

Campi Elettromagnetici

**Corso di Laurea in Ingegneria Informatica,
Biomedica e delle Telecomunicazioni**

a.a. 2019-2020 - Laurea "Triennale" – Secondo semestre - Secondo anno

Università degli Studi di Napoli "Parthenope"

Stefano Perna

Color legend

New formulas, important considerations,
important formulas, important concepts

Very important for the discussion

Memo

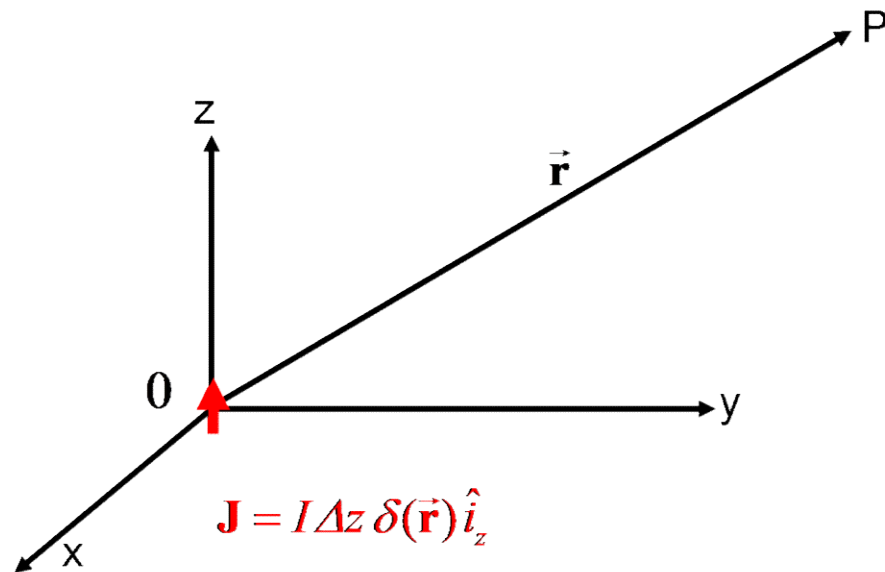
Mathematical tools to be exploited

Mathematics

Elementary electrical dipole

- A δ -source radiating element is also known as elementary electrical dipole.

$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z = I \Delta z \delta(x) \delta(y) \delta(z) \hat{i}_z$$

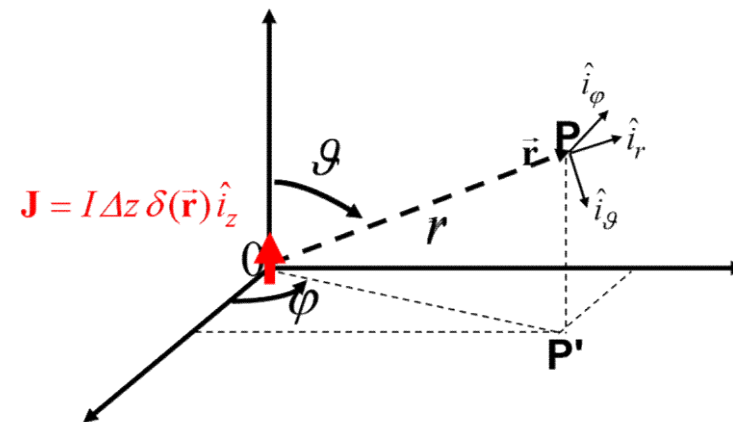


- Why are we interested in such a radiating element?
- Why is such a radiating element referred to as elementary electrical dipole?
- How can we physically approximate an elementary electrical dipole?

Elementary electrical dipole: far field

The E.M. field radiated by the elementary electrical dipole

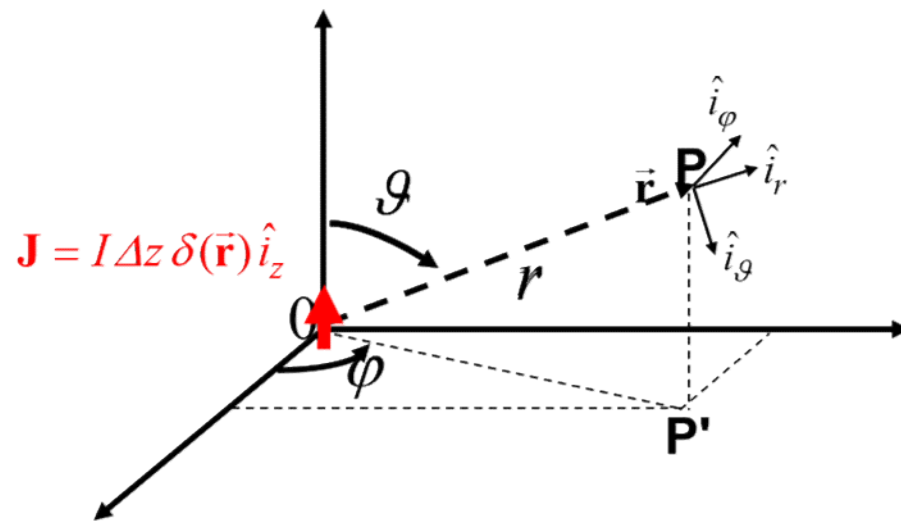
$$\begin{cases} E_r = \zeta \frac{I\Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I\Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I\Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases} \quad \begin{aligned} \vec{\mathbf{E}}(\vec{\mathbf{r}}) &= E_r(r, \vartheta) \hat{i}_r + E_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{H}}(\vec{\mathbf{r}}) &= H_\varphi(r, \vartheta) \hat{i}_\varphi \end{aligned}$$



Elementary electrical dipole: far field

In the far-field case ($r \gg \lambda$) the elementary electrical dipole behaves as follows

$$\begin{aligned} \vec{\mathbf{E}}(\vec{\mathbf{r}}) &= E_{\vartheta}(r, \vartheta) \hat{i}_{\vartheta} \\ \vec{\mathbf{H}}(\vec{\mathbf{r}}) &= H_{\varphi}(r, \vartheta) \hat{i}_{\varphi} \end{aligned} \quad \left\{ \begin{aligned} E_{\vartheta} &= j\zeta \frac{I\Delta z}{2\lambda r} \sin \vartheta \exp(-j\beta r) \\ H_{\varphi} &= \frac{E_{\vartheta}}{\zeta} \end{aligned} \right.$$



- the e.m. field propagates along \hat{i}_r
- the e.m. field lies on the plane orthogonal to the propagation direction
- $|E|$ and $|H|$ exhibit the decaying factor $1/r$
- $|E|$ and $|H|$ are proportional through ζ

$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$

$$\vec{\mathbf{S}} = \frac{1}{2\zeta} |\vec{\mathbf{E}}|^2 \hat{i}_r = \frac{\zeta}{2} |\vec{\mathbf{H}}|^2 \hat{i}_r$$

Elementary electrical dipole: power flux

$$\begin{cases} E_r = \zeta \frac{I \Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

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$$P = \frac{1}{2} \oiint_S [\mathbf{E} \times \mathbf{H}^*] \cdot \hat{i}_r dS = \frac{1}{2} \int_0^{2\pi} d\varphi \int_0^\pi d\vartheta r^2 \sin \vartheta E_\vartheta H_\varphi^*$$

$$dS = r^2 \sin \vartheta d\vartheta d\varphi$$

$$[\mathbf{E} \times \mathbf{H}^*] \cdot \hat{i}_r = \left[(E_\vartheta \hat{i}_\vartheta + E_r \hat{i}_r) \times (H_\varphi^* \hat{i}_\varphi) \right] \cdot \hat{i}_r = E_\vartheta H_\varphi^* \hat{i}_r \cdot \hat{i}_r = E_\vartheta H_\varphi^*$$

$$\begin{aligned} \hat{i}_\varphi &= \hat{i}_r \times \hat{i}_\vartheta \\ \hat{i}_\vartheta &= \hat{i}_\varphi \times \hat{i}_r \\ \hat{i}_r &= \hat{i}_\vartheta \times \hat{i}_\varphi \end{aligned}$$

Elementary electrical dipole: power flux

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$$\begin{aligned} P &= \frac{1}{2} \oiint_S [\mathbf{E} \times \mathbf{H}^*] \cdot \hat{i}_r dS = \frac{1}{2} \int_0^{2\pi} d\varphi \int_0^\pi d\vartheta r^2 \sin \vartheta E_\vartheta H_\varphi^* \\ &= \frac{1}{2} \left[\zeta \frac{|I|^2 \Delta z^2}{(4\pi)^2} \right] \int_0^{2\pi} d\varphi \int_0^\pi d\vartheta r^2 \sin \vartheta \left[\sin^2 \vartheta \right] \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \left(\frac{j\beta}{r} + \frac{1}{r^2} \right)^* \end{aligned}$$

Elementary electrical dipole: power flux

$$\begin{cases} E_r = \zeta \frac{I \Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

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$$\int_0^{2\pi} d\varphi = 2\pi$$

$$\int_0^\pi d\vartheta \sin^3 \vartheta = \frac{4}{3}$$

Elementary electrical dipole: power flux

$$\begin{cases} E_r = \zeta \frac{I \Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

$$\begin{aligned} P &= \frac{1}{2} \oiint_S [\mathbf{E} \times \mathbf{H}^*] \cdot \hat{i}_r dS = \frac{1}{2} \int_0^{2\pi} d\varphi \int_0^\pi d\vartheta r^2 \sin \vartheta E_\vartheta H_\varphi^* = \frac{1}{2} \left[\zeta \frac{|I|^2 \Delta z^2}{(4\pi)^2} \right] r^2 \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \left(\frac{j\beta}{r} + \frac{1}{r^2} \right)^* 2\pi \frac{4}{3} \\ &= \frac{1}{2} \left[\zeta \frac{|I|^2 \Delta z^2}{(4\pi)^2} \right] r^2 \frac{\beta^2}{r^2} \left[1 - j \frac{1}{(\beta r)^3} \right] 2\pi \frac{4}{3} = \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\Delta z}{\lambda} \right)^2 \left[1 - j \frac{1}{(\beta r)^3} \right] |I|^2 \end{aligned}$$

$$\left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \left(\frac{-j\beta}{r} + \frac{1}{r^2} \right) = \left(\frac{\beta}{r} \right)^2 + \cancel{\frac{j\beta}{r^3}} - \cancel{\frac{j\beta}{r^3}} + \cancel{\frac{1}{r^4}} - \cancel{\frac{1}{r^4}} - j \frac{1}{\beta r^5} = \left(\frac{\beta}{r} \right)^2 - j \frac{1}{\beta r^5} = \frac{\beta^2}{r^2} \left[1 - j \frac{1}{(\beta r)^3} \right]$$

Elementary electrical dipole: power flux

$$\begin{cases} E_r = \zeta \frac{I \Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

$$P = P_1 + jP_2$$

$$P_1 = \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\Delta z}{\lambda} \right)^2 |I|^2$$

$$P_2 = -\frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\Delta z}{\lambda} \right)^2 \frac{1}{(\beta r)^3} |I|^2$$

$$P = \frac{1}{2} \oiint_S [\mathbf{E} \times \mathbf{H}^*] \cdot \hat{i}_r dS$$

$$= \frac{1}{2} \frac{2\pi}{3} \zeta \left(\frac{\Delta z}{\lambda} \right)^2 \left[1 - j \frac{1}{(\beta r)^3} \right] |I|^2$$

Elementary electrical dipole: power flux

$$\begin{cases} E_r = \zeta \frac{I \Delta z}{2\pi} \left(\frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I \Delta z}{4\pi} \left(\frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases}$$

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- Note that in the far-field case only the first active power term exists and it does not depend on r
- Note that the real part of the power, in lossless medium, is independent of r , therefore if one consider two different spherical surfaces one gets the same result. Only the so-called radiative terms contribute.
- The reactive part depends on r . Its sign is negative showing that there is an excess of stored **electric** energy in the neighbor of the electrical dipole (see Poynting's theorem)

$$\int_0^{\pi} d\vartheta \sin^3 \vartheta = \frac{4}{3}$$

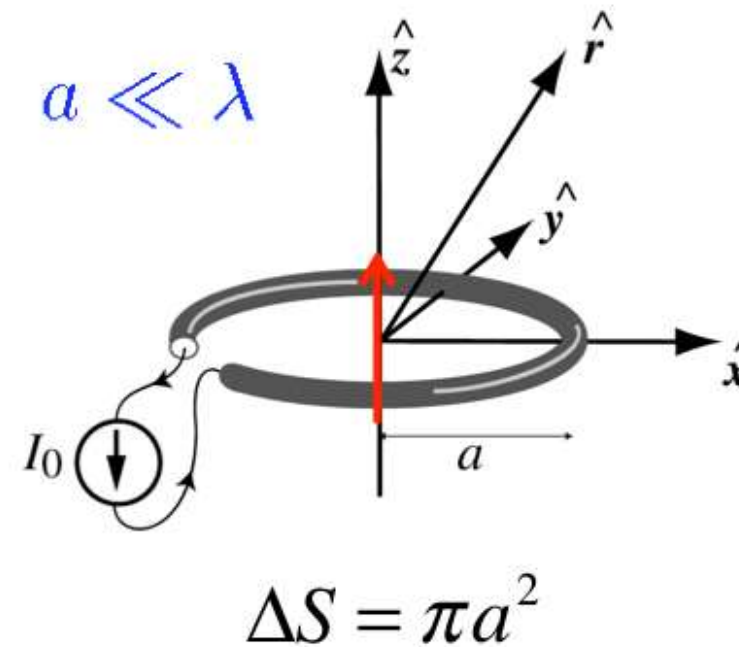
$$\int_0^{\pi} d\vartheta \sin^3 \vartheta = \int_0^{\pi} d\vartheta \sin \vartheta (1 - \cos^2 \vartheta) = \int_{-1}^1 dx (1 - x^2) = \left[x - \frac{1}{3} x^3 \right]_{-1}^1 = \frac{4}{3}$$

$$x = \cos \vartheta$$

$$dx = -d\vartheta \sin \vartheta$$

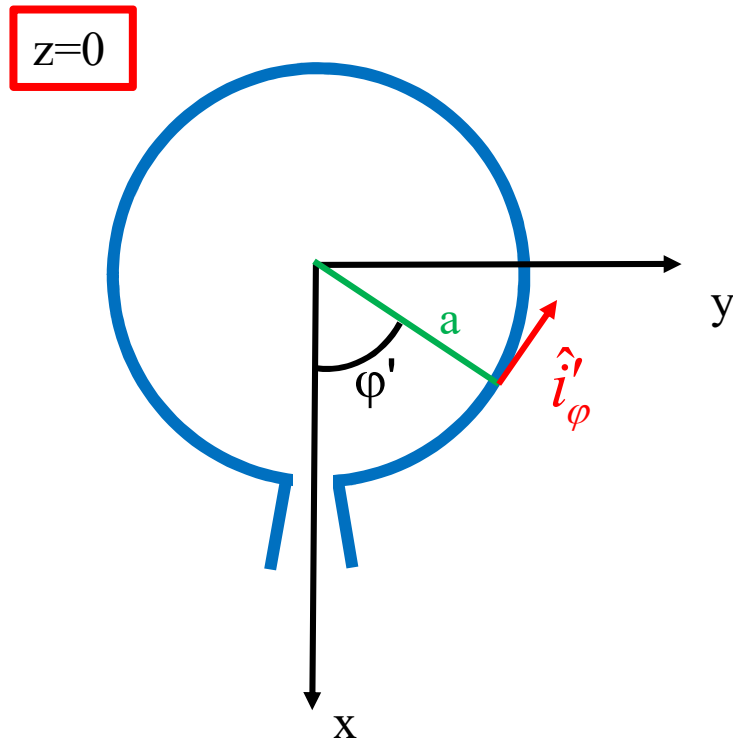
Small loop antenna

A simple and inexpensive antenna type is the loop antenna



Small loop antenna

$$\mathbf{J} = I\delta(z)\delta(r-a)\hat{i}'_{\varphi}$$



↓ \mathbf{J}

$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J} \quad \text{Vector Helmholtz equation}$$

↓

$$\mathbf{A}(\mathbf{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\mathbf{r}') \frac{e^{-jk|\mathbf{r}-\mathbf{r}'|}}{|\mathbf{r}-\mathbf{r}'|} d\mathbf{r}'$$

$$\mathbf{H} = \frac{1}{\mu} \nabla \times \mathbf{A} \quad \text{and} \quad \mathbf{E} = -j\omega \mathbf{A} + \frac{\nabla \nabla \cdot \mathbf{A}}{j\omega \epsilon \mu}$$

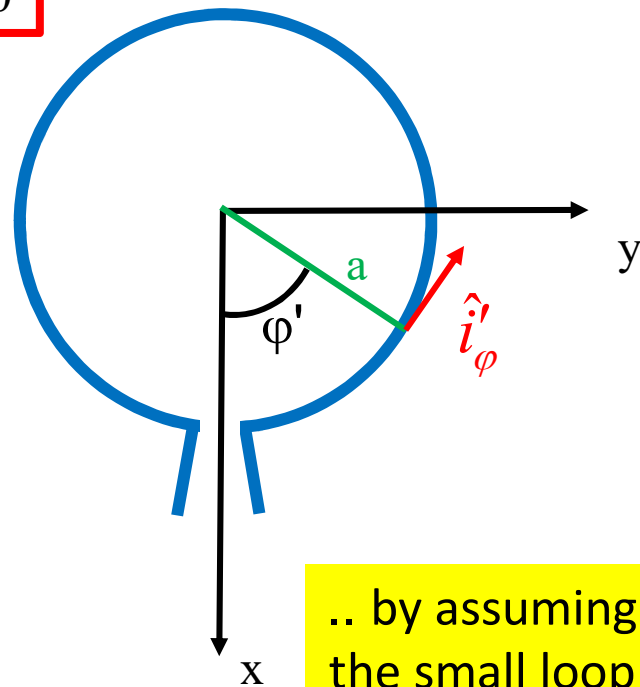
↓

$\mathbf{E}(\mathbf{r})$
 $\mathbf{H}(\mathbf{r})$

Small loop antenna

$$\mathbf{J} = I\delta(z)\delta(r-a)\hat{i}'_{\varphi}$$

$z=0$



.. by assuming that the current I in the small loop is constant and that the radius of the loop $a \ll \lambda$

$\mathbf{J} \downarrow$

$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J} \quad \text{Vector Helmholtz equation}$$

$$\mathbf{A} \approx \frac{j\beta I \mu \Delta S}{4\pi} \frac{e^{-j\beta r}}{r} \left[1 + \frac{1}{j\beta r} \right] \sin \vartheta \hat{i}'_{\varphi}$$

$$\mathbf{H} = \frac{1}{\mu} \nabla \times \mathbf{A} \quad \text{and} \quad \mathbf{E} = -j\omega \mathbf{A} + \frac{\nabla \nabla \cdot \mathbf{A}}{j\omega \epsilon \mu}$$

$\mathbf{E} \downarrow \mathbf{H}$

Small loop antenna

The E.M. field radiated by the the small loop antenna

$$\begin{aligned} \vec{\mathbf{H}}(\vec{\mathbf{r}}) &= H_r(r, \vartheta) \hat{i}_r + H_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{E}}(\vec{\mathbf{r}}) &= E_\varphi(r, \vartheta) \hat{i}_\varphi \end{aligned} \quad \left\{ \begin{aligned} H_r &= \frac{I \Delta S (j\beta)^2}{2\pi r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta &= \frac{I \Delta S (j\beta)^2}{4\pi r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\varphi &= -\frac{\zeta I \Delta S (j\beta)^2}{4\pi r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{aligned} \right.$$

Because of the problem symmetry there is no dependence on the azimuth angle φ .

Small loop antenna: far field

The E.M. field radiated by the the small loop antenna

$$\begin{aligned} \vec{\mathbf{H}}(\vec{\mathbf{r}}) &= H_r(r, \vartheta) \hat{i}_r + H_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{E}}(\vec{\mathbf{r}}) &= E_\varphi(r, \vartheta) \hat{i}_\varphi \end{aligned} \quad \left\{ \begin{aligned} H_r &= \frac{I \Delta S (j\beta)^2}{2\pi r} \left(\frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ H_\vartheta &= \frac{I \Delta S (j\beta)^2}{4\pi r} \left(1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ E_\varphi &= -\frac{\zeta I \Delta S (j\beta)^2}{4\pi r} \left(1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{aligned} \right.$$

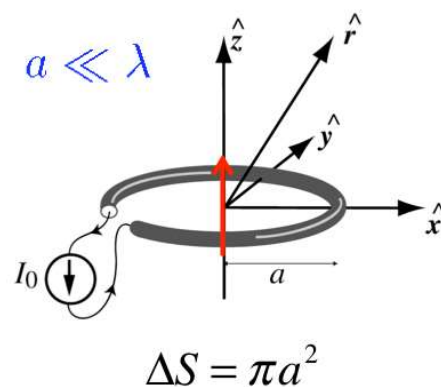
... for $r \gg \lambda$ ($\beta r \gg 1$) simplifies as

$$\left\{ \begin{aligned} H_r &= 0 \\ H_\vartheta &= -\frac{\beta \Delta S I}{2\lambda r} \sin \vartheta \exp(-j\beta r) = \frac{-E_\varphi}{\zeta} \\ E_\varphi &= \frac{\zeta \beta \Delta S I}{2\lambda r} \sin \vartheta \exp(-j\beta r) \end{aligned} \right.$$

Small loop antenna: far field

In the far-field case ($r \gg \lambda$) the small loop antenna behaves as follows

$$\begin{aligned} \vec{\mathbf{E}}(\vec{\mathbf{r}}) &= E_{\varphi}(r, \vartheta) \hat{i}_{\varphi} \\ \vec{\mathbf{H}}(\vec{\mathbf{r}}) &= H_{\vartheta}(r, \vartheta) \hat{i}_{\vartheta} \end{aligned} \quad \left\{ \begin{aligned} E_{\varphi} &= \frac{\zeta \beta \Delta s I}{2\lambda r} \sin \vartheta \exp(-j\beta r) \\ H_{\vartheta} &= -\frac{\beta \Delta s I}{2\lambda r} \sin \vartheta \exp(-j\beta r) = \frac{-E_{\varphi}}{\zeta} \end{aligned} \right.$$



- the e.m. field propagates along \hat{i}_r
- the e.m. field lies on the plane orthogonal to the propagation direction
- $|E|$ and $|H|$ exhibit the decaying factor $1/r$
- $|E|$ and $|H|$ are proportional through ζ

Small loop antenna: far field

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$$\zeta \vec{\mathbf{H}} = \zeta H_\vartheta \hat{i}_\vartheta = -E_\varphi \hat{i}_\vartheta$$

$$\hat{i}_r \times \vec{\mathbf{E}} = \hat{i}_r \times E_\varphi \hat{i}_\varphi = -E_\varphi \hat{i}_\vartheta$$



$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$

$$\hat{i}_\varphi = \hat{i}_r \times \hat{i}_\vartheta$$

$$\hat{i}_\vartheta = \hat{i}_\varphi \times \hat{i}_r$$

$$\hat{i}_r = \hat{i}_\vartheta \times \hat{i}_\varphi$$

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$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$

Small loop antenna: far field

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$$\vec{\mathbf{S}} = \frac{1}{2} \vec{\mathbf{E}} \times \vec{\mathbf{H}}^* = \frac{1}{2} E_\varphi \hat{i}_\varphi \times (H_\vartheta \hat{i}_\vartheta)^* = -\frac{1}{2} E_\varphi H_\vartheta^* \hat{i}_r$$

$$\begin{aligned} \hat{i}_\varphi &= \hat{i}_r \times \hat{i}_\vartheta \\ \hat{i}_\vartheta &= \hat{i}_\varphi \times \hat{i}_r \\ \hat{i}_r &= \hat{i}_\vartheta \times \hat{i}_\varphi \end{aligned}$$

$$-\frac{1}{2} E_\varphi H_\vartheta^* \hat{i}_r = \frac{1}{2\zeta} E_\varphi E_\varphi^* \hat{i}_r = \frac{1}{2\zeta} |E_\varphi|^2 \hat{i}_r = \frac{1}{2\zeta} |\vec{\mathbf{E}}|^2 \hat{i}_r$$

$$-\frac{1}{2} E_\varphi H_\vartheta^* \hat{i}_r = \frac{1}{2} \zeta H_\vartheta H_\vartheta^* \hat{i}_r = \frac{\zeta}{2} |H_\vartheta|^2 \hat{i}_r = \frac{\zeta}{2} |\vec{\mathbf{H}}|^2 \hat{i}_r$$

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$$\begin{aligned} \zeta \vec{\mathbf{H}} &= \hat{i}_r \times \vec{\mathbf{E}} \\ \vec{\mathbf{S}} &= \frac{1}{2\zeta} |\vec{\mathbf{E}}|^2 \hat{i}_r = \frac{\zeta}{2} |\vec{\mathbf{H}}|^2 \hat{i}_r \end{aligned}$$