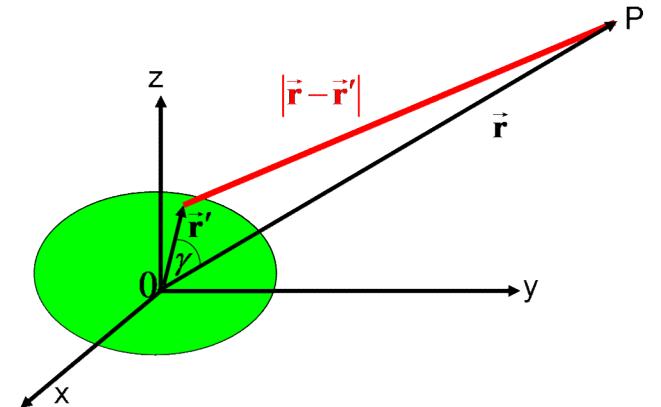
Fraunhofer region

$$\mathbf{E}(\vec{\mathbf{r}}) = \mathbf{E}(r, \vartheta, \varphi) = -\frac{j\omega\mu}{4\pi} \frac{e^{-j\beta r}}{r} [\mathbf{M}(\vartheta, \varphi) - M_r(\vartheta, \varphi) \hat{i}_r]$$

$$\zeta \mathbf{H}(\vec{\mathbf{r}}) = \hat{i}_r \times \mathbf{E}(\vec{\mathbf{r}})$$

- the e.m. field propagates along \hat{i}_r
- the e.m. field lies on the plane orthogonal to the propagation direction
- $|E|$ and $|H|$ exhibit the decaying factor $1/r$
- $|E|$ and $|H|$ are proportional through ζ

$$\mathbf{M}(\vartheta, \varphi) = \int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{i}_r} d\vec{\mathbf{r}'}$$



Field regions

Far-field (Fraunhofer) region is defined as “that region of the field of an antenna where the angular field distribution is essentially independent of the distance from the antenna. If the antenna has a maximum overall dimension D ($D>\lambda$), the far-field region is commonly taken to exist at distances greater than $2D^2/\lambda$ from the antenna, λ being the wavelength”.

In this region, the field components are essentially transverse

Fraunhofer region

$$r \gg D$$

$$r > \frac{2D^2}{\lambda}$$

$$r \gg \lambda$$

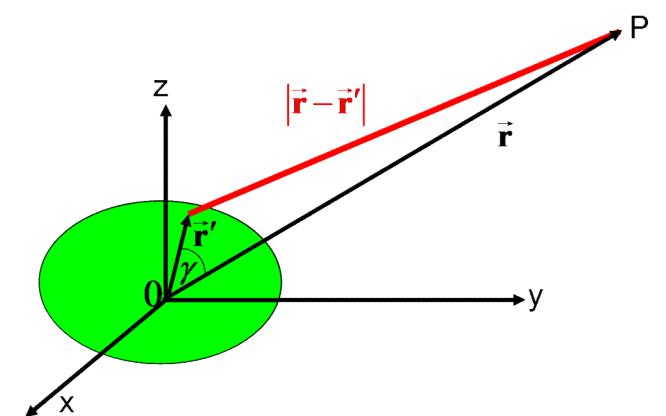
Fraunhofer region

$$\mathbf{E}(\vec{\mathbf{r}}) = \mathbf{E}(r, \vartheta, \phi) = -\frac{j\omega\mu}{4\pi} \frac{e^{-j\beta r}}{r} [\mathbf{M}(\vartheta, \phi) - M_r(\vartheta, \phi) \hat{i}_r]$$

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$$\mathbf{M}(\vartheta, \phi) = \int \mathbf{J}(\vec{\mathbf{r}}') e^{j\beta \vec{\mathbf{r}}' \cdot \hat{i}_r} d\vec{\mathbf{r}'}$$



The radiation condition

$$r \gg D$$

$$r > \frac{2D^2}{\lambda}$$

$$r \gg \lambda$$

Fraunhofer region

$$\mathbf{E} \sim O\left(\frac{1}{r}\right)$$

$$\mathbf{H} \sim O\left(\frac{1}{r}\right)$$

$$\zeta \mathbf{H} - \hat{i}_r \times \mathbf{E} \sim o\left(\frac{1}{r}\right)$$

as $r \rightarrow \infty$

$$\mathbf{E}(\vec{\mathbf{r}}) = \mathbf{E}(r, \vartheta, \phi) = -\frac{j\omega\mu}{4\pi} \frac{e^{-j\beta r}}{r} \left[\mathbf{M}(\vartheta, \phi) - M_r(\vartheta, \phi) \hat{i}_r \right]$$

$$\zeta \mathbf{H}(\vec{\mathbf{r}}) = \hat{i}_r \times \mathbf{E}(\vec{\mathbf{r}})$$

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