

Course of "Automatic Control Systems" 2023/24

Procedure for the controller design

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Procedure for the controller design

▲ The procedure for the controller design consists in three main steps:

♦ STEP 1

Convert the closed loop requirements in requirements on the open loop transfer function F(s)

♦ STEP 2

Design the controller K(s) so that the open loop function F(s) = K(s)G(s) satisfies the requirements in the STEP 1

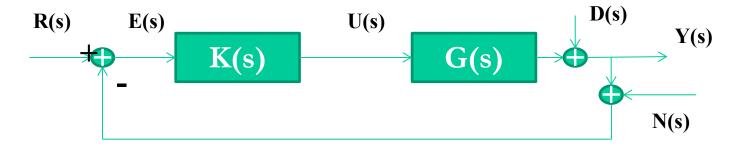
♦ STEP 3

Verify, with the aid of an appropriate software (for example MATLAB), that the closed loop system satisfies the requirements. Otherwise, go back to the STEP1 using more accurate techniques.



STEP 1: convert the requirements on T(s) into requirements on F(s)

- ▲ The performance of the closed loop system are evaluated in terms of
 - ♦ Tracking of the reference input
 - *♦* Rejection of the disturbs
 - **♦** Insensibility to the noise



- Assuming that the stability of the C.L. system is guaranteed, the responses of the system can be divided in a transient and a steady-state parts.
- The *steady-state performance* cares about the steady-state behavior of the closed loop system while the *transient performance* cares about the tracking of the reference signal during the transient phase

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STEP 1: Steady-state performance

- ▲ The steady-state performance are classified in
- \land Tracking of the reference input R(s)
 - ♦ Null or bounded steady-state error to polynomial inputs (step, ramp,...)
 - ♦ Null or bounded steady-state error to sinusoidal inputs at fixed frequency or to multi-frequency reference signal
- \land Rejection of the disturbs D(s)
 - ♦ Null or bounded steady-state error to polynomial inputs
 - ♣ Bounded steady-state error to multi-frequency sinusoidal inputs
- \land Insensibility to the noise N(s)
 - ♣ Bounded steady-state error to multi-frequency sinusoidal inputs



Step1: Steady-state performance

The requirements concerning polynomial reference and/or disturbs define the steady-state part of the controller

$$K'(s) = \frac{k_0}{s^{\mu}}$$

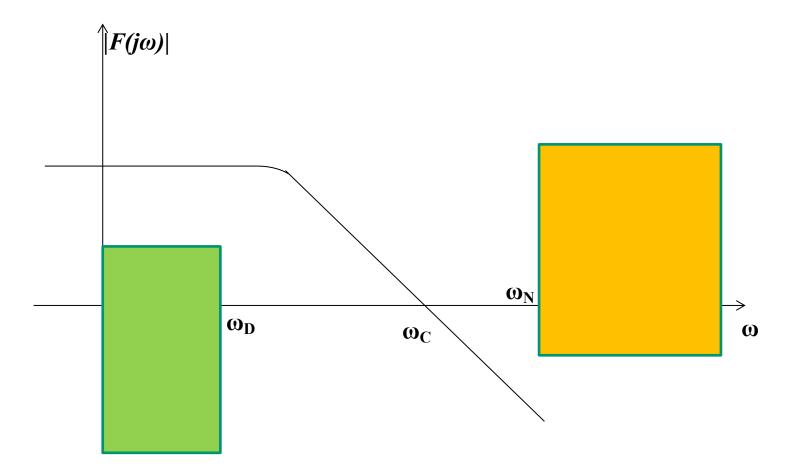
where k_0 and μ depend on

- ♦ the order of the polynomial input
- ♦ bounded or null error requirement
- * structure of the plan transfer function (gain and poles in the origin)



Step1: Steady-state performance

 $^{\perp}$ The requirements concerning *multi-frequency noise and disturbs* defines the set of the forbidden zones for the magnitude Bode diagram of F(s)



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Step 1: Transient performance

- The *transient performance* are usually expressed in terms of tracking properties of a polynomial reference of order 0 (step)
- ▲ The transient performance concerns
 - \Rightarrow Dynamic precision (overshoot) \longrightarrow Phase margin φ_m^*

$$\begin{cases} T_a(s) = \frac{1}{1+\tau s} & \varphi_m^* > 60^{\circ} \\ T_a(s) = \frac{1}{1+2\zeta s/\omega_n + s^2/\omega_n^2} & \varphi_m^* < 60^{\circ} \end{cases}$$

→ Time response (settling time)

$$\begin{cases} \omega_c^* = \frac{1}{\tau} & \varphi_m^* > 60^{\circ} \\ \omega_c^* = \omega_n & \varphi_m^* < 60^{\circ} \end{cases}$$

Crossing frequency ω_c^*



Example

Let us consider the following set of requirements:

- 1. $e_{\infty} = 0$ for a reference signal $r(t) = R_0 1(t)$
- 2. Attenuation $\geq 20_{db}$ for multi-frequency disturbs in the range $[0 \quad 0.01]$ rad/s
- 3. Attenuation $\geq 80\%$ for multi-frequency noise in the range [10 100] rad/s
- 4. Overshoot $s \le 10\%$
- 5. Settling time $t_{s5\%} \le 10s$



Example: steady-state spec. for polynomial reference signal

1. $e_{\infty} = 0$ for a reference signal $r(t) = R_0 \mathbf{1}(t)$

- A Steady-state requirement for polynomial reference signal.
- To assure a null steady state error for a polynomial signal of order 0 it is necessary that F(s) is of type 1, that is F(s) has a pole in the origin.
- Assuming that the plant transfer function G(s) doesn't contain poles in the origin, the steady-state part of the controller is

$$K'(s) = \frac{k_0}{s}$$

where k_0 is a free parameter

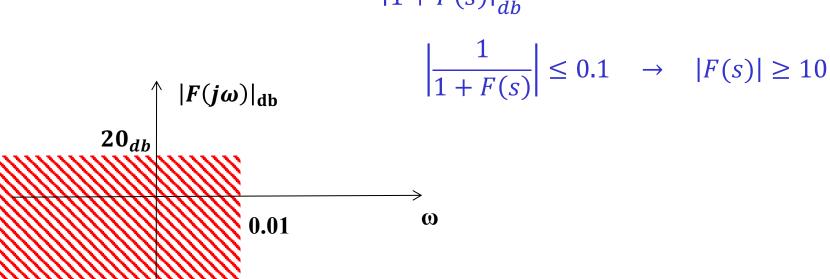


Example: steady-state spec. for multi-frequency disturbs

2. Attenuation $\geq 20_{db}$ for multi-frequency disturbs in the range $\begin{bmatrix} 0 & 0.01 \end{bmatrix}$ rad/s

- ▲ Steady-state requirement for multi-frequency disturb.
- ▲ It implies that, in the range [0 0.01] rad/s,

$$|S(s)|_{db} = \left|\frac{1}{1+F(s)}\right|_{db} \le -20 \quad \to$$



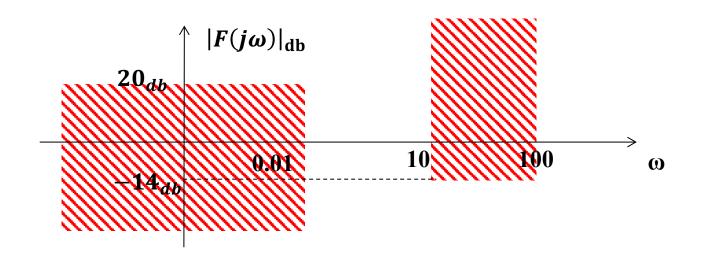


Example: steady-state spec. for multi-frequency noise

3. Attenuation $\geq 80\%$ for multi-frequency noise in the range [10 100] rad/s

- ▲ Steady-state requirement for multi-frequency noise
- ▲ It implies that, in the range [10 100] rad/s,

$$|T(s)| = \left| \frac{F(s)}{1 + F(s)} \right| \le 0.2 \quad \to \quad |F(s)| \le 0.2 \quad \to \quad |F(s)|_{db} \le -14$$





Example: transient spec. on the overshoot

4. Overshoot $s \leq 10\%$

▲ Transient requirement on the overshoot

A Taking into account that $s = e^{\sqrt{1-\zeta^2}}$, we have that

$$s \le 10\% \rightarrow \zeta \ge 0.6 \rightarrow \varphi_m \cong 100\zeta \ge 60^\circ$$

▲ Hence the complementary sensitivity function can be approximated by a first order system

$$T_a(s) = \frac{1}{1 + \tau s}$$

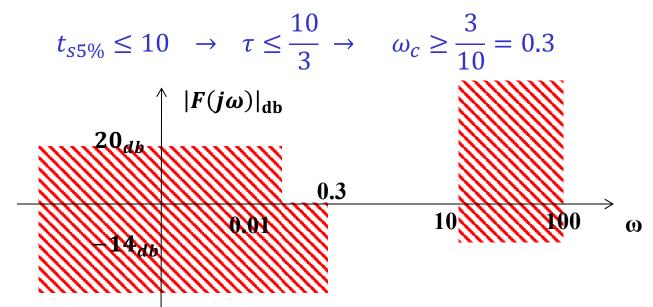
where τ depends on the settling time requirement



Example: transient spec. on the settling time

5. Settling time $t_{s5\%} \leq 10s$

- ▲ Transient requirement on the settling time
- A Taking into account that the settling time at 5% for a first order system is defined as $t_{s5\%} \cong 3\tau$, we have that



The transfer function F(s) should have a crossing frequency $\omega_c > 0.3$ and a phase margin $\varphi_m > 60^\circ$.



Step2: Controller design

- Design the controller K(s) so that the open loop function F(s) = K(s)G(s) satisfies the requirements in STEP 1
- ▲ The controller will be in the form

$$K(s) = K'(s) \cdot K''(s)$$

where

 $\star K'(s)$ have been designed according to the steady-state requirements concerning polynomial reference and/or disturbs

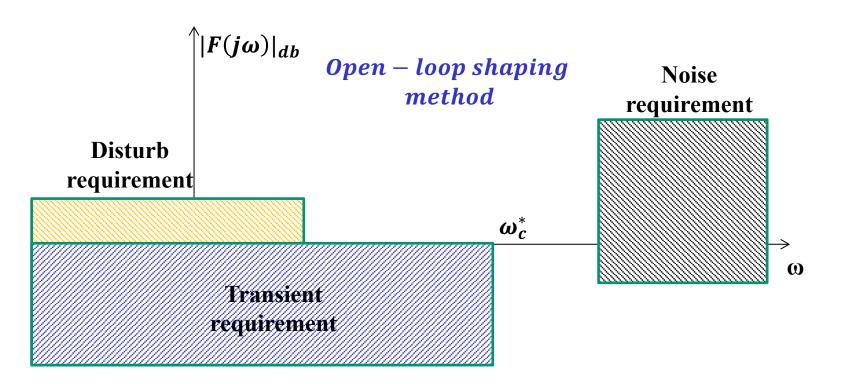
$$K'(s) = \frac{k_0}{s^{\mu}}$$

 $\star K''(s)$ have to be designed according to the steady-state multi-frequency requirements and the transient requirements



Step2: Controller design

 \wedge The control part K''(s) is usually designed so that F(s) doesn't intersect the forbidden zones



with
$$\angle F(j\omega_c^*) = \varphi_m^* - 180^\circ$$

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Step3: Validation of the controller

- ▲ If some of the requirements are not satisfied, more accuracy have been added in the design process:

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- ♦ Use the real Bode diagram instead of the asymptotic Bode diagram
- ♦ Use the Nichols chart to evaluate the desired phase margin
- ♦ Satisfy the requirements with a greater safety factor
- **♦**