



Multimedia Systems

Frequency domain

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Signal processing

- Signal analysis

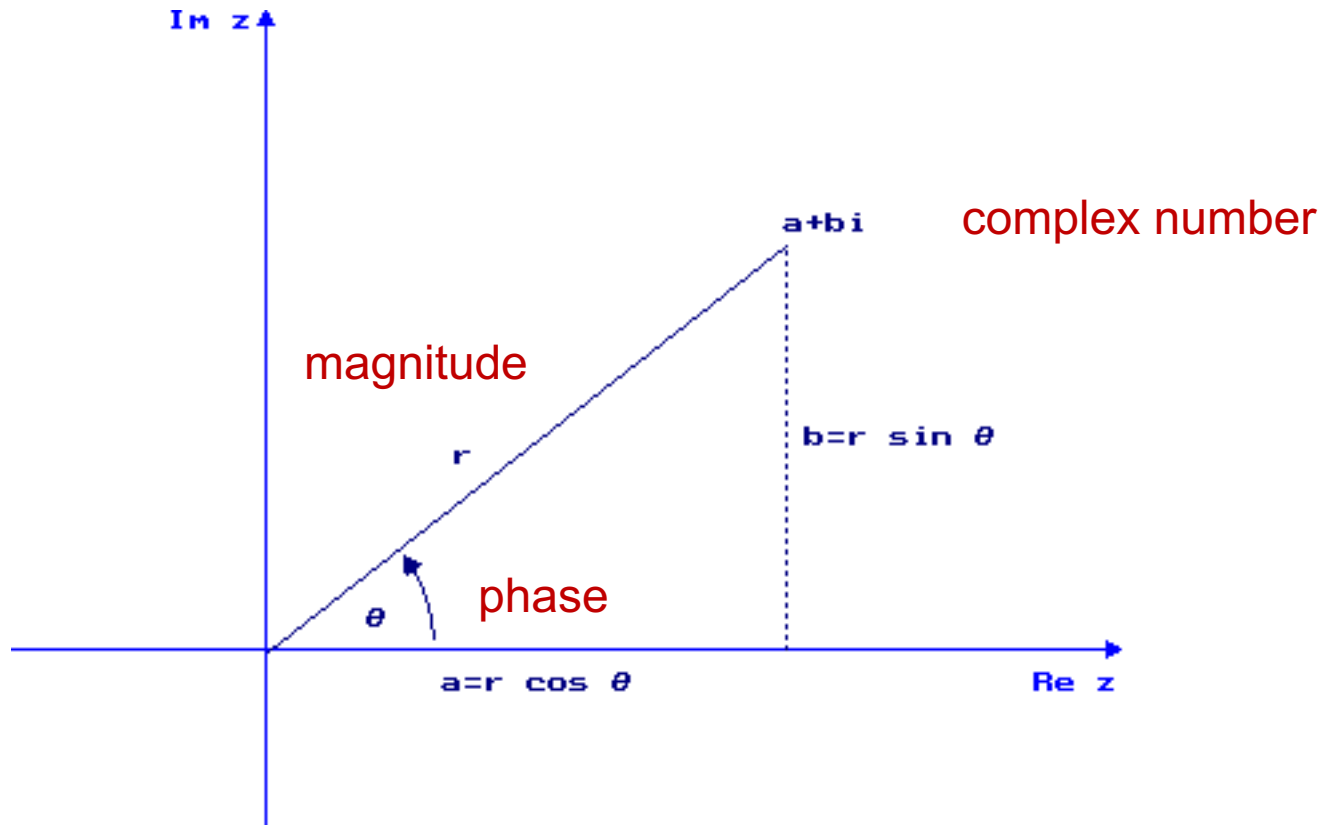
- time domain
- frequency domain

- Classical approaches

- Discrete Time Fourier Transform (DTFT)
- z Transform
- Discrete Fourier Transform (DFT)
 - FFT
- Discrete Cosine Transform (DCT)
- Cepstrum



Polar coordinate system



Polar Coordinates for Complex Numbers



Exponential operator

■ Euler formula

$$z = a + ib = r(\cos \theta + i \sin \theta) = re^{i\theta}$$

■ Magnitude

$$|z| = |a + ib| = \sqrt{a^2 + b^2} = r$$

■ Sequence

$$x(n) = Ae^{j(2\pi fn + \varphi_0)}$$

$$|x(n)| = |A| = \rho \quad \text{magnitude}$$

$$(2\pi fn + \varphi_0) = \mathcal{G} \quad \text{phase}$$

$$x(n) = Ae^{j(2\pi fn + \varphi_0)} = A(\cos(2\pi fn + \varphi_0) + j \sin(2\pi fn + \varphi_0))$$

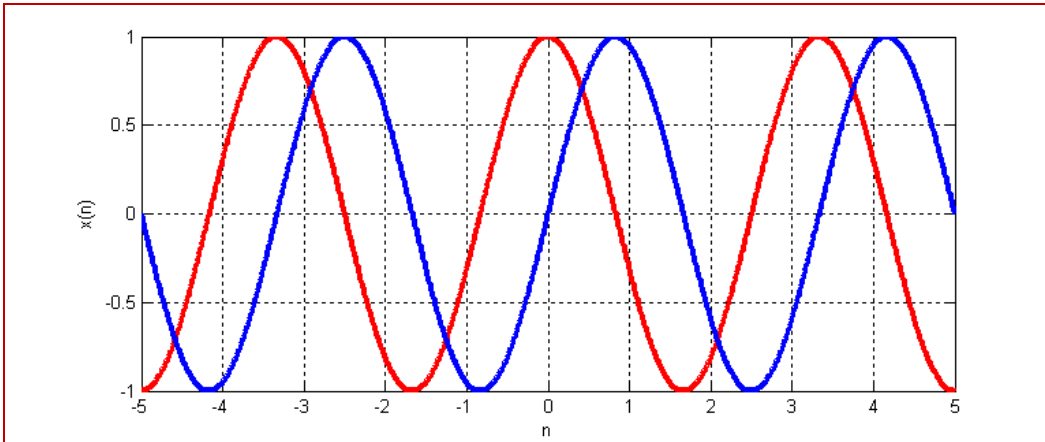


Sinusoidal content

$$x(n) = e^{j(2\pi 0.3n)}$$

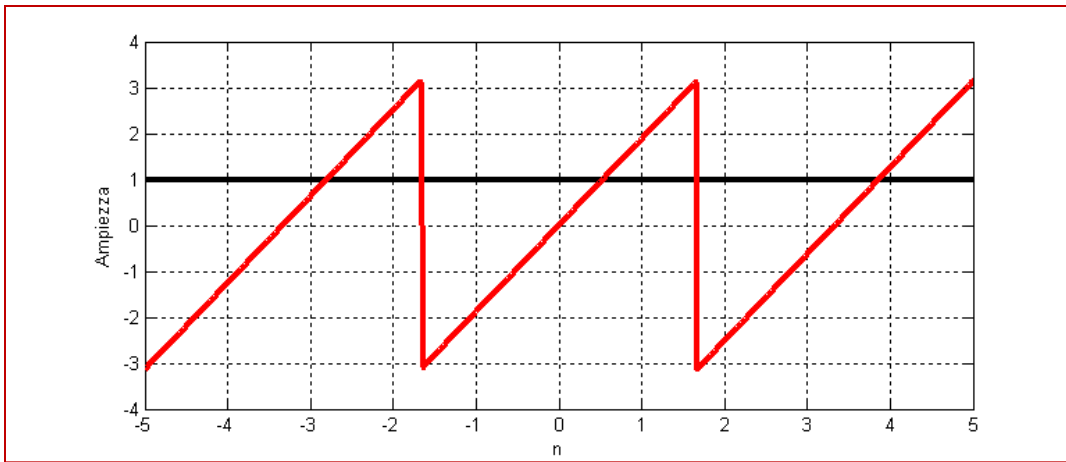
Real

Immaginary



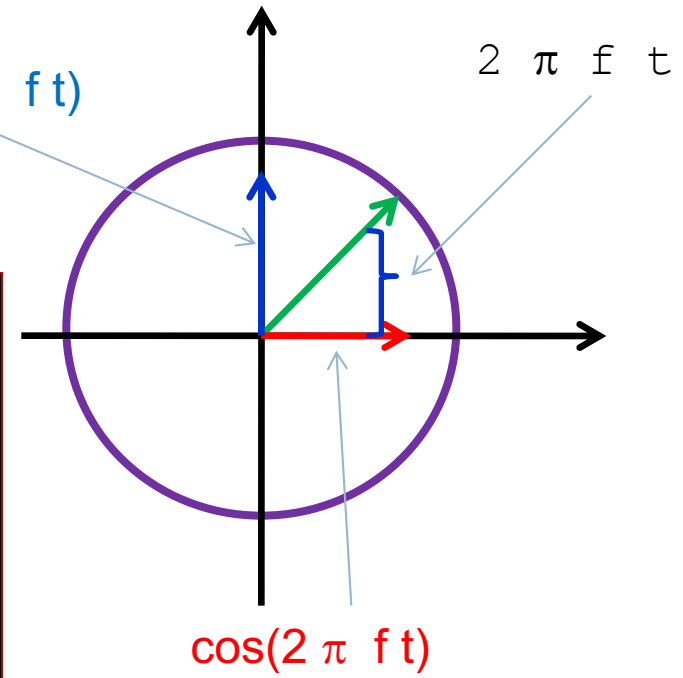
Phase

Amplitude



$\sin(2\pi f t)$

$2\pi f t$



Fourier theorem

■ Theorem

any continuous periodic signal can be obtained by the superposition of simple sine waves, each with its amplitude and phase, and whose frequencies are harmonics of the fundamental frequency of the signal

Continuous periodic signal with period T_0

$$x(t) = A_0 + \sum_{k=1}^{\infty} A_k \cos(2\pi k f_0 t + \theta_k) = A_0 + \sum_{k=1}^{\infty} A_k \cos(\omega_0 k t + \theta_k)$$

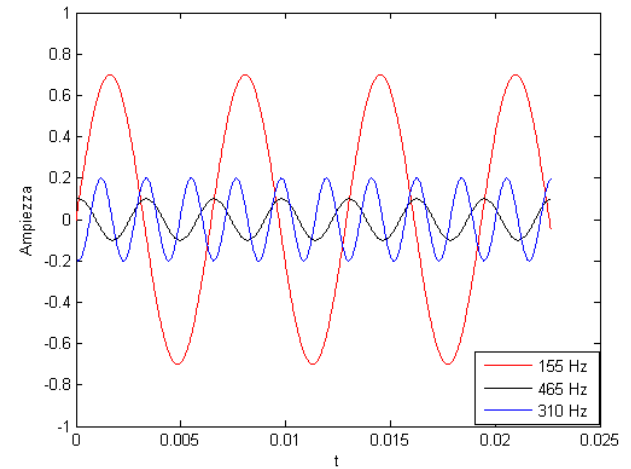
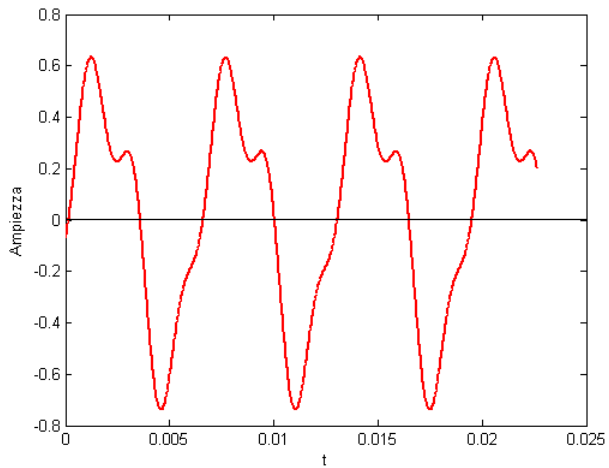
$$f_0 = 1/T_0$$

$$\omega_0 = 2\pi f_0$$

Fourier series



Fourier theorem



Fuorier example



Fourier transform

- Real environments
 - non-periodic signals

Inverse Fourier Transform
(Analysis)

$$x(t) = \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega = \int_{-\infty}^{\infty} X(f) e^{j2\pi f t} df.$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt.$$

$$X(f) df = A_k$$

Direct Fourier Transform
(Synthesis)



Discrete Time Fourier Transform

- For a discrete sequence $x(n)$

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n}$$

DTFT

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega)e^{j\omega n} d\omega$$

Trasformata di Fourier
a tempo discreto inversa



DTFT in C++

```
template <class Type, class Type_1>
class DTFT : public sequenze<Type>{
public:

    DTFT(double wa, double wb, int n):sequenze<Type>(n){

        t = vector<double>(size);
        x = vector<Type>(size);

        double dw = (wb-wa)/size;

        for (int k=0; k<size; k++)
            {
                t[k] = (double)wa + k*dw;
                x[k].put(0.0,0.0);
            }
    }

    void dtft( vector<Type> x, int L){ . . .}
    . . .

private:
    string Name;
};
```

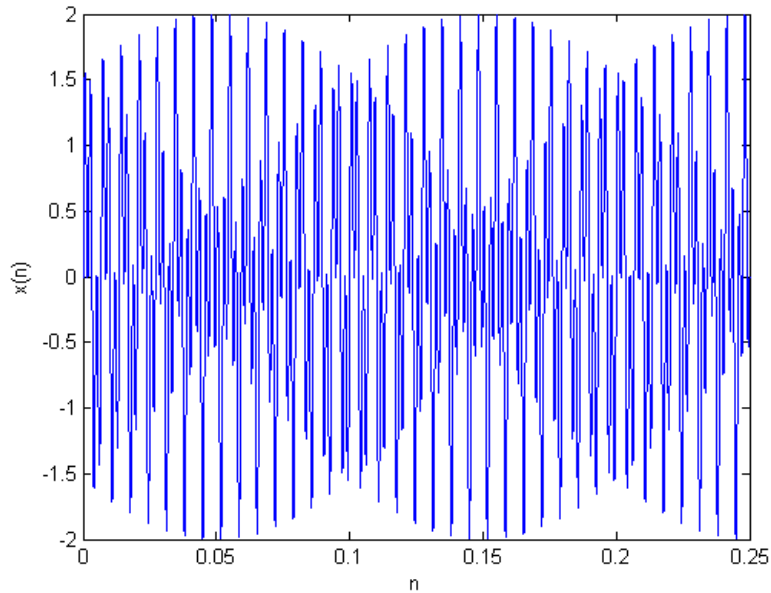


DTFT in C++

```
void dtft( vector<Type_1> x_i, int L) {  
  
    complex Z(0.0,0.0);  
  
    for (int k=0; k<size; k++)  
        {  
  
            Z.c_exp(1.0,-t[k]);  
for (int i=0; i<L; i++)  
            x[k].prod(Z,x[k]).add(complex(x_i[i],0.0));  
        }  
}
```



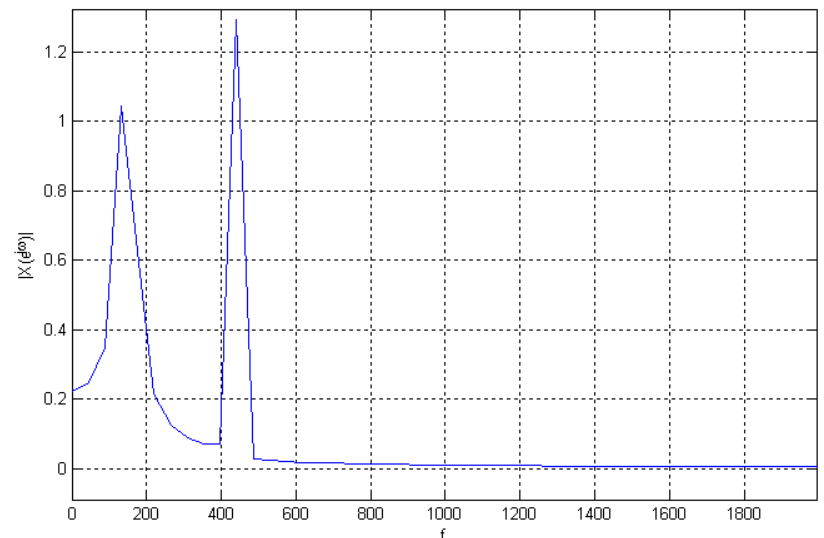
DTFT in C++



Sequence with two pure tones with 150 and 440 Hz, respectively. Sampling frequency of 44100 Hz.

Absolute value of the DTFT

$$0 \leq \omega \leq \pi \rightarrow 0 \leq f \leq 22050$$



Convolution theorem

$$x(n) \xleftrightarrow{\mathcal{F}} X(e^{j\omega})$$

$$h(n) \xleftrightarrow{\mathcal{F}} H(e^{j\omega})$$

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) = x(n) * h(n)$$

$$Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$

■ Highlighting

- A convolution in the time domain corresponds to a product in the frequency domain



Modulation theorem (windowing)

$$x(n) \xleftrightarrow{\mathcal{F}} X(e^{j\omega})$$

$$w(n) \xleftrightarrow{\mathcal{F}} W(e^{j\omega})$$

$$y(n) = x(n)w(n)$$



$$Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})W(e^{j(\omega-\theta)})d\theta$$

■ Highlighting

- The DTFT of the product of sequence corresponds to a periodic convolution of the single DTFTs
- e.g., FIR con windowing



Teorema di Parseval

$$x(n) \begin{array}{c} \xleftrightarrow{\mathcal{F}} \\ \xleftrightarrow{\quad} \end{array} X(e^{j\omega})$$

■ Energy

$$E = \sum_{n=-\infty}^{\infty} |x(n)|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

■ $|X(e^{j\omega})|^2$ Energy Spectral Density

