

# **Campi Elettromagnetici**

**Corso di Laurea in Ingegneria Informatica,  
Biomedica e delle Telecomunicazioni**

**a.a. 2020-2021 - Laurea “Triennale” – Secondo semestre - Secondo anno**

**Università degli Studi di Napoli “Parthenope”**

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# Color legend

New formulas, important considerations,  
important formulas, important concepts

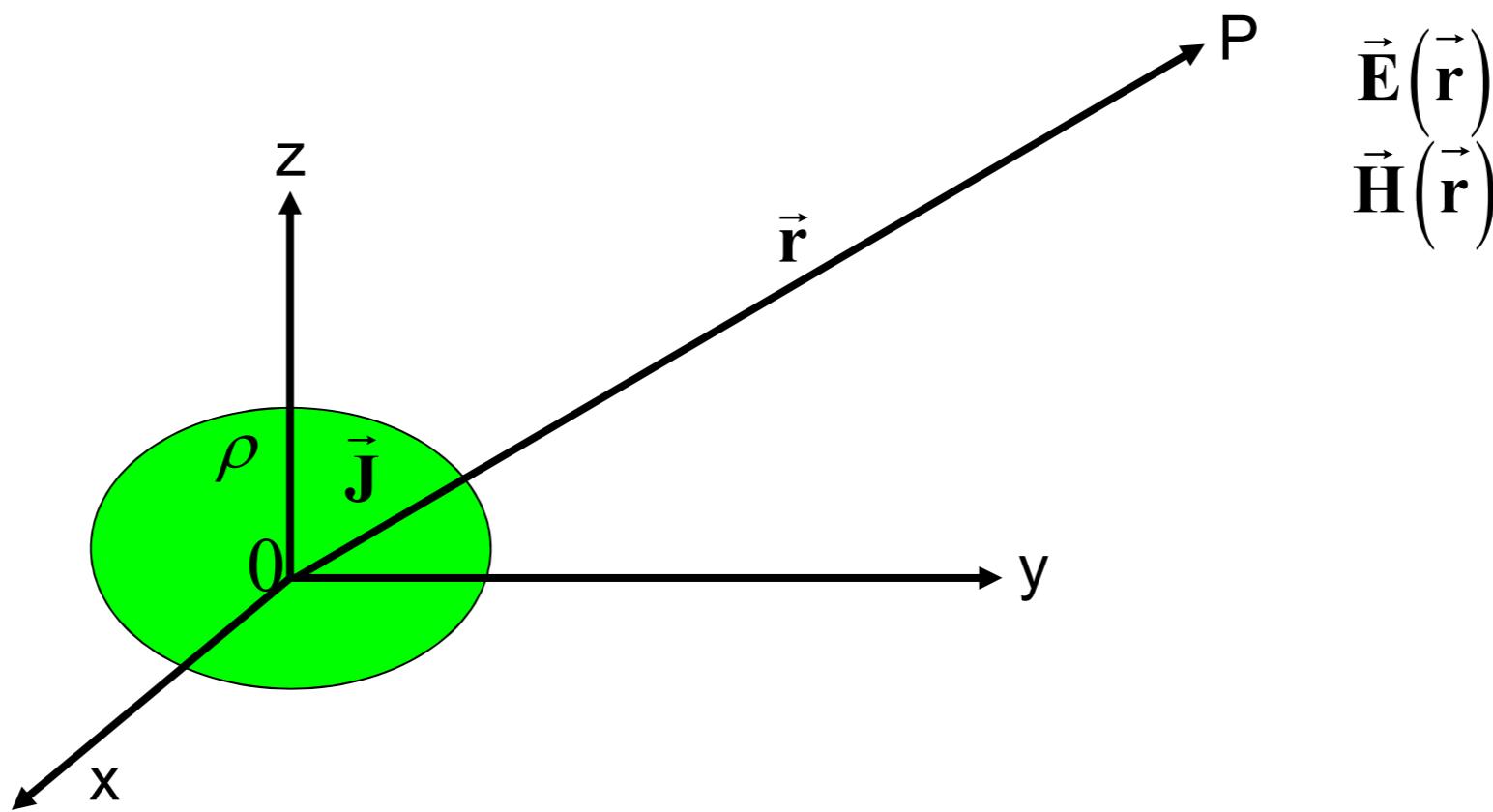
Very important for the discussion

Memo

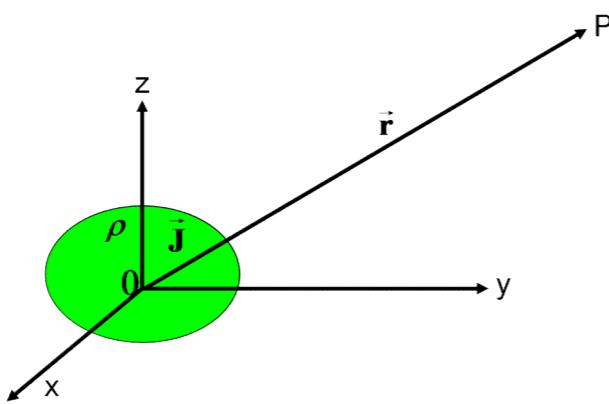
Mathematical tools to be exploited

Mathematics

# Radiation problem



An antenna (or aerial) is an electrical device which converts electric power into radio waves, and vice versa.



# Radiation problem

- We approach the radiation problem of an antenna assuming that the sources are known and the media in which propagation occurs is simple.
- Mathematically we exploit the potentials solution.

Medium: linear, isotropic, homogeneous (time-invariant and space-invariant), non-dispersive in space and time

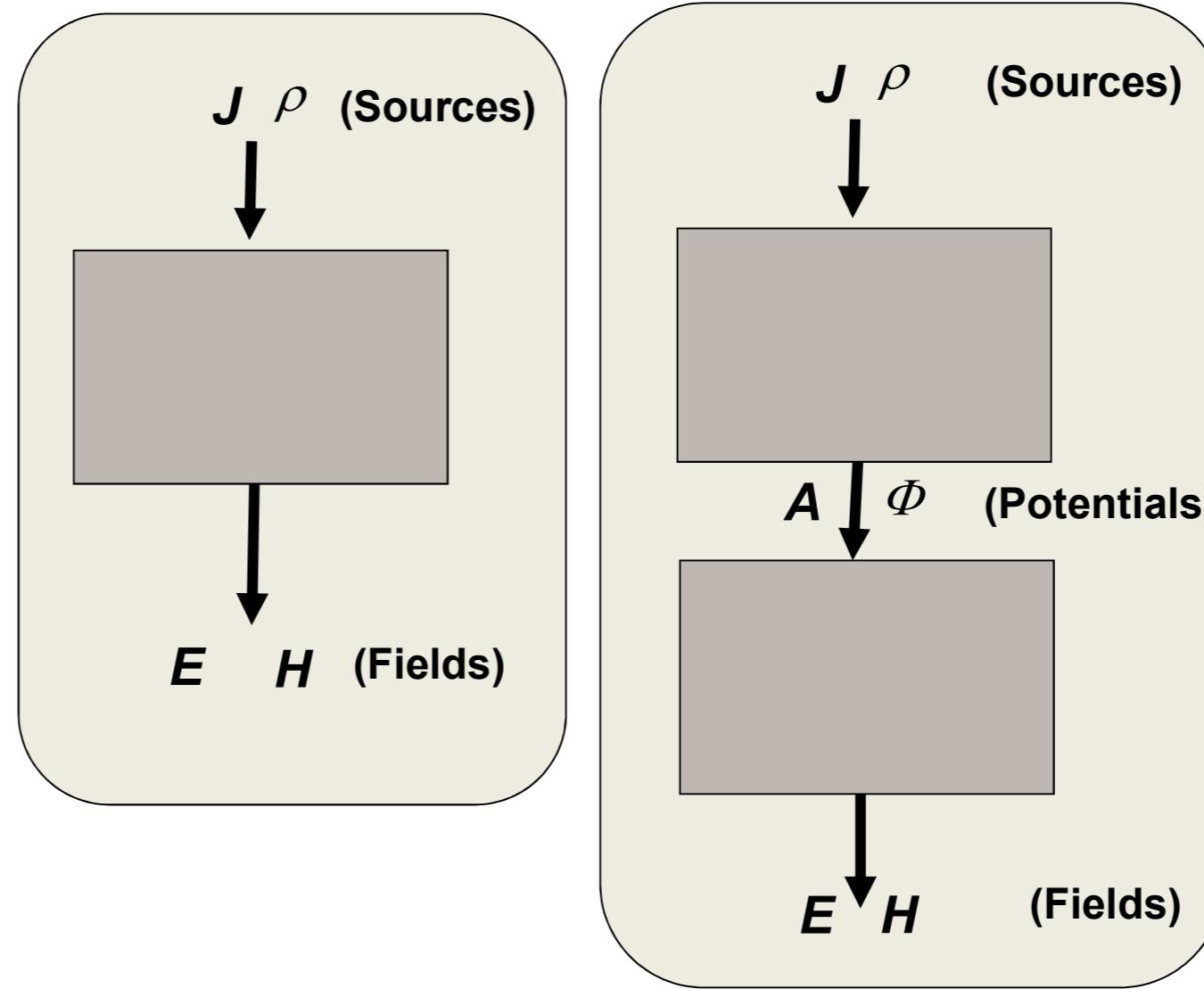
## Phasor domain

$$\left\{ \begin{array}{l} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} + \mathbf{J} \\ \nabla \cdot \epsilon\mathbf{E} = \rho \\ \nabla \cdot \mu\mathbf{H} = 0 \end{array} \right.$$



# Radiation problem & potentials

$$\left\{ \begin{array}{l} \nabla \times \mathbf{E} = -j\omega\mu\mathbf{H} \\ \nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E} + \mathbf{J} \\ \nabla \cdot \epsilon\mathbf{E} = \rho \\ \nabla \cdot \mu\mathbf{H} = 0 \end{array} \right.$$

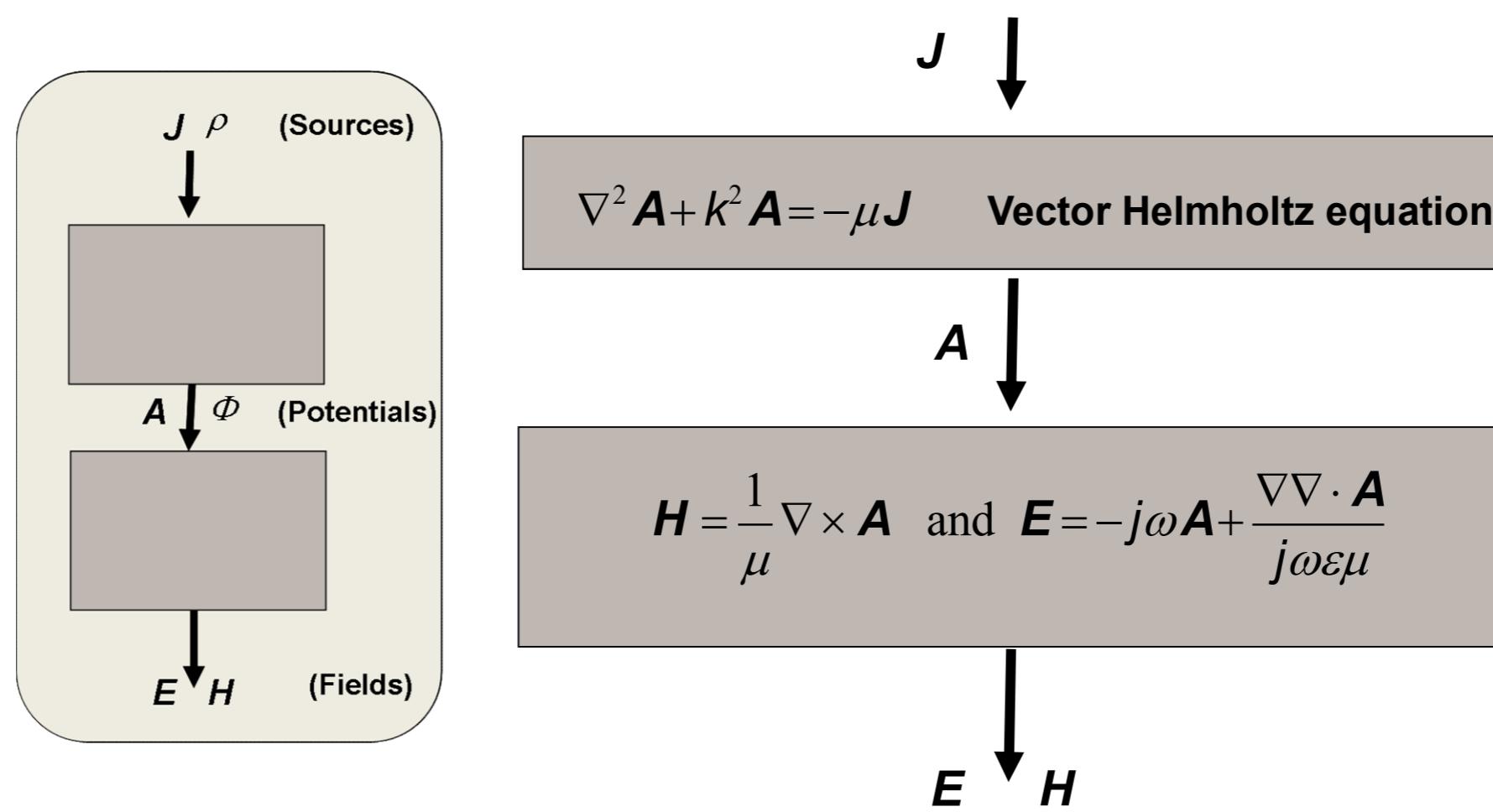


... mathematical tools that we will exploit today...

$$\nabla \times \mathbf{C} = \mathbf{0} \Rightarrow \exists \Phi : \mathbf{C} = \nabla \Phi$$

$$\nabla \cdot \mathbf{C} = 0 \Rightarrow \exists \mathbf{A} : \mathbf{C} = \nabla \times \mathbf{A}$$

# Potentials



# Potentials

$J$   
↓

$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J} \quad \text{Vector Helmholtz equation}$$

$A$   
↓

$$\begin{cases} \nabla^2 A_x + k^2 A_x = -\mu J_x \\ \nabla^2 A_y + k^2 A_y = -\mu J_y \\ \nabla^2 A_z + k^2 A_z = -\mu J_z \end{cases}$$

# Potentials

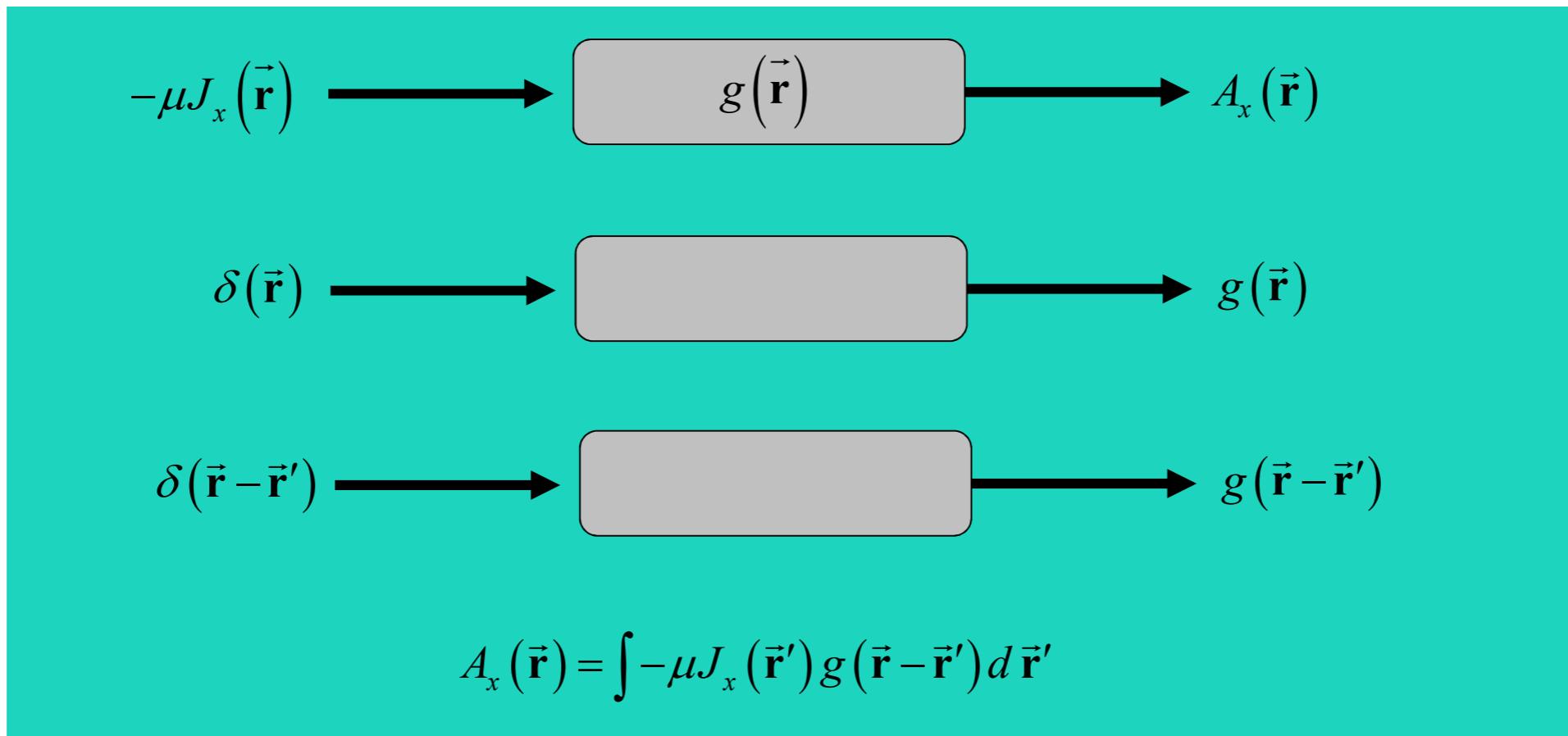
$$\begin{cases} \nabla^2 A_x + k^2 A_x = -\mu J_x \\ \nabla^2 A_y + k^2 A_y = -\mu J_y \\ \nabla^2 A_z + k^2 A_z = -\mu J_z \end{cases}$$

Let us address the solution of the following scalar Helmholtz equation

$$\nabla^2 A_x + k^2 A_x = -\mu J_x$$

# Potentials

$$\nabla^2 A_x + k^2 A_x = -\mu J_x$$

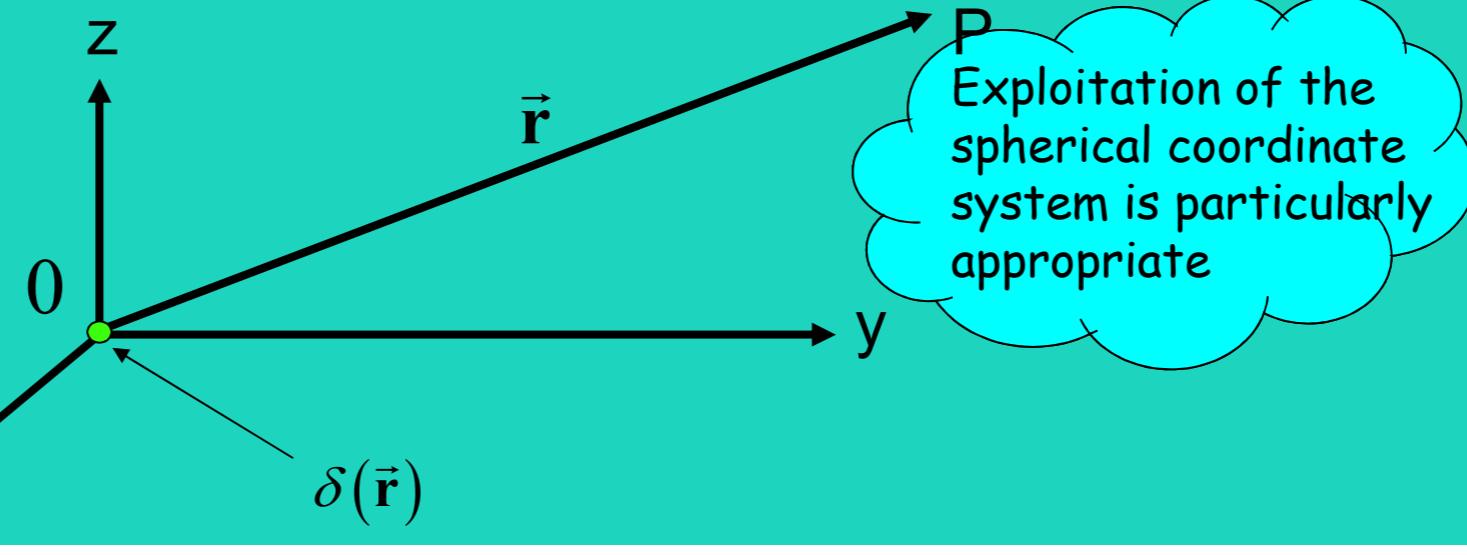


# Potentials

$$\nabla^2 A_x + k^2 A_x = -\mu J_x$$

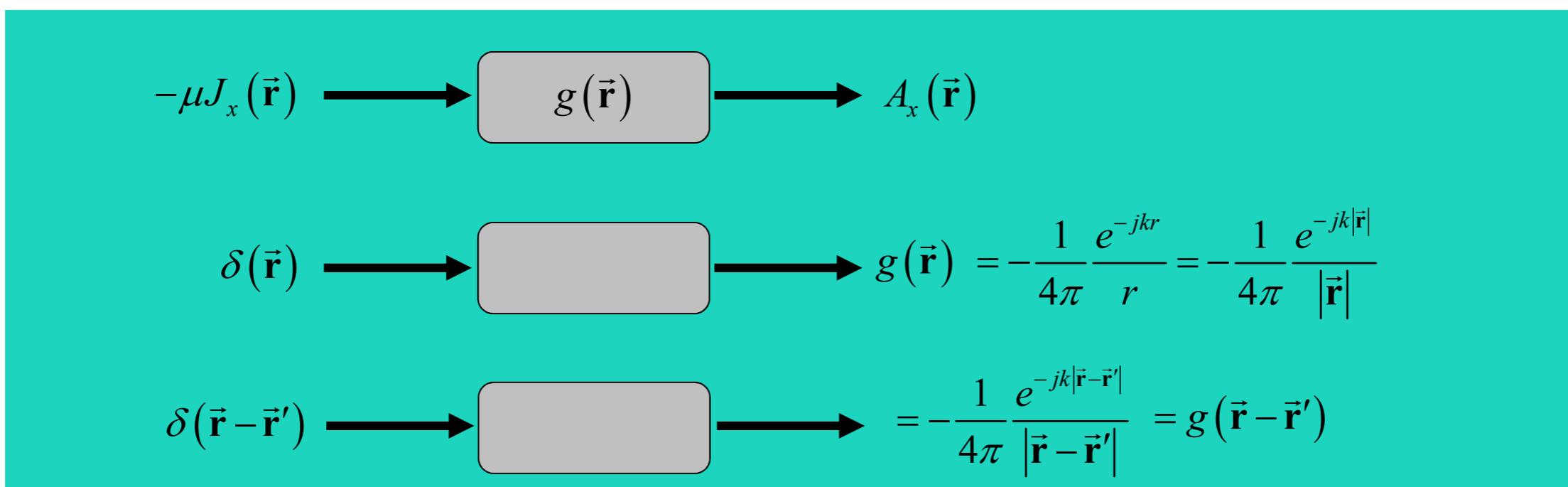
$$\delta(\vec{r}) \longrightarrow \text{[black box]} \longrightarrow g(\vec{r})$$

$$\nabla^2 g(\vec{r}) + k^2 g(\vec{r}) = \delta(\vec{r})$$



# Potentials

$$\nabla^2 A_x + k^2 A_x = -\mu J_x$$



$$A_x(\vec{r}) = \int -\mu J_x(\vec{r}') g(\vec{r} - \vec{r}') d\vec{r}' = \frac{\mu}{4\pi} \int J_x(\vec{r}') \frac{e^{-jk|\vec{r} - \vec{r}'|}}{|\vec{r} - \vec{r}'|} d\vec{r}'$$

# Potentials

$$\downarrow \mathbf{J}(\mathbf{r})$$

$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J} \quad \text{Vector Helmholtz equation}$$

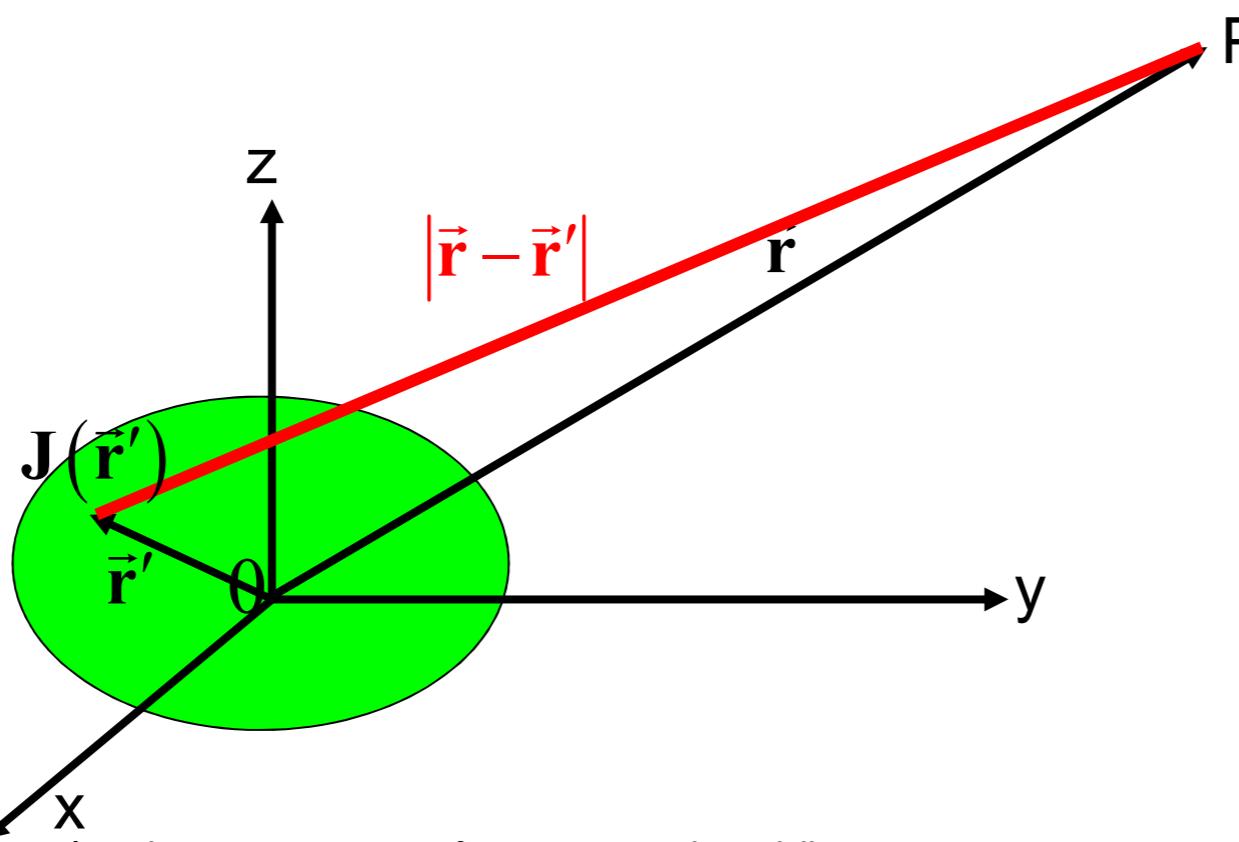
$$\mathbf{A}(\mathbf{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\mathbf{r}') \frac{e^{-jk|\mathbf{r}-\mathbf{r}'|}}{|\mathbf{r}-\mathbf{r}'|} d\mathbf{r}'$$

$$\mathbf{H} = \frac{1}{\mu} \nabla \times \mathbf{A} \quad \text{and} \quad \mathbf{E} = -j\omega \mathbf{A} + \frac{\nabla \nabla \cdot \mathbf{A}}{j\omega \epsilon \mu}$$

$$\downarrow \mathbf{E}(\mathbf{r}) \\ \mathbf{H}(\mathbf{r})$$

# Potentials

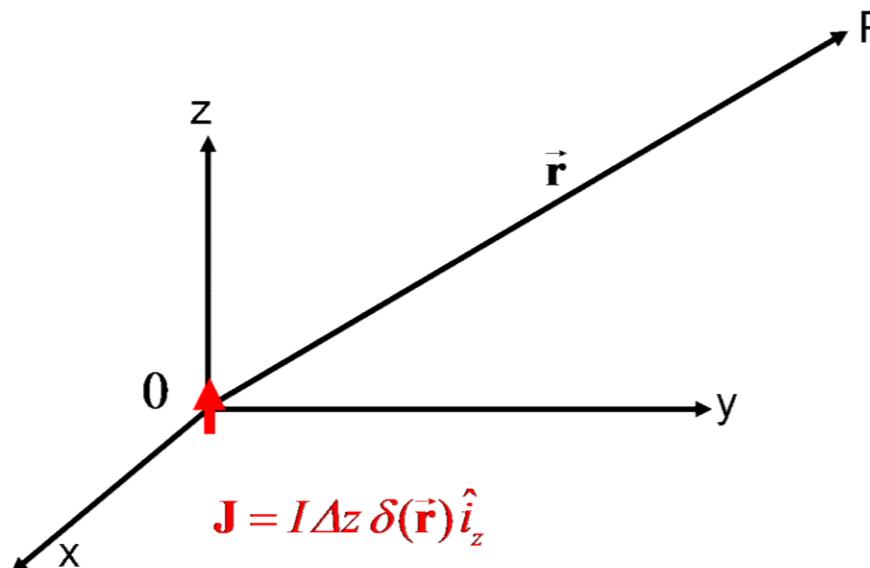
$$\mathbf{A}(\vec{r}) = \frac{\mu}{4\pi} \int \mathbf{J}(\vec{r}') \frac{e^{-jk|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} d\vec{r}'$$



# Elementary electrical dipole

- A  $\delta$ -source radiating element is also known as elementary electrical dipole.

$$\mathbf{J} = I \Delta z \delta(\vec{\mathbf{r}}) \hat{i}_z = I \Delta z \delta(x) \delta(y) \delta(z) \hat{i}_z$$



- Why are we interested in such a radiating element?
- Why is such a radiating element referred to as elementary electrical dipole?
- How can we physically approximate an elementary electrical dipole?

# Elementary electrical dipole

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$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J}$$

**Vector Helmholtz equation**

$$\begin{cases} \nabla^2 A_x + k^2 A_x = -\mu J_x \\ \nabla^2 A_y + k^2 A_y = -\mu J_y \\ \nabla^2 A_z + k^2 A_z = -\mu J_z \end{cases}$$

$\mathbf{J}$   
↓  
 $\mathbf{A}$   
↓

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↓

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Vector Helmholtz equation

$$\begin{cases} \nabla^2 A_x + k^2 A_x = 0 \\ \nabla^2 A_y + k^2 A_y = 0 \\ \nabla^2 A_z + k^2 A_z = -\mu J_z = -\mu I \Delta z \delta(x) \delta(y) \delta(z) = -\mu I \Delta z \delta(\vec{r}) \end{cases}$$

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$J$   
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$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J}$$

Vector Helmholtz equation

$$\begin{aligned}\nabla^2 A_x + k^2 A_x &= 0 & \Rightarrow A_x &= 0 \\ \nabla^2 A_y + k^2 A_y &= 0 & \Rightarrow A_y &= 0\end{aligned}$$

$$\nabla^2 A_z + k^2 A_z = -\mu J_z = -\mu I \Delta z \delta(x) \delta(y) \delta(z) = -\mu I \Delta z \delta(\vec{r})$$

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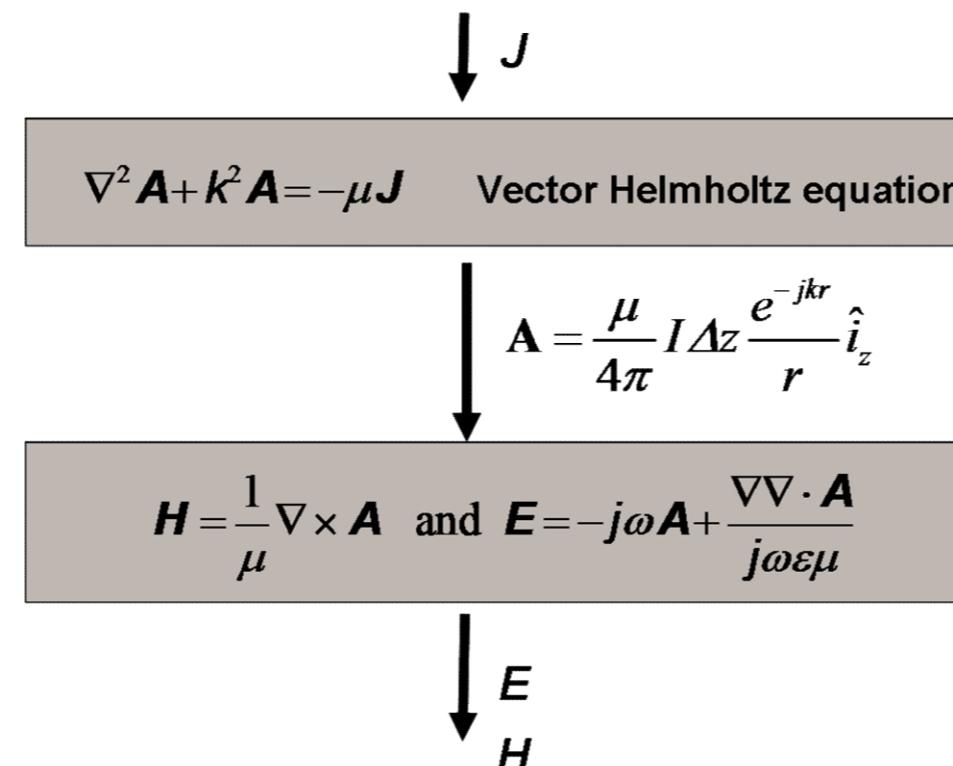
$$\nabla^2 g(\vec{r}) + k^2 g(\vec{r}) = -\mu I \Delta z \delta(\vec{r})$$

$$-\mu I \Delta z \delta(\vec{r}) \rightarrow \text{[redacted]} \rightarrow g(\vec{r}) = (-\mu I \Delta z) - \frac{1}{4\pi} \frac{e^{-jkr}}{r} = \frac{\mu}{4\pi} I \Delta z \frac{e^{-jkr}}{r}$$

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$$\mathbf{J} = I \Delta z \delta(\vec{r}) \hat{i}_z = I \Delta z \delta(x) \delta(y) \delta(z) \hat{i}_z$$

The E.M. field radiated by the elementary electrical dipole

$$\begin{aligned}\vec{\mathbf{E}}(\vec{r}) &= E_r(r, \vartheta) \hat{i}_r + E_\vartheta(r, \vartheta) \hat{i}_\vartheta \\ \vec{\mathbf{H}}(\vec{r}) &= H_\varphi(r, \vartheta) \hat{i}_\varphi\end{aligned}\quad \left\{ \begin{array}{l} E_r = \zeta \frac{I \Delta z}{2\pi} \left( \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I \Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I \Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{array} \right.$$

Because of the problem symmetry there is no dependence on the azimuth angle  $\varphi$ .

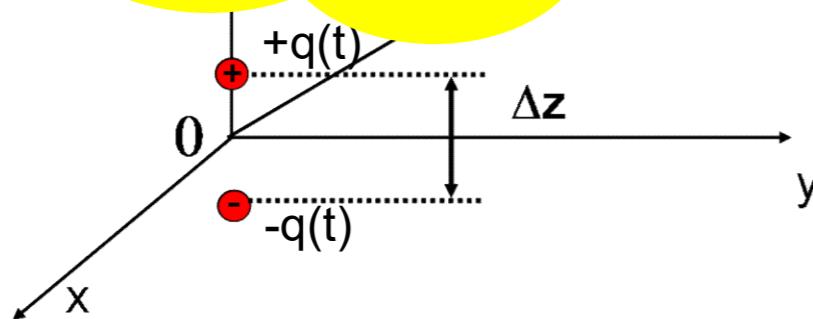
# Elementary electrical dipole

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$$\mathbf{J} = I\Delta z \delta(\vec{r}) \hat{i}_z = I\Delta z \delta(x)\delta(y)\delta(z) \hat{i}_z$$

All the quantities, included the expressions of the fields, can be provided in terms of dipole moment  $U$

$$I\Delta z = j\omega Q\Delta z = j\omega U$$



It can be easily shown that this electric current density is the same as that of an electrical dipole such that:

- 1) the two charges, of opposite sign, have equal time variation;
- 2) in the spectral domain, the relation between  $I$  and the time-varying charge  $Q$  is:

$$j\omega Q = I$$

# Elementary electrical dipole

The E.M. field radiated by the elementary electrical dipole

$$\begin{cases} E_r = \zeta \frac{I\Delta z}{2\pi} \left( \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases} \quad j\omega Q = I$$

# Elementary electrical dipole: static case

The E.M. field radiated by the elementary electrical dipole

$$\begin{cases} E_r = \zeta \frac{j\omega Q \Delta z}{2\pi} \left( \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{j\omega Q \Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{j\omega Q \Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases} \quad j\omega Q = I$$

... for  $\omega=0$  simplifies as

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$j\omega Q = I$

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$$\begin{cases} E_r = \frac{\zeta}{\sqrt{\mu\epsilon}} \frac{Q\Delta z}{2\pi} \frac{1}{r^3} \cos \vartheta \\ E_\vartheta = \frac{\zeta}{\sqrt{\mu\epsilon}} \frac{Q\Delta z}{4\pi} \frac{1}{r^3} \sin \vartheta \\ H_\varphi = 0 \end{cases}$$

$$\beta = \omega \sqrt{\mu\epsilon} \rightarrow 0$$

$$\rightarrow \omega\beta = \omega^2 \sqrt{\mu\epsilon} \rightarrow 0$$

$$\rightarrow \omega/\beta = 1/\sqrt{\mu\epsilon}$$

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$$\zeta = \sqrt{\frac{\mu}{\epsilon}}$$

$$\rightarrow \frac{\zeta}{\sqrt{\mu\epsilon}} = \frac{1}{\epsilon}$$

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The E.M. field radiated by the elementary electrical dipole

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$j\omega Q = I$

... for  $\omega=0$  simplifies as

$$\begin{cases} E_r = \frac{\zeta}{\sqrt{\mu\epsilon}} \frac{Q\Delta z}{2\pi} \frac{1}{r^3} \cos \vartheta = \frac{Q\Delta z}{2\pi} \frac{1}{\epsilon r^3} \cos \vartheta \\ E_\vartheta = \frac{\zeta}{\sqrt{\mu\epsilon}} \frac{Q\Delta z}{4\pi} \frac{1}{r^3} \sin \vartheta = \frac{Q\Delta z}{4\pi} \frac{1}{\epsilon r^3} \sin \vartheta \\ H_\varphi = 0 \end{cases}$$

$$\zeta = \sqrt{\frac{\mu}{\epsilon}}$$

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# Elementary electrical dipole: far field

The E.M. field radiated by the elementary electrical dipole

$$\begin{cases} E_r = \zeta \frac{I\Delta z}{2\pi} \left( \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\phi = \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases} = \zeta \frac{I\Delta z}{2\pi} \frac{j\beta}{r} \left( \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) = \zeta \frac{I\Delta z}{4\pi} \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) = \frac{I\Delta z}{4\pi} \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r)$$

$$\left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) = \frac{j\beta}{r} \left( 1 + \frac{r}{j\beta} \frac{1}{r^2} + \frac{r}{j\beta} \frac{1}{j\beta r^3} \right) = \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right)$$

# Elementary electrical dipole: far field

The E.M. field radiated by the elementary electrical dipole

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# Elementary electrical dipole: far field

The E.M. field radiated by the elementary electrical dipole

$$\begin{cases} E_r = \zeta \frac{I\Delta z}{2\pi} \left( \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\phi = \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases} \quad \begin{aligned} &= \zeta \frac{I\Delta z}{2\pi} \frac{j\beta}{r} \left( \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ &= \zeta \frac{I\Delta z}{4\pi} \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ &= \frac{I\Delta z}{4\pi} \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{aligned}$$

... for  $\beta r \gg 1$  simplifies as

$$\frac{\beta}{4\pi} = \frac{2\pi}{\lambda} \frac{1}{4\pi} = \frac{1}{2\lambda}$$

$$\begin{cases} E_r = 0 \\ E_\vartheta = j\zeta \frac{I\Delta z}{r} \frac{\beta}{4\pi} \sin \vartheta \exp(-j\beta r) = j\zeta \frac{I\Delta z}{2\lambda r} \sin \vartheta \exp(-j\beta r) \\ H_\phi = j \frac{I\Delta z}{r} \frac{\beta}{4\pi} \sin \vartheta \exp(-j\beta r) = \frac{E_\vartheta}{\zeta} \end{cases}$$

$$\beta = \omega \sqrt{\mu \epsilon} = \frac{2\pi f}{c} = \frac{2\pi}{\lambda}$$

$$\begin{aligned} \beta &= \omega \sqrt{\mu \epsilon} \\ 2\pi f &= \omega \\ c &= 1/\sqrt{\mu \epsilon} \\ \lambda &= c/f \end{aligned}$$

# Elementary electrical dipole: far field

The E.M. field radiated by the elementary electrical dipole

$$\begin{cases} E_r = \zeta \frac{I\Delta z}{2\pi} \left( \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \cos \vartheta \exp(-j\beta r) \\ E_\vartheta = \zeta \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} + \frac{1}{j\beta r^3} \right) \sin \vartheta \exp(-j\beta r) \\ H_\varphi = \frac{I\Delta z}{4\pi} \left( \frac{j\beta}{r} + \frac{1}{r^2} \right) \sin \vartheta \exp(-j\beta r) \end{cases} \quad \begin{aligned} &= \zeta \frac{I\Delta z}{2\pi} \frac{j\beta}{r} \left( \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \cos \vartheta \exp(-j\beta r) \\ &= \zeta \frac{I\Delta z}{4\pi} \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} + \frac{1}{[j\beta r]^2} \right) \sin \vartheta \exp(-j\beta r) \\ &= \frac{I\Delta z}{4\pi} \frac{j\beta}{r} \left( 1 + \frac{1}{j\beta r} \right) \sin \vartheta \exp(-j\beta r) \end{aligned}$$

... for  $\beta r \gg 1$  simplifies as

$$\beta r \gg 1 \Rightarrow \frac{2\pi}{\lambda} r \gg 1 \Rightarrow r \gg \lambda$$

$$\begin{cases} E_r = 0 \\ E_\vartheta = j\zeta \frac{I\Delta z}{r} \frac{\beta}{4\pi} \sin \vartheta \exp(-j\beta r) = j\zeta \frac{I\Delta z}{2\lambda r} \sin \vartheta \exp(-j\beta r) \\ H_\varphi = j \frac{I\Delta z}{r} \frac{\beta}{4\pi} \sin \vartheta \exp(-j\beta r) = \frac{E_\vartheta}{\zeta} \end{cases}$$

$$\beta = \omega \sqrt{\mu \epsilon} = \frac{2\pi f}{c} = \frac{2\pi}{\lambda}$$

$$\begin{aligned} \beta &= \omega \sqrt{\mu \epsilon} \\ 2\pi f &= \omega \\ c &= 1/\sqrt{\mu \epsilon} \\ \lambda &= c/f \end{aligned}$$

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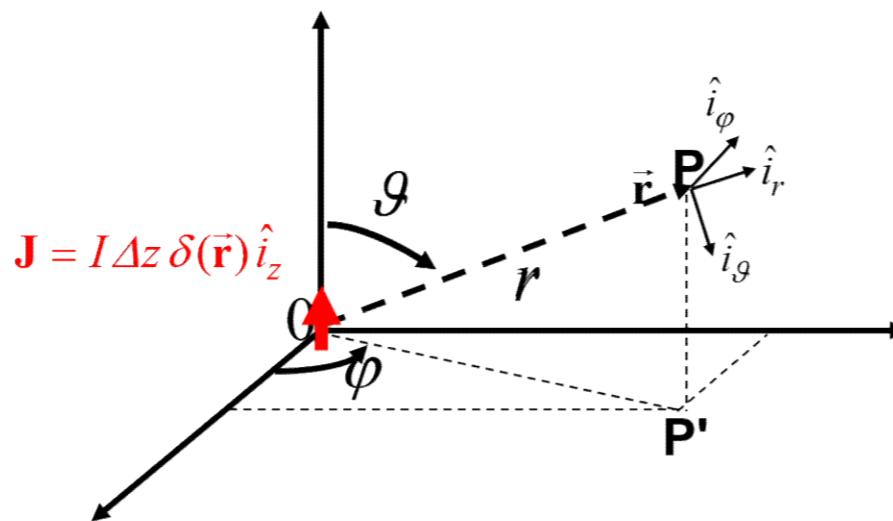
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In the far-field case ( $r \gg \lambda$ ) the elementary electrical dipole behaves as follows

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- the e.m. field propagates along  $\hat{i}_r$
- the e.m. field lies on the plane orthogonal to the propagation direction
- $|E|$  and  $|H|$  exhibit the decaying factor  $1/r$
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$$\zeta \vec{H} = \zeta H_\varphi \hat{i}_\varphi = E_\vartheta \hat{i}_\varphi$$



$$\zeta \vec{H} = \hat{i}_r \times \vec{E}$$

$$\hat{i}_r \times \vec{E} = \hat{i}_r \times E_\vartheta \hat{i}_\vartheta = E_\vartheta \hat{i}_\varphi$$

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$$\vec{S} = \frac{1}{2} \vec{E} \times \vec{H}^* = \frac{1}{2} E_\vartheta \hat{i}_\vartheta \times (H_\varphi \hat{i}_\varphi)^* = \frac{1}{2} E_\vartheta H_\varphi^* \hat{i}_r$$

$$\begin{aligned} \hat{i}_\varphi &= \hat{i}_r \times \hat{i}_\vartheta \\ \hat{i}_\vartheta &= \hat{i}_\varphi \times \hat{i}_r \\ \hat{i}_r &= \hat{i}_\vartheta \times \hat{i}_\varphi \end{aligned}$$

$$\frac{1}{2} E_\vartheta H_\varphi^* \hat{i}_r = \frac{1}{2\zeta} E_\vartheta E_\vartheta^* \hat{i}_r = \frac{1}{2\zeta} |E_\vartheta|^2 \hat{i}_r = \frac{1}{2\zeta} |\vec{E}|^2 \hat{i}_r$$

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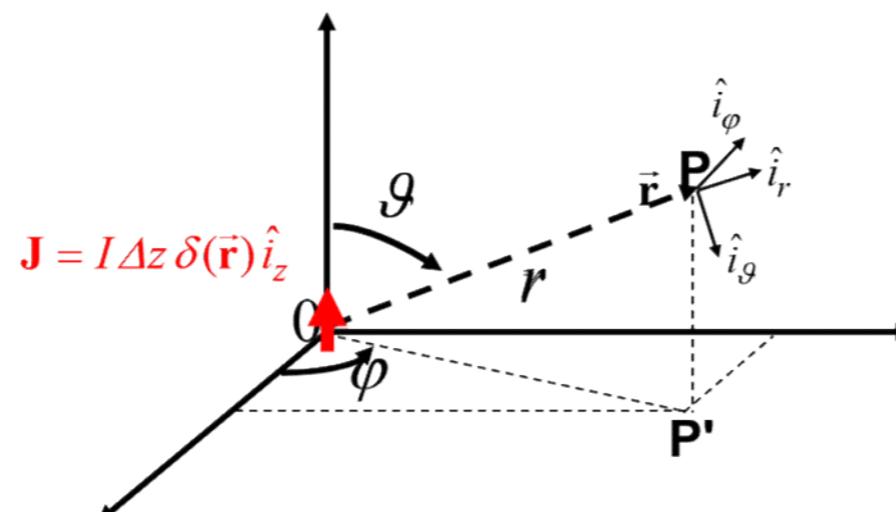
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$$\begin{aligned} \zeta \vec{H} &= \hat{i}_r \times \vec{E} \\ \vec{S} &= \frac{1}{2\zeta} |\vec{E}|^2 \hat{i}_r = \frac{\zeta}{2} |\vec{H}|^2 \hat{i}_r \end{aligned}$$

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MEMO .....

The radiation condition

$\vec{\mathbf{E}} \sim O\left(\frac{1}{r}\right)$	$\vec{\mathbf{H}} \sim O\left(\frac{1}{r}\right)$	$\vec{\mathbf{E}} - \zeta \vec{\mathbf{H}} \times \hat{\mathbf{n}} \sim o\left(\frac{1}{r}\right)$	$\left( \text{and } \zeta \vec{\mathbf{H}} - \hat{\mathbf{n}} \times \vec{\mathbf{E}} \sim o\left(\frac{1}{r}\right) \right)$	as $r \rightarrow \infty$	PD
$\hat{\mathbf{n}} \cdot \vec{\mathbf{E}} = \hat{\mathbf{n}} \cdot \vec{\mathbf{H}} = 0$					

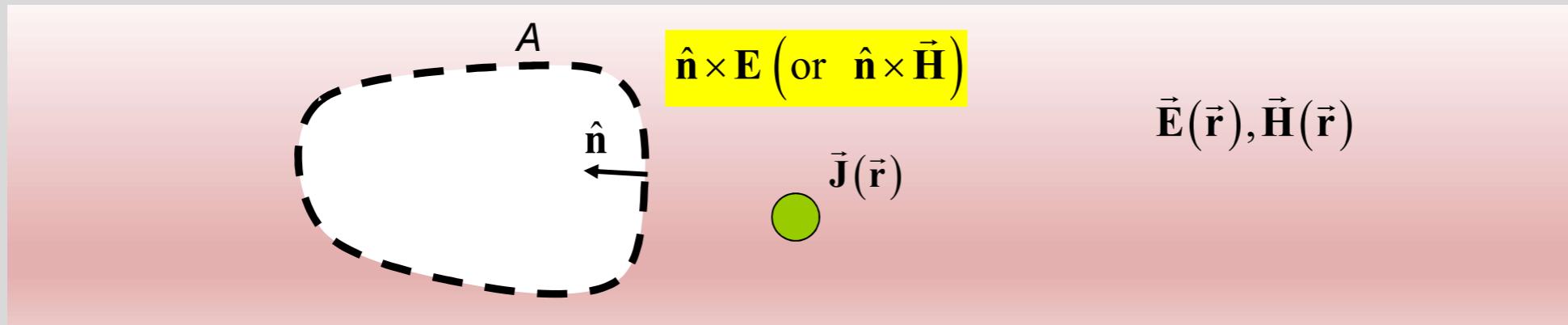


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# MEMO.....Uniqueness (PD-Exterior Problem)



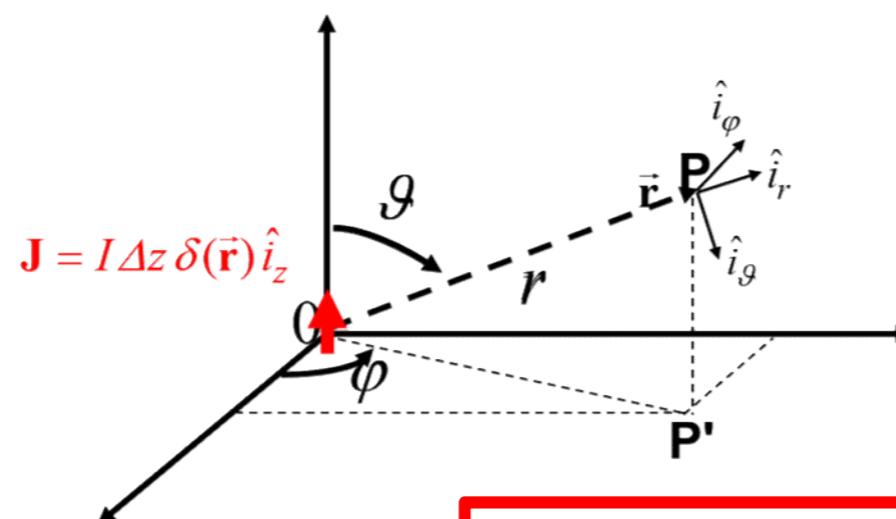
- I Consider a source distribution  $\vec{J}(\vec{r})$  with its associated electromagnetic field  $\vec{E}(\vec{r}), \vec{H}(\vec{r})$
- II Consider a (smooth) surface  $A$  with an everywhere defined unit normal  $\hat{n}$
- IV Consider the values of the tangential component of the electric (or magnetic) field upon the surface  $A$ ; that is, consider  $\hat{n} \times \mathbf{E}$  (or  $\hat{n} \times \mathbf{H}$ ) **on the boundary**

The Uniqueness Theorem states that the electromagnetic field produced by the source in (I) within the **infinite volume V outside** the surface  $A$  in (II), enforcing **the boundary condition** in (IV) **as well as the radiation condition at infinity** is unique.

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Uniqueness is guaranteed

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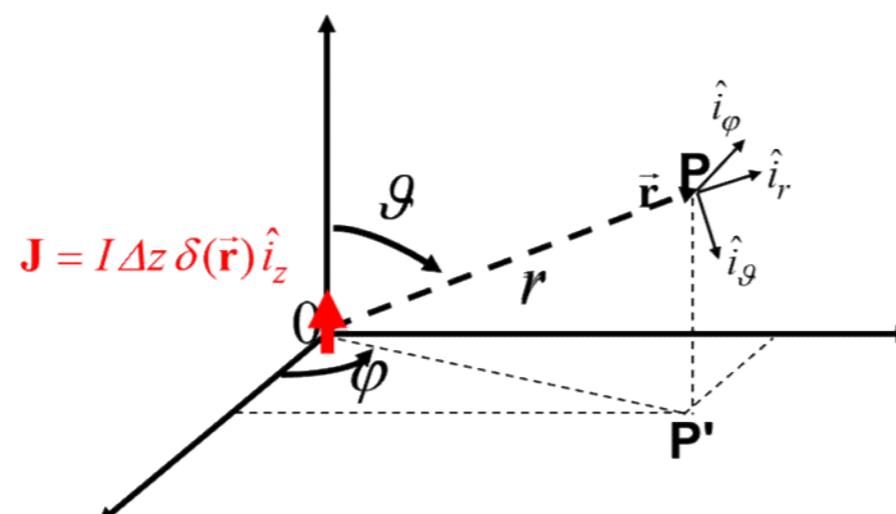
$$\zeta \mathbf{H} = \hat{i}_r \times \mathbf{E}$$

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# ....MEMO.....Plane Waves (TD)

$$\frac{\partial^2 e_x}{\partial z^2} - \frac{1}{c^2} \frac{\partial^2 e_x}{\partial t^2} = 0$$

$\{e_x, h_y\}$

$\{e_x^+, h_y^+\}$      $\vec{s}^+ = \frac{|e_x^+(z-ct)|^2}{\zeta} \hat{i}_z$

$\{e_x^-, h_y^-\}$      $\vec{s}^- = -\frac{|e_x^-(z+ct)|^2}{\zeta} \hat{i}_z$

$$\frac{\partial^2 e_y}{\partial z^2} - \frac{1}{c^2} \frac{\partial^2 e_y}{\partial t^2} = 0$$

$\{e_y, h_x\}$

$\{e_y^+, h_x^+\}$      $\vec{s}^+ = \frac{|e_y^+(z-ct)|^2}{\zeta} \hat{i}_z$

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<b>Source-free</b> <b>Medium</b> - Linear - Local (TND & SND) - Isotropic - Homogeneous (TI – SI) - Lossless
$\frac{\partial}{\partial x} = \frac{\partial}{\partial y} = 0$ $\vec{e}(\vec{r}, t) = \vec{e}(z, t)$ $\vec{h}(\vec{r}, t) = \vec{h}(z, t)$

↓

$$e_z(z, t) = h_z(z, t) = 0$$

$\{e_x^+, h_y^+\}$      $\{e_x^-, h_y^-\}$

$\{e_y^+, h_x^+\}$      $\{e_y^-, h_x^-\}$

In all these 4 cases the electric and magnetic fields are related each other through the following relation:

$$\zeta \vec{h} = \hat{i}_p \times \vec{e}$$

where  $\hat{i}_p$  points to the propagation direction

$\{e_y, h_x\}$     Independent each other  
 $\{e_x, h_y\}$

# ..MEMO... Plane Waves (Phasor Domain)

$$\{E_x, H_y\}$$

$$E_x^+(z) = E^+ e^{-j\beta z}$$

$$\zeta H_y^+(z) = E^+ e^{-j\beta z}$$

$$k(\omega_0) = \omega_0 \sqrt{\mu \epsilon} = \beta(\omega_0)$$

$$v_p = \frac{\omega_0}{\beta} = \frac{\omega_0}{\omega_0 \sqrt{\mu \epsilon}} = \frac{1}{\sqrt{\mu \epsilon}} = c$$

$$\zeta = \sqrt{\frac{\mu}{\epsilon}}$$

Source-free

Medium

- Linear
- Time nondispersive
- Space non-dispersive
- Isotropic
- Homogeneous (TI – SI)
- Lossless

$$E^+ e^{-j\beta z} \rightarrow e_x^+(z, t) = |E^+| \cos(\omega_o t - \beta z + \varphi^+) = e_x^+(z - v_p t)$$

**Progressive plane wave**

$$\vec{S}(\vec{r}) = \frac{1}{2} \vec{E}(\vec{r}) \times \vec{H}^*(\vec{r})$$

$$\vec{S}^+(\vec{r}) = \frac{1}{2} \vec{E}^+(\vec{r}) \times \vec{H}^{+*}(\vec{r}) = \frac{|E_x^+(z)|^2}{2\zeta} \hat{i}_z = \zeta \frac{|H_y^+(z)|^2}{2} \hat{i}_z$$

Time nondispersive & lossless

$$\begin{cases} \epsilon : real \\ \mu : real \\ \sigma = 0 \end{cases}$$

$$\epsilon_{eq} = \epsilon$$

$$c = \frac{1}{\sqrt{\mu \epsilon}}$$

$$\frac{\partial}{\partial x} = \frac{\partial}{\partial y} = 0$$

$$\vec{e}(\vec{r}, t) = \vec{e}(z, t)$$

$$\vec{h}(\vec{r}, t) = \vec{h}(z, t)$$

$$E_z = H_z = 0$$

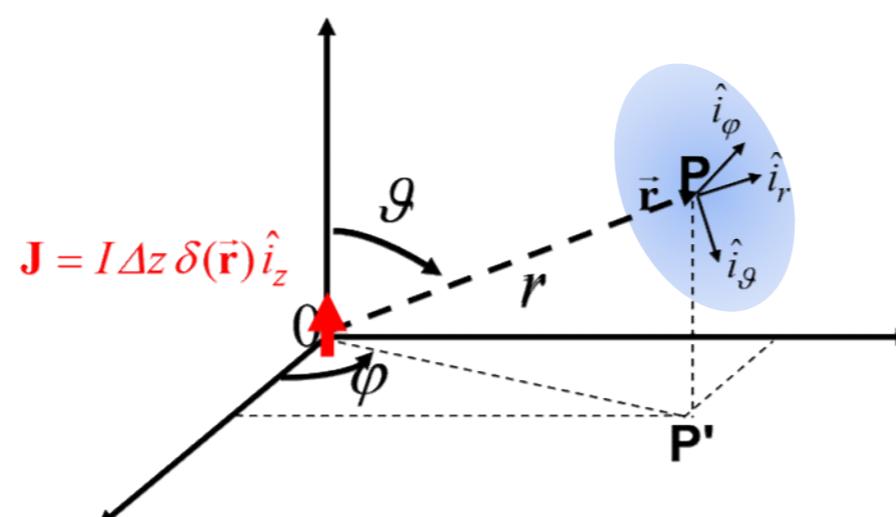
$$k(\omega_0) = \omega_0 \sqrt{\mu \epsilon} = \beta(\omega_0)$$

$\{E_y, H_x\}$  Independent  
 $\{E_x, H_y\}$  each other

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In the far field case, this field locally behaves as a plane wave

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