

Intelligent Signal Processing

Fast Fourier Transform

Angelo Ciaramella

Introduction

- The **Discrete Fourier Transform (DFT)** has an important role for signal analysis
- In the sixties of the last century **a fast approach** for DFT was introduced
 - **Fast Fourier Transform**
 - Work of **Cooley** and **Tukey**



Classis

- Decimation in time
 - The source signal $x(n)$ is divided in shorter sequences
- Decimation in frequency
 - The DFT coefficients $X(k)$ are divided in shorter sequences



DFT

$$X(k) = \begin{cases} \sum_{n=0}^{N-1} x(n)W_N^{kn} & 0 \leq k \leq N-1 \\ 0 & \text{altrove} \end{cases}$$

Analysis

$$x(n) = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X(k)W_N^{-kn} & 0 \leq n \leq N-1 \\ 0 & \text{altrove} \end{cases}$$

Synthesis

$$X(k) = \sum_{n=0}^{N-1} \left\{ \left(\operatorname{Re}[x(n)] \operatorname{Re}[W_N^{kn}] - \operatorname{Im}[x(n)] \operatorname{Im}[W_N^{kn}] \right) + j \left(\operatorname{Re}[x(n)] \operatorname{Im}[W_N^{kn}] + \operatorname{Im}[x(n)] \operatorname{Re}[W_N^{kn}] \right) \right\}$$

$$k = 0, 1, \dots, N-1$$



DFT

$$X(k) = \begin{cases} \sum_{n=0}^{N-1} x(n)W_N^{kn} & 0 \leq k \leq N-1 \\ 0 & \text{altrove} \end{cases}$$

Analysis

$$x(n) = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X(k)W_N^{-kn} & 0 \leq n \leq N-1 \\ 0 & \text{altrove} \end{cases}$$

Synthesis

$$X(k) = \sum_{n=0}^{N-1} \left\{ \left(\operatorname{Re}[x(n)] \operatorname{Re}[W_N^{kn}] - \operatorname{Im}[x(n)] \operatorname{Im}[W_N^{kn}] \right) + j \left(\operatorname{Re}[x(n)] \operatorname{Im}[W_N^{kn}] + \operatorname{Im}[x(n)] \operatorname{Re}[W_N^{kn}] \right) \right\}$$

$$k = 0, 1, \dots, N-1$$

$X(k)$ needs of $4N$ real products and $(4N-1)$ real sums for each k . Totally, we have $4N^2$ real products e $N(4N-1)$ real sums.



Time decimation

- We use the **symmetry** and **periodicity** of the complex exponential

$$W_N^{kn} = e^{-j\left(\frac{2\pi}{N}\right)kn}$$

- The **sequence** is a **power of two**

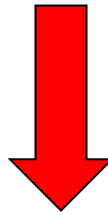
$$N = 2^v$$



Time decimation

- $X(k)$ is calculated dividing $x(n)$ in **two** subsequences

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j\left(\frac{2\pi}{N}\right)kn}$$



$$X(k) = \sum_{\substack{n=0 \\ \text{n even}}}^{N-1} x(n) e^{-j\left(\frac{2\pi}{N}\right)kn} + \sum_{\substack{n=0 \\ \text{n odd}}}^{N-1} x(n) e^{-j\left(\frac{2\pi}{N}\right)kn}$$



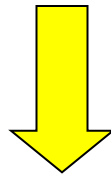
Time decimation

$$n = 2r$$

$$n = 2r+1$$

$$\begin{aligned} X(k) &= \sum_{r=0}^{N/2-1} x(2r)W_N^{2rk} + \sum_{r=0}^{N/2-1} x(2r+1)W_N^{(2r+1)k} \\ &= \sum_{r=0}^{N/2-1} x(2r)(W_N^2)^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1)(W_N^2)^{rk} \end{aligned}$$

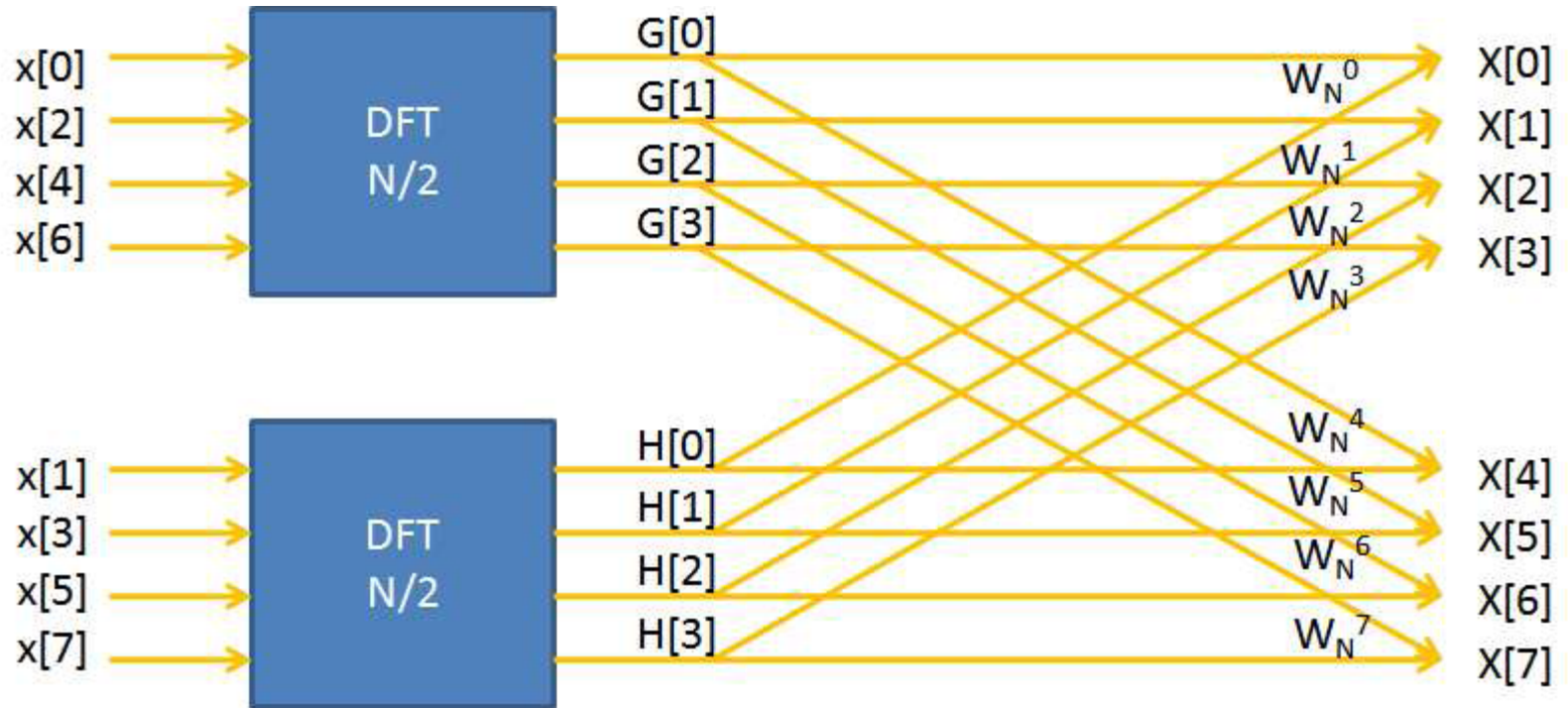
$$\begin{aligned} W_N^2 &= e^{-j2\pi/N} = \\ e^{-j2\pi/(N/2)} &= W_{N/2} \end{aligned}$$



$$\begin{aligned} X(k) &= \sum_{r=0}^{N/2-1} x(2r)W_{N/2}^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1)W_{N/2}^{rk} \\ &= G(k) + W_N^k H(k) \end{aligned}$$



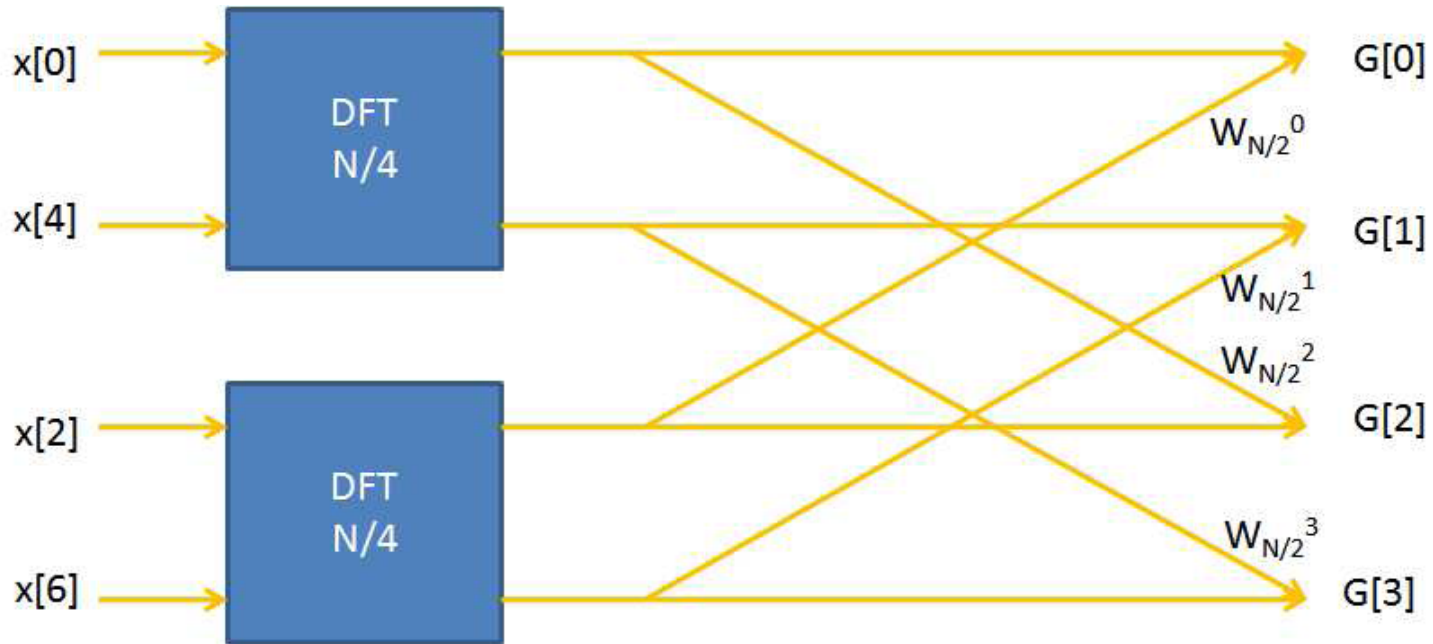
Flow graph



Flow Graph for a DFT with $N=8$



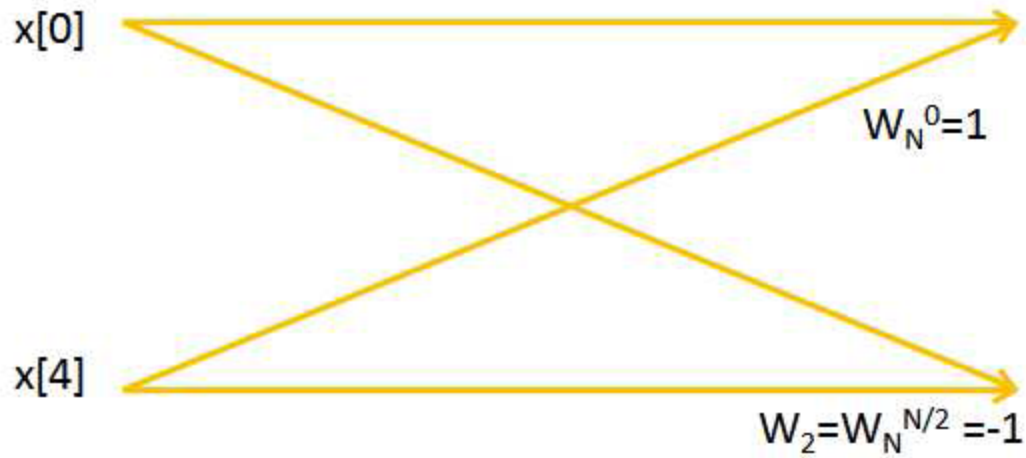
Flow graph



Flow Graph for a DFT with $N=4$



Flow graph



Flow Graph for a DFT with $N=2$



FFT algorithm

- The FFT is obtained by a **recursive algorithm** based on a **divide-et-impera strategy**

- **Fourier coefficients**

$$X[k] = \sum_{j=0}^{N/2-1} x[j] W_N^{kj}$$

$$x = (x[0], x[1], \dots, x[N - 1])$$



FFT algorithm

```
FFT-Ricorsiva(x)
  N = length(x);
  if N==1
    then return x[0]
  WN = exp(j 2 PI/N)
  W = 1
  Xp = FFT-Ricorsiva([x[0], x[2], x[4], . . . , x[N-2]])
  Xd = FFT-Ricorsiva([x[1], x[3], x[5], . . . , x[N-1]])
  for k = 0 to N/2 - 1
    do
      X[k] = Xp + W Xd
      X[k+N/2] = Xp - W Xd
      W = W WN
  return X
```



Time complexity

- The asymptotic time complexity is

$$T(N) = 2T(N/2) + \Theta(N) = \Theta(N \log N)$$

- It is the same also for the inverse transform



Convolution theorem

- A faster convolution can be obtained

$$a * b = \text{DFT}_{2N}^{-1} \left(\text{DFT}_{2N}(a) \text{DFT}_{2N}(b) \right)$$

zero padding

